

IC-701

HF ALL BAND ALL SOLID STATE
TRANSCEIVER

INSTRUCTION MANUAL

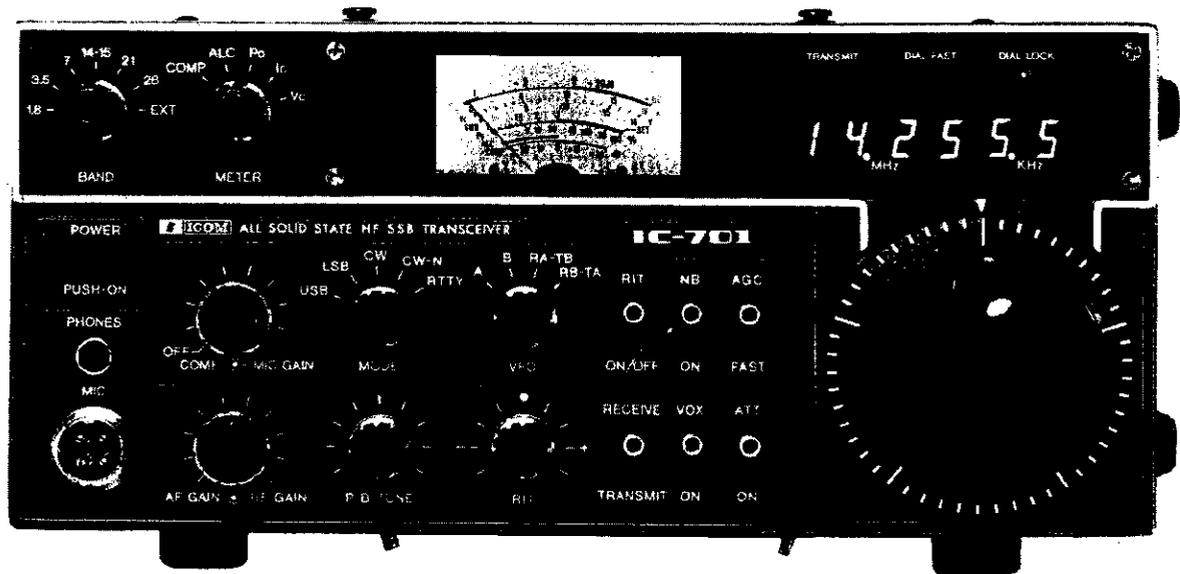


TABLE OF CONTENTS

SECTION 1	SPECIFICATIONS.....	1
SECTION 2	FEATURES.....	2
SECTION 3	INSTALLATION.....	3
SECTION 4	OPERATING CONTROLS.....	6
4-1	FRONT PANEL.....	6
4-2	CONTROLS UNDER ACCESS COVER.....	9
4-3	REAR PANEL CONNECTIONS.....	10
SECTION 5	OPERATING INSTRUCTIONS.....	11
5-1	HOW TO TUNE.....	11
5-2	SSB OPERATION.....	13
5-3	CW OPERATION.....	15
5-4	RTTY OPERATION.....	15
5-5	OTHER OPERATIONS.....	17
SECTION 6	THEORY.....	20
6-1	RECEIVER.....	20
6-2	TRANSMITTER.....	23
6-3	PLL (Phase Locked Loop) UNIT.....	26
6-4	DRIVER UNIT.....	28
6-5	POWER UNIT.....	32
6-6	OTHER CIRCUITRY.....	32
SECTION 7	MAINTENANCE AND ADJUSTMENT.....	34
7-1	MEASURING INSTRUMENTS REQUIRED FOR ADJUSTMENT.....	34
7-2	RECEIVER ADJUSTMENT.....	34
7-3	TRANSMITTER ADJUSTMENT.....	36
7-4	PLL UNIT ADJUSTMENT.....	38
SECTION 8	TROUBLE SHOOTING.....	40
SECTION 9	INSIDE VIEW.....	42
SECTION 10	VOLTAGE CHARTS.....	44
SECTION 11	BLOCK DIAGRAM.....	48

SECTION I SPECIFICATIONS

GENERAL

Number of Semi-Conductors:

Transistors	128
F E T	23
IC (Includes LSI)	56
Diodes	265

Frequency Coverage:

1.8MHz ~ 2.0MHz
3.5MHz ~ 4.0MHz
7.0MHz ~ 7.5MHz (7.3MHz~7.5MHz Receive Only)
14.0MHz~15.2MHz (14.35MHz~15.2MHz Receive Only)
21.0MHz~21.5MHz
28.0MHz~30.0MHz (28.0MHz~29.7MHz Transmit Range)

Frequency Control:

LSI based 100Hz step Digital PLL synthesizer.
Independent Transmit-Receive Frequency Available
on same band.

TRANSMITTER

DC Input Power:

SSB (A ₃ J)	200 Watts
CW (A ₁), RTTY (F ₁)	200 Watts

Continuously Adjustable Output power 0~Max.

Emission Mode:

A ₃ J	SSB (Upper sideband and Lower sideband)
A ₁	CW
F ₁	RTTY (Frequency Shift Keying)

Harmonic Output:

More than 40dB below peak power output

Spurious Output:

More than 60dB below peak power output

Carrier Suppression:

More than 40dB below peak power output

Unwanted Sideband:

More than 40dB down at 1000Hz AF input

Microphone:

Impedance 600 ohms
Input Level 10 millivolts typical
Dynamic or Electret Condenser Microphone

Frequency Readout:

6 digit LED 100Hz readout.

Frequency Stability:

Less than 500Hz after switch on 1 min to 60 mins, and
less than 100Hz after 1 hour. Less than 1KHz in the
range of -10°C to +60°C.

Power Supply Requirements:

DC 13.6V ±15% Negative ground Current drain
18 A max. (at 200W input)
AC power supply speaker console is available for AC
operation.

Antenna Impedance:

50 ohms Unbalanced

Weight:

7.3Kg

Dimensions:

111mm x 241mm x 311mm (H x W x L)

RECEIVER

Receiving System:

Triple Conversion Superheterodyne with continuous
Bandwidth Control

Receiving Mode:

A₁, A₃J (USB, LSB) F₁ (Output FSK audio signal)

IF Frequencies:

1st	9.0115MHz
2nd	10.75MHz
3rd	9.0115MHz

with continuous Bandwidth Control

Sensitivity:

Less than 0.25 microvolts for 10dB S+N/N

Selectivity:

SSB, RTTY	±1.2KHz at -6dB (Adjustable to ±0.5KHz Min)
	±2.0KHz at -60dB
CW	±250Hz at -6dB
	±700Hz at -60dB
CW-N	±100Hz at -6dB
	±500Hz at -60dB (with Audio Filter)

Spurious Response Rejection Ratio:

More than 60dB

Audio Output:

More than 1.5 Watts

Audio Output Impedance:

8 ohms

SECTION 2 FEATURES

ALL BAND, ALL MODE, ALL SOLID STATE

The IC-701 covers all the Amateur HF frequencies from 1.8MHz to 29.999.9MHz plus 15MHz WWV. It offers not only SSB, but also CW and RTTY. All of the circuits in the IC-701, including the driver and final power stages are completely solid state, and provide about 100 Watts output.

COMPUTER COMPATIBLE TUNING SYSTEM

The local oscillator circuit (VFO) employs a C-MOS LSI for the PLL that has been custom-made on the basis of ICOM's advanced digital technology. The VFO circuit is a digital PLL circuit that controls frequency determination by pulses produced by the tuning control. Unlike conventional PLLs, it controls the VCO by combining and dividing crystal oscillator frequencies. Therefore, its stability is much higher than conventional VFOs. In addition, the pulse control system makes it possible to set and change frequencies with external digital signals.

With a computer programmed and connected to the rear accessory socket, you can easily control frequencies, change bands, scan, on the IC-701. This system will give you the maximum of performance and versatility.

DUAL DIGITAL VFO'S

The IC-701 contains two built-in digital VFO's located in ICOM's custom LSI, providing the ultimate in versatility.

DIGITAL READOUT AND FULL METERING

The frequency is presented in bright easy to read LEDs, with no backlash or waiting time. The multi-function meter shows you relative signal strength in receive, power output, ALC, SWR, compression level, final transistor voltage and current. All the important information you need to know.

PASS BAND TUNING AND SPEECH PROCESSOR

The IC-701 has a built-in Pass Band Tuning system developed by ICOM that allows you to continuously adjust the pass band of the IF. By turning the control, you can eliminate interference from a nearby signal, thus providing clear reception. It can also be used as a tone control. During transmit, the Pass Band Tuning circuit can be used as a Speech Processor providing increased "talk power" for outstanding DXing.

OUTSTANDING PERFORMANCE

The RF amplifier circuit using MOS FETs, the Schottky barrier diode balanced mixer and other circuits provide excellent Cross Modulation and Two-Signal Selectivity characteristics. The IC-701 has excellent sensitivity demanded especially for mobile operation, high stability, and with two Crystal Filters having high shape factors, exceptional selectivity.

The Same Schottky diode mixer is used in the transmitter, and, with the band pass filter and high performance K-Constant two stage low-pass filter, ensures transmission of pure signals with very low spurious emission.

ADDITIONAL CIRCUITS

The IC-701 has a built-in Noise Blanker, VOX, an Auto Dimmer for control of the readout and meter illumination, CW Monitor, APC, SWR detector, and many other circuits for your convenience.

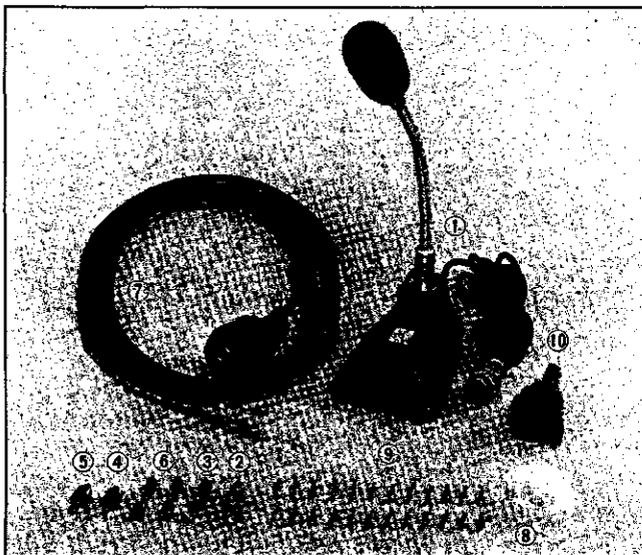
The IC-701 has everything you need to really enjoy HF operation, in an extremely compact, rugged transceiver.

SECTION 3 INSTALLATION

BE SURE TO READ THE FOLLOWING INSTRUCTIONS BEFORE USE.

3-1 UNPACKING

Carefully remove your transceiver from the packing carton and examine it for signs of shipping damage. Should any be apparent, notify the delivering carrier or dealer immediately, stating the full extent of the damage. It is recommended you keep the shipping cartons. In the event storage, moving, or reshipment becomes necessary they will be handy. Accessory cables, plugs, etc., are packed with the transceiver. Make sure you have not overlooked anything.



- | | |
|------------------------------------|----|
| 1. Microphone (IC-SM2) | 1 |
| 2. External Speaker Plug | 1 |
| 3. Key Plug | 1 |
| 4. Scope Terminal Plug | 1 |
| 5. Transverter terminal Plug | 1 |
| 6. Spare Fuse (20 Amp) | 2 |
| 7. DC Power Cord | 1 |
| 8. Accessory Plug (24 Pin) | 1 |
| 9. Accessory Plug Pins | 24 |
| 10. Accessory Plug Boot | 1 |

3-2 RECOMMENDATIONS FOR INSTALLATION

1. Avoid placing the IC-701 in direct sunlight, high temperature, dusty or humid places.
2. The left side of the unit, when facing it, functions also as a heatsink. The temperature there will usually become relatively warm during transmission. Any equipment should be at least 1 inch (3cm) away from the unit so as to provide good ventilation. Be sure that nothing is on and just behind the rear PA heatsink to ensure good ventilation. Also avoid places near outlets of heaters, air conditioners etc.
3. Place the unit so that the controls and switches can easily be handled and the frequency indication and meter can easily be read.
4. For mobile installation, an optional mounting bracket is available. Select the best location that can stand the weight of the unit and that does not interfere with your

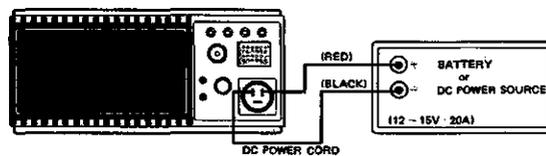
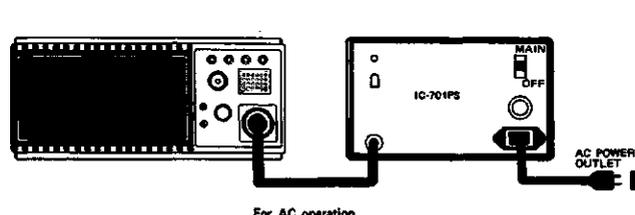
driving in any way.

5. Use the Ground Lug!

3-3 POWER SUPPLY

For AC operation, use the special power supply IC-701PS. If you would like to use your car battery or any other DC power supply, be sure that its output voltage is 12-15 Volts and the current capacity is at least 20 Amps. The maximum power consumption of the set during transmission runs from 16-20 Amps, so keep that in mind if the unit is installed in your automobile, and turn it on after you have started the engine. Attention should also be paid to the condition of the battery and electrical system.

The connection of the DC power cord supplied with the IC-701 is done in the following way: First make sure that the power switch of the unit is in the OFF position and the T/R switch is in the receive position. Connect the cord to the DC power supply with the RED lead to the positive terminal and the BLACK lead to the negative terminal. (Reverse connection will cause the protection circuit to operate and blow the fuse.) Connect the DC plug to the socket on the rear panel of the IC-701. Refer to the drawing below.



3-4 ANTENNA

Antennas play a very important role in radio communication. If the antenna is inferior, your transceiver cannot give you the best performance. With a good antenna and feeder cable having 50 ohm impedance, you should easily get the desired matching and performance. Carefully install a high performance antenna that suits the frequency band(s) you wish to operate on and place it as high as possible. Be especially careful of the condition of the connectors as loose connections will deteriorate the performance. Be sure to connect the ground terminal of a whip antenna, if used, to the body of your car.

As the output is quite high, avoid connecting the antenna connector to open lines and do not transmit under mismatched conditions. Otherwise the final stage could be overloaded and cause a malfunction of the unit.

3 - 9 EXTERNAL SPEAKER

The IC-701 contains an internal speaker, and is also designed so that it can drive an external speaker from the external (EXT) speaker jack on the rear panel. Be sure the impedance of the external speaker is 8 ohms, and remember that with the external speaker connected, the internal speaker is disabled.

3 - 10 HEADPHONES

Any good headphone set, including stereo type, that have 4-16 ohms impedance can be used. With the plug inserted halfway into the PHONES jack, both the headphone and speaker will operate. This is convenient when others wish to listen in on the station, or you wish to record contacts using a tape recorder connected to the headphone jack. With a stereo headphone set inserted this way, however, the headphone will lose the sound on one side. With the plug inserted completely, only the headphone works.

3 - 11 POWER SUPPLY

It is highly recommended that you use the IC-701PS power supply for base operation. If you wish to use another power source, make sure that it meets the voltage, current, and regulation requirements. Pay particular attention to the overvoltage protection as a runaway regulator can destroy the IC-701. Make certain that more than 18 Volts does not get applied to the unit. When the power switch is in the ON position, do not connect the power cord, antenna, external speaker, Accessory plug, or the microphone to avoid possible trouble. Be sure not to transmit with the antenna disconnected. If the fuse blows, replace it with a new 20 Amp fuse after fixing the cause. Avoid turning the power switch ON and OFF repeatedly over a short time. Frequent repetitions may cause the frequency indicator to display the wrong frequencies. If such a malfunction should occur, turn OFF the switch and a few seconds later turn it ON again.

3 - 12 CAUTIONS

As the unit has already been closely adjusted with highly sophisticated measuring instruments, never tamper with the turnable resistors, coils, trimmers, etc.

IN THE PLL UNIT, THE HEART OF THE TRANSCIEIVER, A C-MOS LSI IS INCORPORATED. IF THE SEAL IS BROKEN WHEN A FAILURE OCCURS WITHIN THE TERMS OF THE GUARANTEE, THE REPAIR MAY BE AT THE USERS EXPENSE, SO DO NOT ATTEMPT TO OPEN THE PLL CASE!

C-MOS is also used in the digital driver unit as well as the PLL. C-MOS ICs are very susceptible to excessive static charges and over current and care must be used when handling them. Therefore, avoid touching the driver unit and the nearby circuitry unless absolutely necessary. When it is necessary to check the circuitry, observe the following points.

Ground all measuring instruments, the soldering iron, and other tools. Do not connect or disconnect the C-MOS IC from its socket, or solder it when the power is ON. Do

not apply voltage of less than -0.5 or more than $+5$ Volts to the input terminals of the IC. DO NOT MEASURE WITH AN OHMMETER.

3 - 13 COOLING FAN

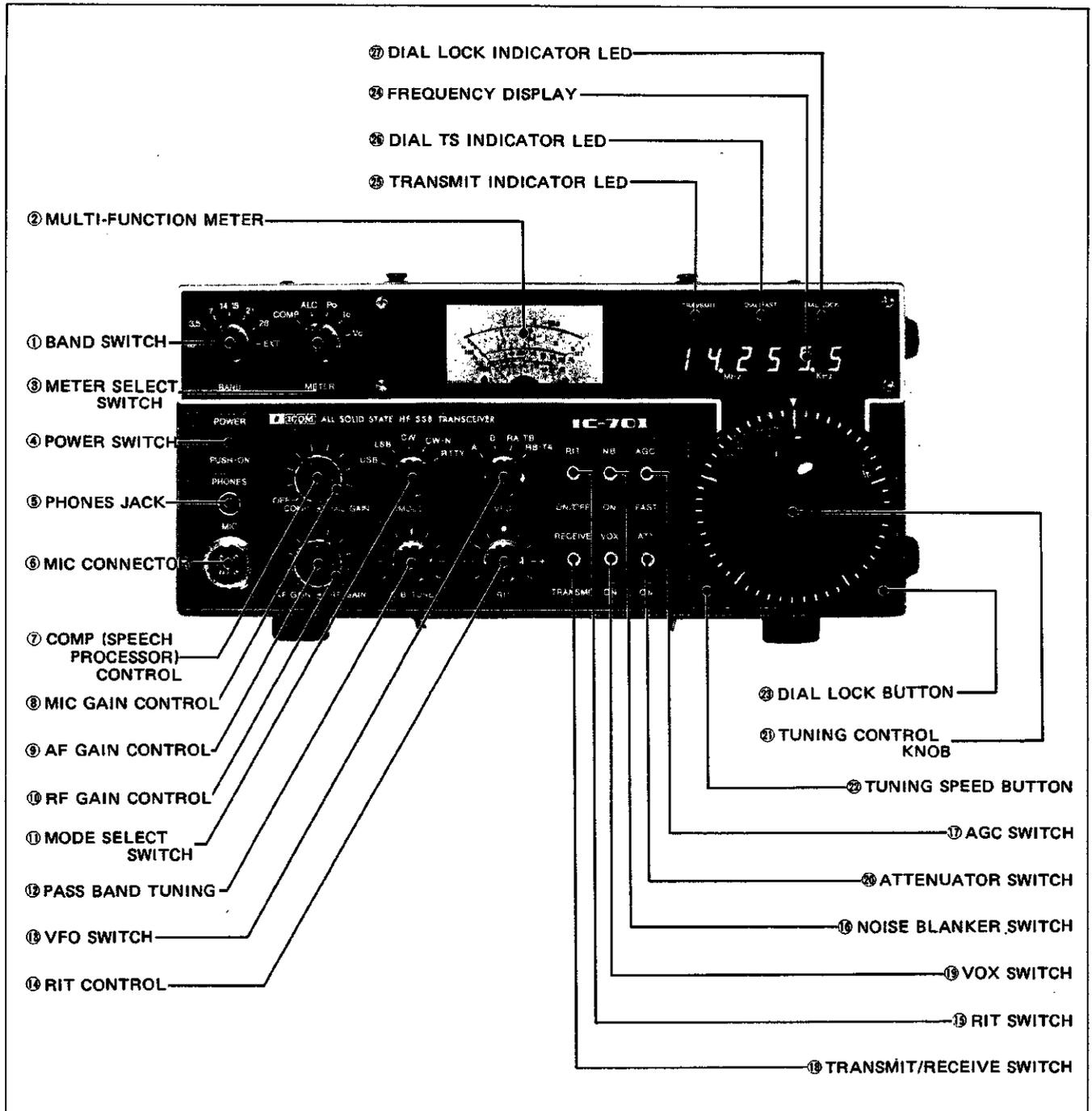
The rear of the PA unit is designed to achieve effective natural cooling. But with 200 Watts input, the final stage circuit produces quite a bit of heat and its temperature may rise during a very long, continuous transmission. If the temperature becomes high enough, the built-in cooling fan will start to spin. The fan is connected to the temperature monitoring circuit which detects the temperature of the final stage transistors. The operation of this fan is as follows:

1. The fan does not work during all reception.
2. During transmission, it does not work if the detected temperature is below a certain level.
3. When the temperature exceeds that level during a transmission, the fan will start to move air across the heatsink to cool it.
4. If the temperature of the final stage transistors rises to the danger point, the fan will turn at a much higher speed and frequency display will start to flash.
THIS MEANS STOP TRANSMITTING!

1, 2 and 3 are normal conditions. 4 is abnormal. If the meter and display start to flash, stop transmitting and check for the cause of the excess heat. This could be antenna mismatch, environmental problems, etc., but be sure to rectify the problem before beginning to operate again.

SECTION 4 OPERATING CONTROLS

4-1 FRONT PANEL



1. BAND SWITCH

The BAND SELECT SWITCH is a 7 position rotary switch used for selecting one of the 6 Amateur bands covered by the IC-701 with a 7th position for external control. The selectable bands are 1.8MHz, 3.5MHz, 7MHz, 14MHz, 21MHz and 28MHz. In the EXT position, the band can be selected remotely. However, the switch should not be put in the EXT position without having the remote control connected. This will cause the rotary switch to rotate continuously and may cause excessive wear or damage.

2. MULTI-FUNCTION METER

With the IC-701 in the receive mode, this meter acts as an S meter regardless of the position of the meter select

switch. Signal strength is indicated on a scale of S1-S9 and S9+20dB to S9+60dB.

In the transmit mode, the meter has 5 functions as outlined below which are selected by the meter select switch.

3. METER SELECT SWITCH

In the transmit mode the meter has 5 functions.

1. Vc Indicates the collector voltage of the final transistors.
2. Ic Indicates the collector current of the final transistors.
3. Po Indicates the relative output power. SWR can

be measured by placing the SWR/SWR SET switch in the SWR position. (This switch is located under the access cover on the top of the IC-701.

4. ALC In this position, the meter begins to function when the RF output power reaches a certain level.
5. COMP Indicates the compression level when the speech processor is in use.

4. POWER SWITCH

The POWER SWITCH is a push-lock type switch which controls the input DC power to the IC-701. When the external AC power supply (IC-701PS) is used, the switch also acts as the AC power supply switch. When the switch is pushed in and locked, power is supplied to the set. When the switch is pushed again and released, power is cut to all circuits except the PA unit.

5. PHONE JACK

This jack accepts a standard 1/4 inch headphone plug for a 4-16 ohm headphone set. Stereo headphones can be used without modification.

6. MIC CONNECTOR

Connect the supplied microphone or optional microphone to this jack. If you wish to use a different microphone than the IC-SM2, refer to the drawing on page 4. Microphone impedance should be 500-600 ohms.

7. COMP (SPEECH PROCESSOR) CONTROL

The IC-701 contains a built-in RF speech processor which is turned ON by turning the center knob clockwise from the OFF position (a click will be heard). In the ON position it works as the RF drive control as explained below. In the CW and RTTY modes, it works as the output level control of the IC-701. When RF speech processor operation is not desired, the COMP control should be turned completely counter-clockwise and clicked OFF.

8. MIC GAIN CONTROL

Adjusts the level of modulation according to the input of the microphone. Clockwise rotation increases microphone gain. As the input will vary with different microphones and different voices, the knob should be turned until the MULTI-FUNCTION METER needle, in the ALC mode, starts to move slightly within the ALC zone. In the SSB mode, with the RF speech processor in use, the MIC GAIN CONTROL sets the clipping limits, while the COMP control sets the RF drive level up to the maximum power level where ALC begins at the saturation point of the amplifiers.

9. AF GAIN CONTROL

Controls the audio output level in the receive mode. Clockwise rotation increases audio output.

10. RF GAIN CONTROL

Controls the gain of the RF section in the receive mode. Clockwise rotation gives the maximum gain. As the control is rotated counter-clockwise, the needle of the

MULTI-FUNCTION METER rises, and only signals stronger than the level indicated by the needle will be heard.

11. MODE SELECT SWITCH

This switch selects the mode of operation for both transmit and receive.

- USB Upper Sideband, mainly for 14, 21 and 28MHz bands.
- LSB Lower Sideband, mainly for 1.8, 3.5 and 7MHz bands.
- CW Continuous Wave, for CW operation on all bands.
- CW-N Narrow CW. The audio band-pass filter is automatically turned ON in this position to improve selectivity when receiving.
- RTTY Radio Teletype. For use on all available bands.

12. P. B. TUNE (PASS BAND TUNING) CONTROL

This control allows continuous tuning of the pass band selectivity moving the edge of the filter up to 1.5KHz from either the upper or lower side in SSB and RTTY. This not only improves selectivity, but also can improve the audio quality. The normal position is in the center (12:00 o'clock) position and is 2.4KHz wide in SSB.

13. VFO SWITCH

With this switch you can select the VFO you want to operate from two separate VFO's that are built-in to the IC-701. It also selects the relationship of the two VFO's. The switch performs the following operations according to its position.

- * A : Selects "A" VFO for both transmit and receive.
- * B : Selects "B" VFO for both transmit and receive.
- * RA-TB : Selects "A" VFO for receive and "B" VFO for transmit.
- * RB-TA : Selects "B" VFO for receive and "A" VFO for transmit.

In addition, when the VFO is switched from "A" VFO to "B" VFO, the frequency indicated on the frequency display just prior to switching goes into a memory inside the LSI. Thus even if "B" VFO is being used, switching to "A" again will enable you to operate at the initial frequency. Switching from "B" to "A" results in the same operation. However, if switching is done with the BAND SWITCH (or external control), the VFO frequency displayed will disappear and the lower end frequency of the band selected will be displayed. If you wish to operate on the frequency that was memorized in the VFO that was not used when switching bands, simply return to the original band and then to the other VFO and the initial memorized frequency will appear. This is explained in more detail later in this manual.

14. RIT CONTROL

Shifts the receiver frequency ± 800 Hz either side of the transmit frequency. When the RIT is ON, the RIT LED is illuminated. The LED also represents the "0" position. Rotating the control to the (+) side raises the receiver frequency, and rotating to the (-) side lowers the receiver frequency. With the RIT ON, if the tuning knob (21) is moved one increment, the RIT circuit is automatically pulsed OFF. Therefore it is not necessary to manually turn

OFF the RIT when changing operating frequency. The frequency shift by tuning the RIT Control is not indicated on the frequency display.

15. RIT SWITCH

This is a spring-loaded switch for the RIT control. To turn ON the RIT, push down once. If OFF is desired, push down again. When the RIT is ON the LED above the RIT control knob is lit. This also indicates the center position of the RIT control. With the RIT circuit OFF the LED is also OFF. Also if the RIT is ON and the Tuning Knob is rotated one increment it will pulse the RIT OFF unless the internal jumper has been changed.

16. NB (NOISE BLANKER) SWITCH

When pulse type noise such as automobile ignition noise is present, set this switch to the ON (down) position. The noise will be reduced to provide comfortable reception.

17. AGC (AUTOMATIC GAIN CONTROL) SWITCH

With this switch you can change the time constant on the AGC circuit. With the switch in the AGC (up) position, the circuit operates as a "hang" AGC and maintains the AGC voltage for a specific time, thus it is suitable for SSB reception. With the switch in the FAST (down) position, the circuit is OFF, and AGC is suitable for stations suffering from fast fading or operating in the CW mode.

18. TRANSMIT/RECEIVE (T/R) SWITCH

This switch is for manually switching from transmit to receive and vice versa. Set the switch to RECEIVE (up) and the IC-701 is in the receive mode. Set the switch to TRANSMIT (down) and it switches to transmit. When switching with the PTT switch on the microphone or with the VOX switch set to ON the T/R switch must be in the RECEIVE position.

19. VOX SWITCH

This switches the VOX circuit ON and OFF. When it is in the ON (down) position, in SSB, T/R switching is accomplished by means of a voice signal. In CW operation, semi-break-in switching by means of keying is possible.

20. ATT (ATTENUATOR) SWITCH

Turn the ATT switch ON (down) when strong nearby signals disturb signal reception or make "S" readings difficult. In the ON position, a 10dB attenuator is inserted into the receiving antenna circuit, reducing interfering signals and giving more stable reception. In normal operation the ATT switch is left in the OFF (up) position.

21. TUNING CONTROL KNOB

Rotating the TUNING CONTROL KNOB clockwise increases the frequency, while rotating it counter-clockwise decreases the frequency. The frequency is changed in 100Hz steps which correspond to the smaller vernier marks on the knob. The larger marks represent 1KHz steps. One complete rotation of the tuning knob results in a 5KHz frequency increase or decrease in 100Hz steps exactly as shown by the small vernier marks.

22. TUNING SPEED BUTTON

Pushing the TUNING SPEED BUTTON will illuminate the DIAL FAST display LED, and the small vernier marks on the tuning knob are changed to correspond to 10KHz steps. At the same time, the 1KHz and 100Hz digits are cleared on the display to show "0.0". When the Tuning Speed Button is depressed again, the dial fast display LED goes OFF and the small vernier marks will again correspond to 100Hz steps. This button will allow you to quickly OSY over a great frequency range.

23. DIAL LOCK BUTTON

After the IC-701 is set to a certain frequency for rag chewing, mobile operation, etc., pushing the DIAL LOCK BUTTON will illuminate the DIAL LOCK display LED and the VFO is electronically locked at the displayed frequency, thus inactivating the operation of the tuning knob. To change frequency, the dial lock must first be disengaged by pushing the dial lock button again. The dial lock display LED will go OFF and the frequency will change with the direction of the tuning knob rotation. Turning the power OFF will release the dial lock and clear the displayed frequency.

24. FREQUENCY DISPLAY

The frequency of the IC-701 is displayed on 7 segment LEDs. Since the 1MHz and 1KHz decimal points are displayed, the frequency can easily be read. The frequencies indicated are the carrier frequencies of each mode in USB, LSB and CW. In the RTTY mode, the space frequency is displayed. Since the local oscillation frequency is shifted in accordance with modes, no tuning will be required if you change modes. Remember, if you turn the RIT control ON to change the receive frequency and rotate the RIT control knob, the frequency displayed will not change.

25. TRANSMIT INDICATOR LED

When your set is in the transmit mode, this LED is lit.

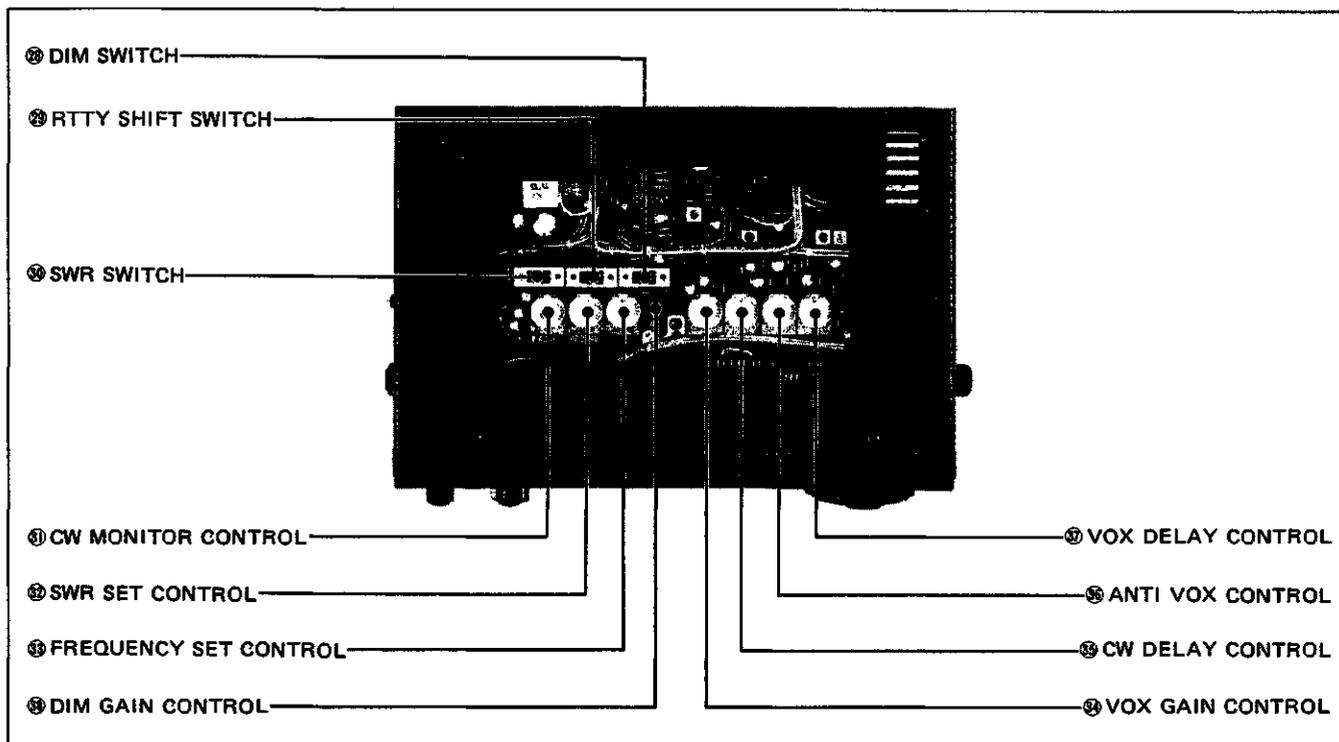
26. DIAL TUNING SPEED INDICATOR LED

Illuminates when the TUNING SPEED BUTTON is pressed to set the dial to fast tuning.

27. DIAL LOCK INDICATOR LED

Illuminates when the DIAL LOCK BUTTON is pressed to lock the dial.

4 - 2 CONTROLS UNDER ACCESS COVER



28. DIM SWITCH

When this switch is set to DIM, the luminosity of the frequency display and the meter automatically decreases when the ambient level is reduced below a set level. Therefore, under dim lighting such as at night, it decreases the luminosity to a lower level for easy indicator reading. When the switch is at the OFF position, the dimmer function is disabled.

29. RTTY SHIFT SWITCH

This switch shifts transmit frequency and shift spacing. Normally the shift is 170Hz, but in some cases 850Hz is still used. Set the switch for the desired shift.

30. SWR SWITCH

When measuring SWR, calibration SET and SWR reading functions are selected with this switch. Remember, at the factory, the switch is in the SET position and fixed in place with a plastic screw to prevent it from switching to SWR. Remove the screw before attempting to read SWR! SWR must be read only in the Po meter position.

31. CW MONITOR (MONI) CONTROL

This control adjusts the audio volume of the side tone (monitor) audio during CW transmit operation. Adjust it to your desired level for easy listening.

32. SWR SET CONTROL

This control calibrates the meter needle to the Po SET position when you want to determine the value of SWR. The meter must be in the Po position and the set must be transmitting a carrier.

33. FREQUENCY SET CONTROL

This control is for fine adjustment of the VXO section frequency of the PLL unit, which is the local oscillator frequency. Do not turn it unless you want to change the frequency.

34. VOX GAIN CONTROL

This control adjusts input signal level via the microphone to the VOX circuit. For VOX operation in SSB, adjust the control so that the VOX circuit will operate with normal speech.

35. CW DELAY (CW time delay constant) CONTROL
In Semi-break-in CW operation, this controls the TX/RX switchover time delay. Adjust it to suit your keying speed.

36. ANTI VOX CONTROL

In VOX (SSB) operation, the VOX circuit may be operated by sound from the speaker causing a switch to transmit. This trouble can be prevented by adjusting the input level of the ANTI VOX circuit with this control along with the VOX gain control so that the VOX circuit only operates from the operator's voice, not by sound from the speaker.

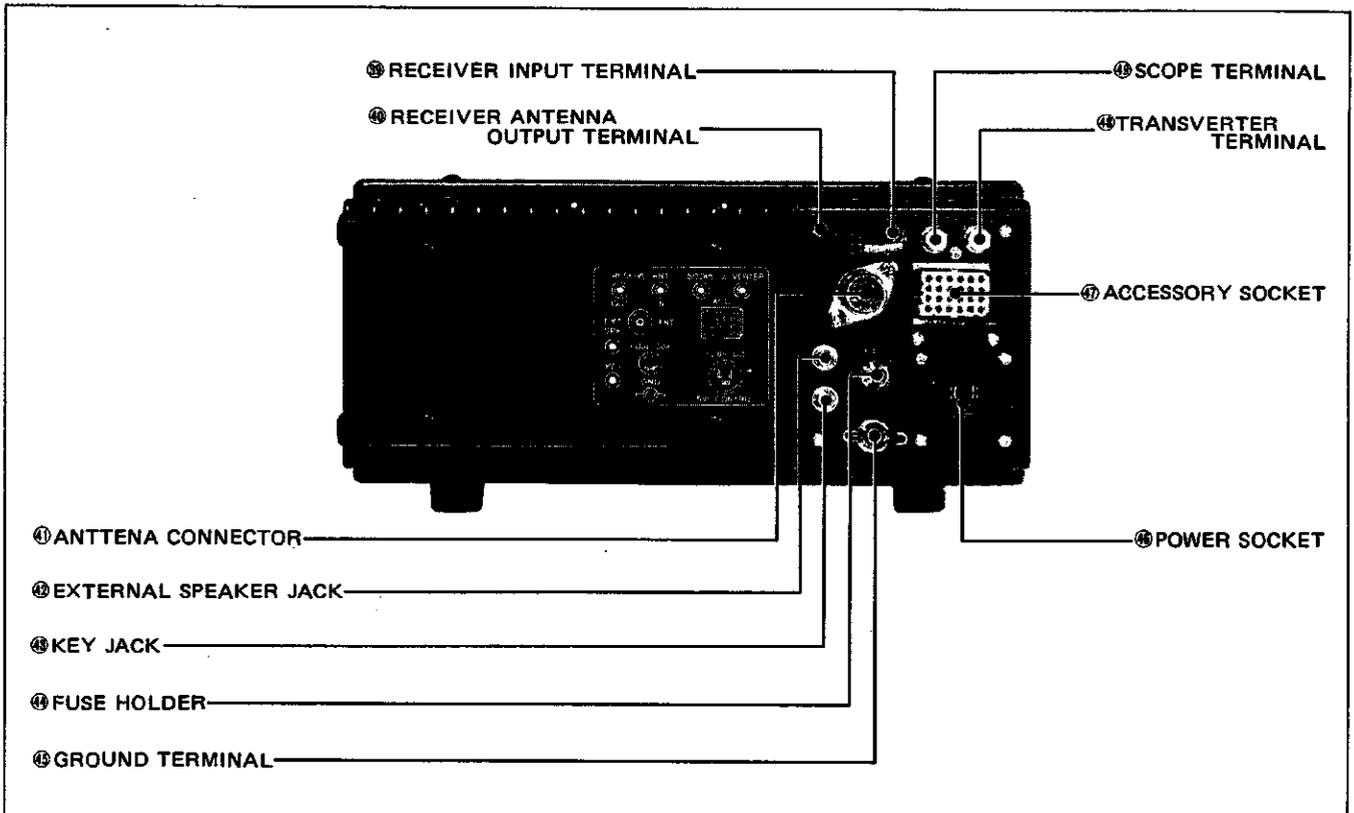
37. VOX DELAY (VOX time constant) CONTROL

This controls the TX to RX switching time. Adjust it so TX to RX switching will not occur during short pauses in normal speech.

38. DIM GAIN (dimmer circuit) CONTROL

The threshold for the functioning of the dimmer circuit can be adjusted by turning the DIM GAIN CONTROL. Turning the control counter-clockwise will lower the operation threshold of the circuit.

4-3 REAR PANEL CONNECTIONS



39. RECEIVER INPUT TERMINAL

This is an input terminal which is connected directly to the receiver.

40. RECEIVER ANTENNA OUTPUT TERMINAL

This is a terminal to which received signals from the antenna connector are conducted after the signal passes through the transmit/receive antenna switching circuit. Usually the receiver IN and OUT terminals are jumpered. The receiver antenna output terminal is usually used when another receiver is used or a pre-amplifier is connected to the IC-701.

41. ANTENNA (ANT) CONNECTOR

This is used to connect the antenna to the set. Its impedance is 50 ohms and connects with a PL 259 connector.

42. EXTERNAL SPEAKER JACK

When an external speaker is used, connect it to this jack. Use a speaker with an impedance of 8 ohms. When the external speaker is connected, the built-in speaker does not function.

43. KEY JACK

For CW operation, connect the key here. For electronic keying the terminal voltage must be less than 0.4V DC.

44. FUSE HOLDER

This holds the fuse for the DC power circuit. If the fuse is blown, replace it with a new 20 Amp fuse after checking the cause. Open the fuse holder with a Phillips head driver.

45. GROUND TERMINAL

To prevent electrical shock, TVI, BCI and other problems, be sure to ground the equipment through the GROUND TERMINAL. For best results use as heavy a gauge wire or strap as possible and make the connection as short as possible, even in mobile installations.

46. POWER SOCKET

Connect the DC power cord or cable from the IC-701PS or suitable DC source to this connector.

47. ACC (ACCESSORY) SOCKET

Accessories and controls can be used with the IC-701 and connected to the ACC socket on the rear panel. NOTE: Pins 14-24 are terminals for digital control and are connected to the input terminal of the C-MOS IC. By no means apply any voltage not within the range of -0.5 to +5 Volts. For details, refer to page 20.

48. X-VERTER (TRANSVERTER) TERMINAL

VHF and UHF operation using a suitable transverter with the IC-701 is possible. This terminal is for Transverter connection. The output is a few milliwatts. Refer to the transverters instruction for recommended hookup.

49. SCOPE TERMINAL

This terminal brings out the 9.011MHz IF signal from the mixer in the receiver. Not only observation of the received signal, but also of signals of a the selected band width are possible by using a panadaptor or panascope.

SECTION 5 OPERATING INSTRUCTIONS

5-1 HOW TO TUNE

The following instructions are for tuning in any mode. Please read carefully and understand fully before turning on your unit. Proper tuning is necessary for optimum operation.

5-1-1 TUNING KNOB

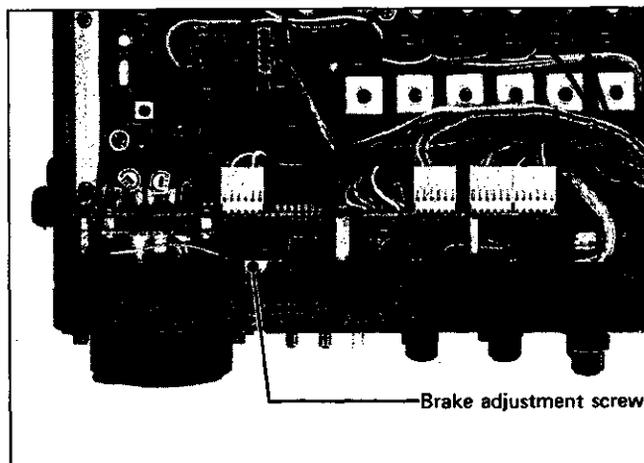
The transmit or receive frequency is displayed on 7 segment LEDs down to 100Hz digits. Rotating the tuning knob clockwise increases the frequency, while turning counterclockwise decreases the frequency in 100Hz steps. The smaller vernier marks on the knob represent 100Hz and the larger marks represent 1KHz. One complete rotation of the tuning knob results in a 5KHz frequency increase or decrease. When you receive your set, calibration of the tuning knob may be desirable, as the large increments may not correspond exactly to a change of 1KHz displayed on the LEDs. Exact calibration is not necessary as the displayed LED frequency will always be correct, but you may calibrate the scale on the tuning knob for use as an analog type frequency display. To calibrate, set a large increment on the tuning scale to line up with the indicator arrow directly above the scale and then turn ON the power.

To calibrate the scale during operation, if desired, set a large increment on the tuning scale to line up with the indicator arrow and then push the tuning speed button TWICE. By doing so, the 1KHz and 100Hz digits are cleared on the display to show "0.0".

The tuning knob scale may require recalibration if you:

1. Turn the knob while the frequency is locked by use of the Dial Lock Button.
2. Use the tuning speed button when a large increment is not lined up with the indicator.
3. Turn the knob beyond the band edge.

If the knob is too loose or too stiff for comfortable use, you can adjust the torque by tightening or loosening the brake adjustment screw accessible from underneath the set.



Brake Adjustment

The following instructions should be used to adjust the tension of the Tuning Knob.

1. Remove the bottom cover of the IC-701.
2. The Tuning knob tension will become tighter by turning the brake adjustment screw clockwise, and will become looser by turning the screw counterclockwise.
3. While performing this adjustment, the Tuning knob must be turned continuously as the screw is adjusted in order to set the tension for a comfortable touch.
4. After the adjustment has been made, replace the bottom cover.

On all but the 28MHz band, the displayed frequency will not go beyond the highest band edge even by turning the tuning knob clockwise, or beyond the lowest band edge by turning the knob counter-clockwise. On the 28MHz band, since this band width is much wider than the others, the IC-701 provides "automatic revert" for operating convenience. When the frequency display shows 29.999.9MHz and the tuning knob is turned clockwise, the displayed frequency jumps down to 28.000.0MHz and continues up from there. Likewise, by continuing to turn counter-clockwise beyond the bottom edge of the band, the frequency jumps up from 28.000.0MHz to 29.999.9MHz and goes down from that point.

The displayed frequency shows the carrier frequency in the USB, LSB and CW modes, and the SPACE frequency in the RTTY mode. Therefore recalibration of the frequency is unnecessary when the mode is changed.

5-1-2 DIAL LOCK BUTTON

After the IC-701 is set at a certain frequency for rag chewing, mobile operation, etc., pushing the Dial Lock button will illuminate the Dial Lock LED and the VFO is locked at the displayed frequency, thus inactivating the operation of the tuning knob. To change the frequency, the Dial Lock must first be disengaged by pushing the dial lock button again. The Dial Lock display LED will go OFF, and the frequency will change with the direction of the tuning knob rotation. Turning OFF the power will release the Dial Lock and clear the displayed frequency.

5-1-3 TUNING SPEED BUTTON

Pushing the Tuning Speed Button will illuminate the dial fast LED and the small vernier marks on the Tuning Knob is changed to correspond to 10KHz steps. At the same time, the 1KHz and 100Hz digits are cleared on the display to show "0.0". When the Tuning Speed Button is depressed again, the dial fast display LED goes OFF and the small vernier marks will again correspond to 100Hz steps. This button will allow you to quickly QSY over a great frequency range.

5-1-4 VFO SELECT SWITCH

The LSI in the IC-701 contains two "VFO's" for both receiving and transmitting. The VFO's are labeled "A", VFO and "B" VFO and are selectable with the VFO Select Switch. This dual VFO system gives the IC-701 many very convenient features. Please read this section very carefully and perform the operation several times until you are comfortable with the system.

1. "A" VFO is for both transmitting and receiving and is selected by placing the VFO Select switch in the "A" position. The transmit and receive frequency will be controlled by "A" VFO, displayed on the LED frequency readout, and stored in "A" memory.
2. "B" VFO is for both transmitting and receiving and is selected by placing the VFO Select switch in the "B" position. The transmit and receive frequency will be controlled by "B" VFO, displayed on the LED frequency readout, and stored in "B" memory.
3. RA-TB: The receive frequency is controlled by "A" VFO and the transmit frequency is controlled by "B" VFO. Both frequencies are stored in their respective memories, and displayed during their respective operation times ("A" frequency during receive, and "B" frequency during transmit).
4. RB-TA: The receive frequency is controlled by "B" VFO and the transmit frequency is controlled by "A" VFO, the reverse of the above. Again, both frequencies are stored in their respective memories, and displayed during their respective operation times (this time "B" frequency during receive and "A" frequency during transmit).

When the IC-701 is first turned ON, the frequency displayed on the LED readout is the lowest frequency limit of the band selected by the band switch.

FOR EXAMPLE:

If the band switch is set to the 7MHz band, when the set is turned ON, 7.000.0 will be displayed on the LED readout. This will occur whether the VFO switch is in either the "A" or "B" or combination VFO position. Rotating the tuning knob clockwise will increase the frequency in 100Hz steps. Rotating the tuning knob counter-clockwise will decrease the frequency in 100Hz steps. Note: Rotating the knob counter-clockwise immediately after applying power will not lower the frequency as it is already at the lowest limit.

Switching from one VFO to the other VFO does not clear the first VFO. The frequency is retained in the VFO's memory.

EXAMPLE

If 14.125.0MHz is set with "A" VFO, and the VFO switch is set to "B" VFO, the frequency LED readout will show "B" VFO's frequency, but 14.125.0MHz is still stored in "A" VFO's memory. Returning the VFO switch to "A" VFO position, and 14.125.0 will be displayed on the readout. Accordingly, if the switch is placed in the "B" VFO position, the frequency that was set with the "B" VFO will appear. This allows you to set a certain frequency with one VFO, work up and down the band with the other VFO, and periodically check the set frequency simply by switching between "A" and "B" VFO. It also allows you to search for a clear frequency with one VFO, while keeping your operating frequency on the other VFO. When you

have found a clear frequency, switch back to your operating frequency, inform the station you are in contact with of the new frequency, and switch back. It's that simple!

The IC-701 will perform this operation even on two different bands. A frequency is set in either "A" or "B" VFO on one band, the VFO switch is set to the other VFO, and the band switch set to any other band you wish to work. When you want to go back to the set frequency, first return to the original band and then to the original VFO and you're back on the set frequency.

FOR EXAMPLE

Set "A" VFO to 14.255.0MHz. Set the VFO switch to "B" VFO. Switch to the 21MHz band. The frequency readout will read 21.000.0MHz. Operate anywhere you wish with "B" VFO. When you want to return to 14.255.0MHz, first switch back to the 14MHz band. Then switch to "A" VFO and the frequency readout will show 14.255.0MHz.

When recalling a memorized frequency be sure to turn OFF the dial fast circuit. If the circuit is not turned OFF, the memorized frequency's 1KHz and 100Hz digits will both revert to "0".

RA-TB/RB-TA

These positions will allow you to operate split transmit/receive frequencies on the same band.

EXAMPLE:

Set "A" VFO to 14.350.0MHz and "B" VFO to 14.255.0MHz. Then set the VFO select switch to the RA-TB position. 14.350MHz will shown on the readout during receive ("A" VFO) and 14.255.0MHz during transmit ("B" VFO). You are now receiving on 14.350.0MHz and transmitting on 14.255.0MHz.

Setting the VFO select switch to RB-TA reverses the above. Changing bands in these modes resets both "A" and "B" VFO to the lowest band limit.

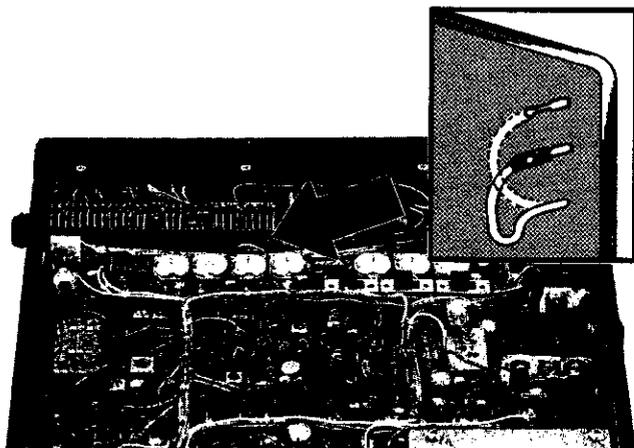
5 - 1 - 5 RIT (RECEIVER INCREMENTAL TUNING)

By using the RIT circuit, you can shift the receive frequency ± 800 Hz either side of the transmit frequency without moving the transmit frequency itself. Therefore, when you get a call slightly off frequency, or when the other station's frequency has shifted, you can tune in the frequency without disturbing the transmitting frequency. By pushing the RIT switch down once (it's a spring loaded switch) the RIT circuit is turned ON and the RIT LED is lit. The receive frequency is shifted with the RIT knob.

When the RIT knob is in the "0" position, the transmitting and receiving frequencies are the same. Rotating the control to the (+) side raises the receiving frequency, and rotating to the (-) side lowers the frequency. To turn OFF the RIT control, again push on the RIT switch and the RIT LED will go OFF. When the RIT circuit is OFF, the transmit and receive frequencies are the same regardless of

the setting of the RIT control knob.

When the RIT circuit is ON, if the tuning knob is moved by one increment, the RIT circuit is automatically pulsed OFF. However, this function can be eliminated by changing the connecting pins in the driver unit as shown below, in case you would like the RIT to remain operational when you turn the tuning knob. Or, by continuously depressing the RIT switch with your finger while turning the tuning knob, the RIT circuit will remain operational while tuning.



Other things to remember are: The RIT circuit is operational when the frequency is locked with the dial lock button. The frequency shifted by turning the RIT control is not indicated on the frequency display. When the transmitting and receiving frequencies differ by more than 800Hz, use "A" and "B" VFOs as outlined previously.

5 - 2 SSB OPERATION

5 - 2 - 1 RECEIVING:

After connecting an antenna, microphone, etc., set knobs and switches as follows.

POWER SWITCH	OFF
T/R SWITCH	RECEIVE (UP)
VOX SWITCH	OFF (UP)
ATT SWITCH	OFF (UP)
AGC SWITCH	SLOW (UP)
NOISE BLANKER (NB) SWITCH	OFF (UP)
MODE SELECT SWITCH	USB or LSB
VFO SWITCH	"A" or "B"
BAND SELECT SWITCH	Desired Band
AF GAIN CONTROL	Completely Counter-clockwise
RF GAIN CONTROL	Completely Clockwise
P.B. TUNE CONTROL	Center (12 o'clock) position
RIT CONTROL	Center (12 o'clock) position

The other knobs are unrelated to receiving and need not be set at this time.

In SSB operation there is both an USB (upper side band) and an LSB (lower side band). LSB is usually used on the 1.8, 3.5 and 7MHz bands, while USB is usually used on the

14, 21 and 28MHz bands. Select the mode according to the band in which you wish to operate.

Now turn ON the power switch. The meter lamp will be illuminated and the lowest frequency of the selected band will be shown on the LED readout. Slowly turn the AF gain control clockwise to a comfortable level. Rotate the tuning knob until a signal is received. The multi-function meter needle will move according to the signal strength, so tune for the highest possible meter reading and the clearest audio. If you cannot get a clear signal, you may be receiving in the opposite sideband. If so, change the mode switch to the proper sideband.

The VFOs in the IC-701 change in 100Hz steps. Therefore the maximum possible difference in your receiving frequency and the received signal frequency is 50Hz. However, for normal voice signal reception, a difference of 50Hz is negligible and indeed may be hard to distinguish. Since the frequency is changed in 100Hz steps with the IC-701, as opposed to continuous variable frequency systems, tuning is simplified and more accurate.

5 - 2 - 2 NB (NOISE BLANKER)

Turn the noise blanker ON (down) when there is pulse type noise, such as ignition noise from automobile motors. Noise will be suppressed and even weak signals will be received comfortably. The noise blanker may not work as effectively when there is a strong signal on a nearby frequency or when there is continuous (not pulse type) noise.

5 - 2 - 3 AGC (AUTOMATIC GAIN CONTROL)

The IC-701 has a "hang" AGC system which holds the peak voltage of rectified AF signals from the detector circuit for a certain period. Therefore, during the pauses in normal speech of the received signal, uncomfortable noises will not be heard. The multi-function meter indicates the peak value for a certain period, facilitating reading of the meter "S" function.

For normal SSB reception, set the AGC switch in the UP (slow) position. Set the AGC switch in the FAST (down) position, when tuning or receiving signals with short interval fading. When in the FAST position, the hang AGC circuit is turned OFF and the time constant is shortened. Use the FAST position for RTTY reception.

5 - 2 - 4 ATT (ATTENUATOR) SWITCH

Turn the ATT switch ON (down) when strong nearby signals disturb signal reception or make "S" reading difficult. In the ON position a 10dB attenuator is inserted into the receiving antenna circuit, reducing interfering signals and giving more stable reception. In normal operation the ATT switch is left in the OFF (up) position.

5 - 2 - 5 P. B. (PASS BAND) TUNING

Pass Band Tuning is a system in the receive mode to narrow the band width (selectivity) of the frequencies that will pass through the crystal filter electrically from either the upper or lower side continuously by up to 1.5KHz. This is very

access cover as follows:

VOX GAIN CONTROL
FULLY COUNTER-CLOCKWISE
ANTI VOX CONTROL
FULLY COUNTER-CLOCKWISE
VOX DELAY CONTROL
FULLY CLOCKWISE

Turn the VOX switch on the front panel to the ON (down) position. Leaving the T/R switch in the receive position and without pushing the PTT switch, turn the VOX gain control clockwise while speaking into the microphone. At a certain point, the T/R switching circuit will be activated by your voice. This is the proper position for the VOX Gain control. Set the VOX Gain control at a level which provides for T/R switching at your normal voice level. Transmit-release time (the time delay before the set automatically returns to receive when you stop talking) is controlled by the VOX delay control. Turning the control counter-clockwise makes the time shorter. Set it at a position which is comfortable and which allows for short pauses in normal speech.

Adjust the ANTI VOX control so that the VOX circuit is not activated by sounds from the speaker by turning the control clockwise while receiving a signal.

5-3 CW OPERATION

5-3-1 RECEIVING

For CW reception, set the MODE Select Switch to the CW or CW-N position. Other switches and knobs are set the same as for SSB reception. The Pass Band Tune circuit is preset internally for CW reception, therefore, when the MODE Select Switch is set at the CW or CW-N (CW NARROW) position, the Pass Band tuning control on the front panel is inactivated.

In addition to a crystal band pass filter, an audio active filter is built-into this unit. When the MODE Select Switch is set at the CW-N position, this audio filter is activated and the total selectivity of CW reception is improved. Also, with this filter, internal noise is reduced for comfortable CW reception and an improved signal to noise (S/N) ratio.

5-3-2 TRANSMITTING

Insert the keyer plug into the KEY Jack on the rear panel of the unit, and set knobs and switches as follows:

COMP CONTROL OFF (Fully counter-clockwise)
METER SELECT SWITCH P₀ POSITION
Other knobs and switches are set the same as for CW reception.

By setting the T/R switch to Transmit, the transmit LED is lit and shows that you are ready for CW transmission. When you key the keyer, the meter needle moves and your

CW signal is transmitted. To decrease the transmitting power, turn the COMP Control clockwise to ON and rotate to adjust while watching the meter needle on the P₀ scale for the desired output power.

5-3-3 CW MONITOR

When keying, the side tone oscillator circuit is activated and a 900Hz tone is heard from the speaker. The loudness of this tone is adjusted by rotating the CW MONI Control, located under the access cover. Rotating the CW MONI Control clockwise makes the tone louder. (In the receive mode, this side tone is audible and can therefore be used for Morse Code Practice, adjustment of the keyer, etc.) To receive again, place the T/R switch in the receive position.

5-3-4 SEMI BREAK-IN OPERATION

The IC-701 has Semi Break-In CW capability when using the VOX function. When keying, the unit is automatically set in the transmit mode. After keying, it is returned to the receive mode, also automatically, after a given transmit-release delay time constant. Leave the T/R switch in the Receive position, and set the VOX switch in the ON (down) position.

The transmit release delay time constant is set by adjusting the CW DELAY Control located under the access cover. Turning the CW DELAY Control clockwise will make the transmit release time longer. Set for your keying speed.

5-4 RTTY OPERATION

For RTTY operation, a teletypewriter and a demodulator (terminal unit) which is operational with audio input are necessary. Any demodulator with 2125/2295Hz (narrow, 170Hz shift) or 2125/2975Hz (wide, 850Hz shift) filters can be used with the IC-701. You can select NARROW or WIDE shifts by setting the RTTY SHIFT Switch, located under the Access Cover, to N or W. Fine adjustment of the MARK frequency can be accomplished by rotating the NARROW or WIDE variable resistors, located behind the RTTY SHIFT Switch.

5-4-1 RECEIVING

Audio signals for the demodulator can be supplied from Pin 4 of the ACC socket on the rear panel, or from the PHONES jack on the front panel. The level of the audio signals from Pin 4 of the ACC socket does not vary by turning the AF GAIN Control.

Set the MODE Select Switch to RTTY. The frequencies for NARROW shift are 2295Hz for MARK and 2125Hz for SPACE. For WIDE SHIFT, they are 2975Hz for MARK and 2125Hz for SPACE.

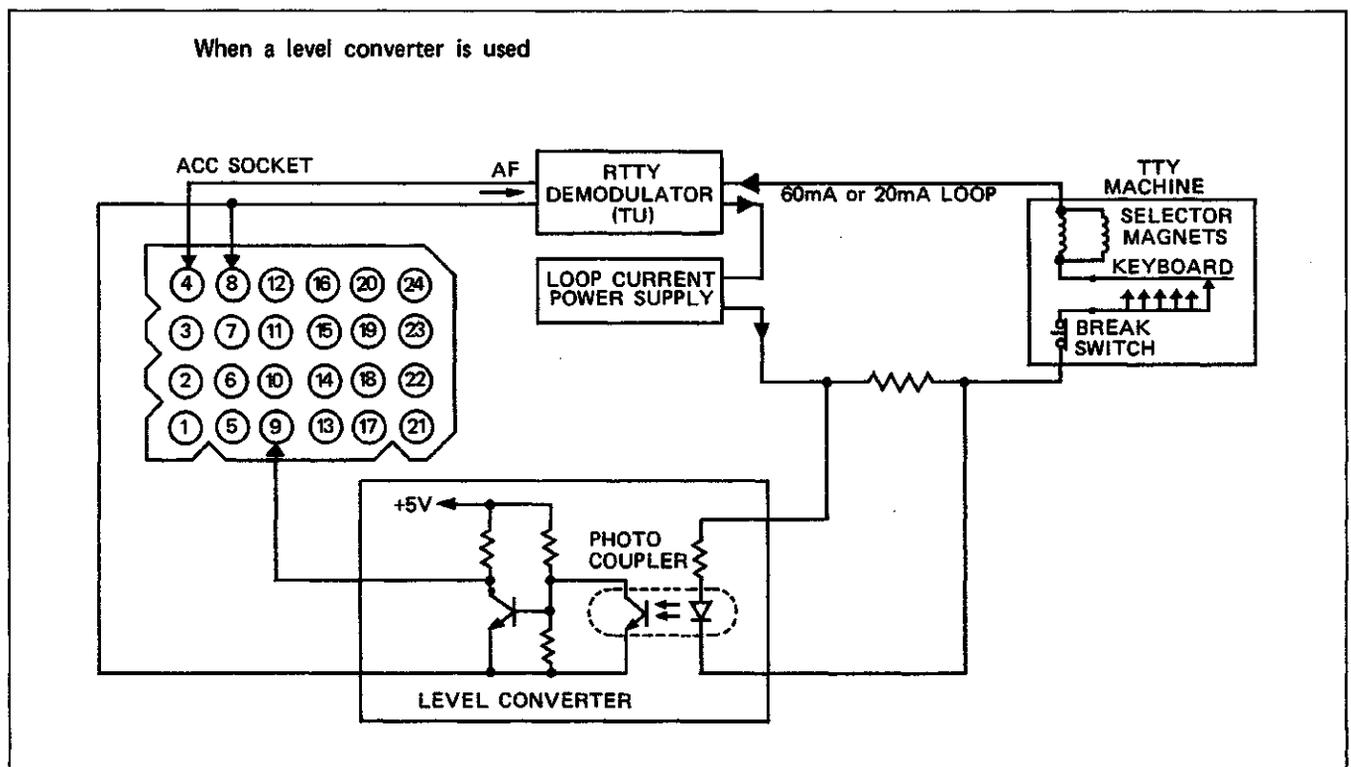
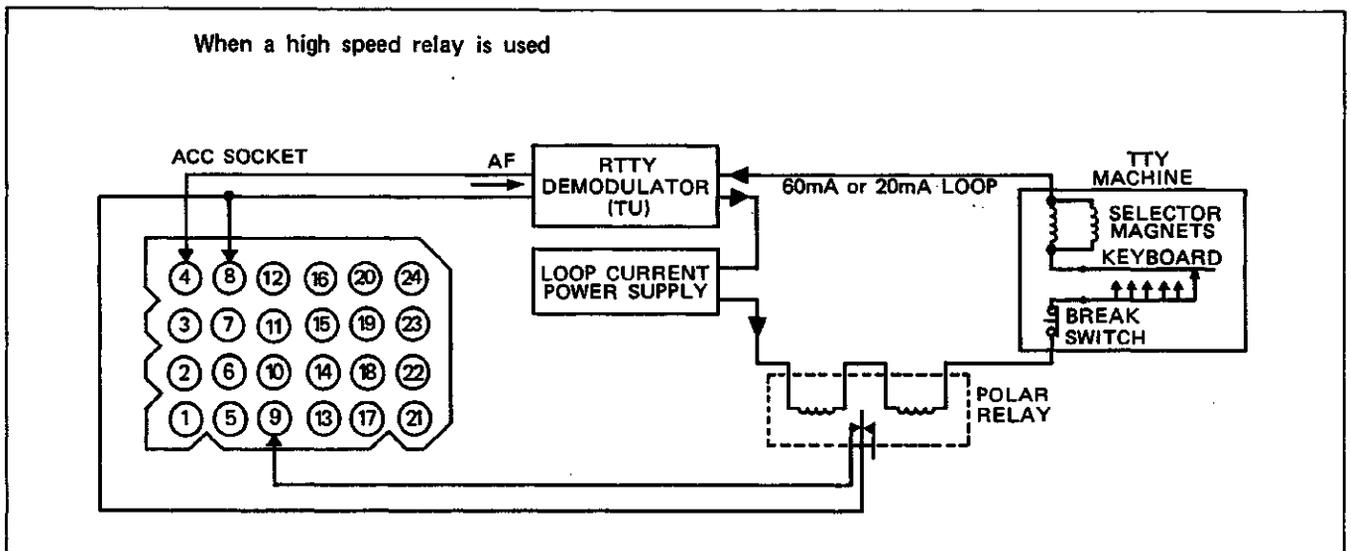
If you have an AFSK generator, set the MODE Select Switch on LSB for RTTY operation. Then the frequencies for NARROW SHIFT are 2295Hz for SPACE and 2125Hz

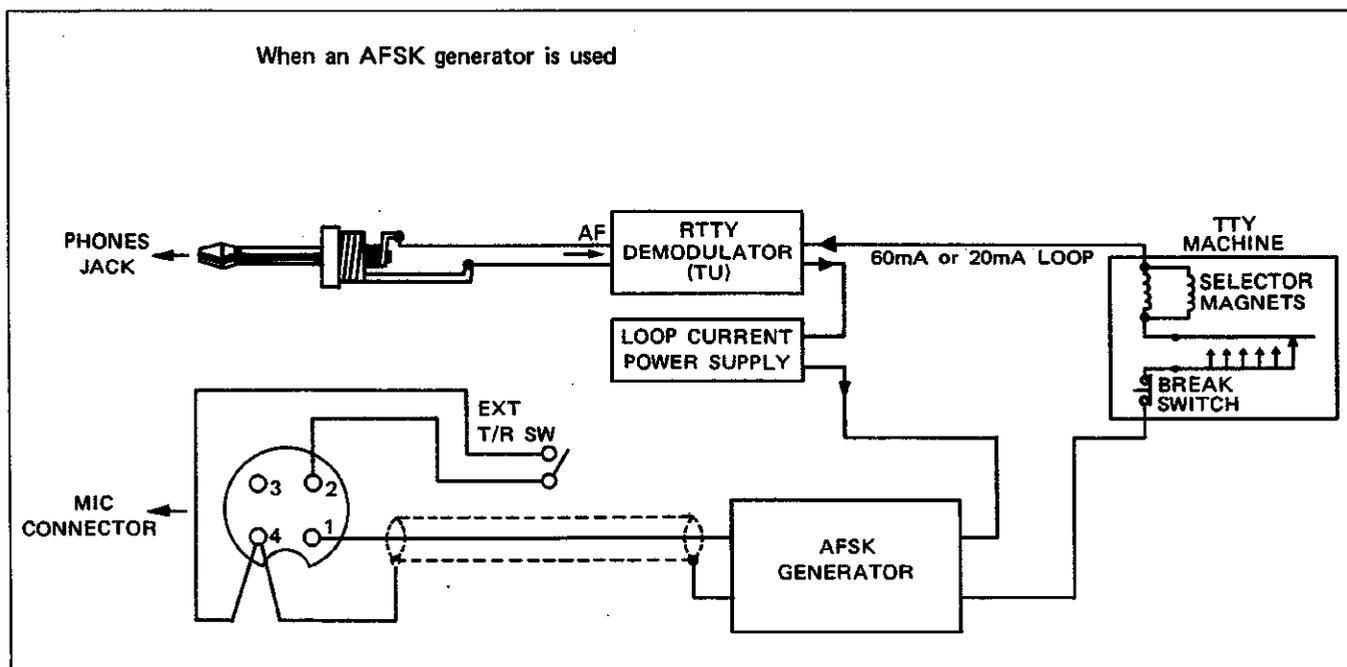
for MARK. For WIDE SHIFT, they are 2975Hz for SPACE and 2125Hz for MARK. Also turn the P. B. Tune control fully clockwise for clear reception.

5-4-2 TRANSMITTING

For Keying of the FSK (Frequency Shift Keying) circuit, insert a high speed relay's coil into the loop current circuit of the teletypewriter and connect the relay contacts to Pins 8 and 9 of the ACC socket on the rear panel. The relay contacts make during the SPACE and break during the MARK.

When a level converter for TTL level signals is used, connect the output of the converter to Pins 8 (ground) and 9 of the ACC socket. Apply HIGH level (5V) signals for the MARK and LOW (0V) signals for the SPACE. When using an AFSK generator that has 2125Hz for MARK and 2295Hz (2975Hz in wide shift) for SPACE, connect the output signals for the AFSK generator to the MIC connector on the front panel and turn the MODE Select switch to the LSB mode. In this case, you can use the VOX operation available in the SSB mode, so receive to transmit change over is very easy.





5-5 OTHER OPERATIONS

5-5-1 VSWR READING

The IC-701 has a built-in VSWR meter for checking antenna matching in order to avoid problems caused by VSWR. Set the METER Select switch to the Po position, the T/R switch to Transmit, and Mode switch to the RTTY position. Adjust the SWR SET control so that the meter needle points to "SET" on the meter scale. Set the SWR switch, located under the access cover, to the SWR position. (Before your unit is shipped, the SWR switch is fixed in place in the SET position with a plastic screw. Remove this screw before trying to make VSWR readings). With the switch in the SWR position, SWR reading can be seen on the meter. Although this unit is built to handle VSWR of up to 2:1, it is recommended that the antenna(s) be adjusted for the lowest possible VSWR. After taking the reading, return the switch to the SET position. **ALSO BE SURE THAT THE ANTENNA IMPEDANCE IS 52 OHMS OR THERE MAY NOT BE ANY OUTPUT. OTHERWISE THERE WILL BE DAMAGE TO THE TRANSCEIVER.**

The final transistors used in the IC-701 are of good design and are protected to a reasonable extent by circuits incorporated in the set. These devices can be expected to have an indefinite lifetime since there are no cathodes to burn out. Under some conditions, however, they can be abused beyond tolerance and may have to be replaced.

When in doubt about antenna systems, use the lowest power setting possible to achieve meaningful readings. Use a good tuner or transmatch when necessary. Always use caution and exercise judgement when testing RF power generators.

5-5-2 AUTO DIMMER

For optimum visual comfort, the meter lamp and frequency

display LEDs are automatically dimmed and brightened depending on the brightness or darkness of the surrounding light by use of a built-in Dimmer Photosensor, located on the right side of the frequency display on the front panel. For the AUTO DIMMER to function properly, take care that the dimmer photosensor is not covered by your hand or other object.

The threshold for the functioning of the AUTO DIMMER can be adjusted by turning the DIM GAIN Control, located under the access cover. Rotating the control counter-clockwise will lower the operation threshold of the circuit. The circuit can be turned OFF by the DIM switch located also under the access cover.

5-5-3 WWV RECEPTION

To receive WWV (or other standard frequency station), set the BAND SELECT switch to the 14MHz band and the MODE switch to the USB or LSB position. Tune to 15.000.0MHz on the frequency display. The WWV signal can be used for alignment of a frequency counter, marker oscillator, or the frequency display.

5-5-4 SIMPLE FREQUENCY ALIGNMENT

A very accurate frequency counter is necessary to align the frequency of the IC-701. However, the frequency can be aligned simply by receiving the WWV signal.

1. Set the frequency display to 15.000.0MHz and be sure that you are receiving the WWV signal.
2. Set the Mode Select switch to CW. A 900Hz beat can be heard.
3. Short the KEY Jack on the rear panel so that the CW sidetone also becomes audible.
4. Rotate the FREQUENCY SET Control, located under the access cover, so that the two tones are of the same pitch. If the tones are difficult to adjust because of a

difference in their strengths, adjust the CW sidetone level with the CW MONI control, also located under the access cover, until the strengths are the same.

WHEN ALIGNING THE FREQUENCY, DO NOT PUT THE UNIT IN THE TRANSMIT MODE.

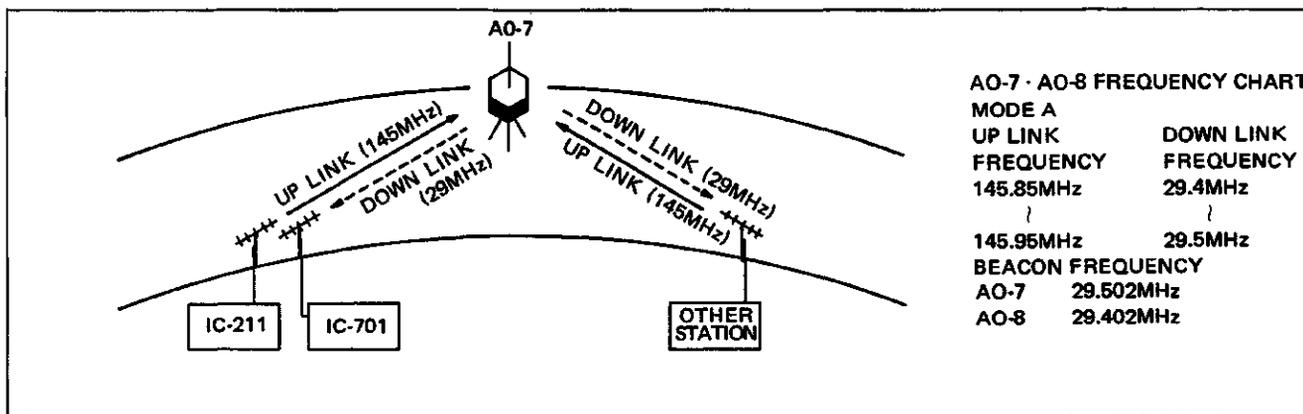
Be sure the T/R switch is in the Receive position, the VOX switch is OFF, and that you do not touch the PTT switch on the microphone.

5-5-5 SATELLITE COMMUNICATION

At present there are two active satellites for amateur radio use, though more are being planned. AMSAT OSCAR 7

was launched in November, 1974, and AMSAT OSCAR 8 was launched in March, 1978. Satellite communications can be made by using an IC-211/211E for transmitting and an IC-701 for receiving. For operation of the IC-211/211E and the IC-701 and for frequency information, refer to the chart below.

Since communication satellites are moving at very high speeds, the down-link signals can be heard on a higher frequency than the real frequency as a satellite approaches, and on a lower frequency as it travels further away. Further information on communicating through satellites can be found in various amateur radio books and periodicals.



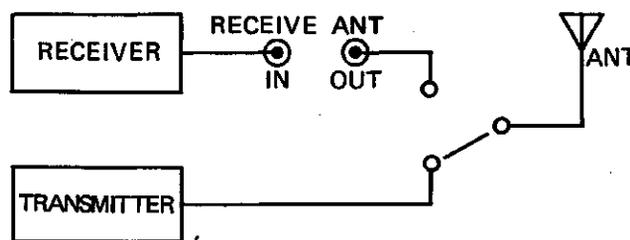
5-5-6 RECEIVE ANTENNA TERMINALS

The RECEIVE ANT IN jack is connected to the input terminal of the receiving section, and the RECEIVE ANT OUT jack is connected to the antenna connector through the internal T/R antenna switching circuit. These two jacks are normally jumpered with a cable, but can be used with:

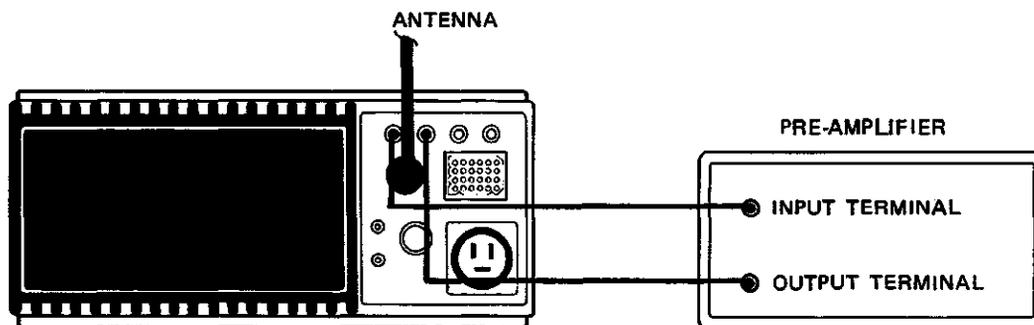
1. A receiving pre-amplifier
2. A separate receiver
3. Separate receiver and transmitter antennas.

If you wish to use a receiver pre-amplifier, connect it between the receiver input and antenna output terminals.

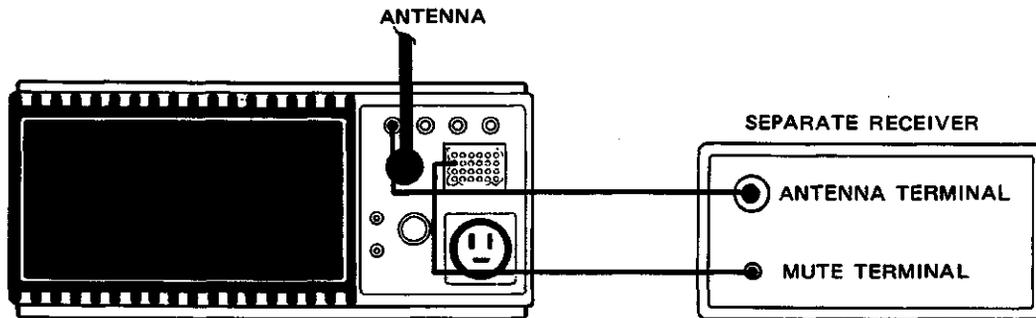
If a separate receiver is used, connect it to the receiver antenna output terminal. For a separate receive antenna, connect it to the receiver input terminal.



CONNECTION OF A PRE-AMPLIFIER

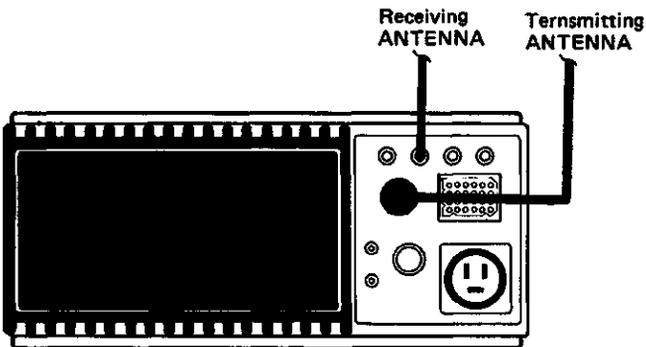


CONNECTION OF A SEPARATE RECEIVER



Connect a 12V small relay between Pins 2 and 3 of ACC socket, and control the separate receiver by contacts of the relay.

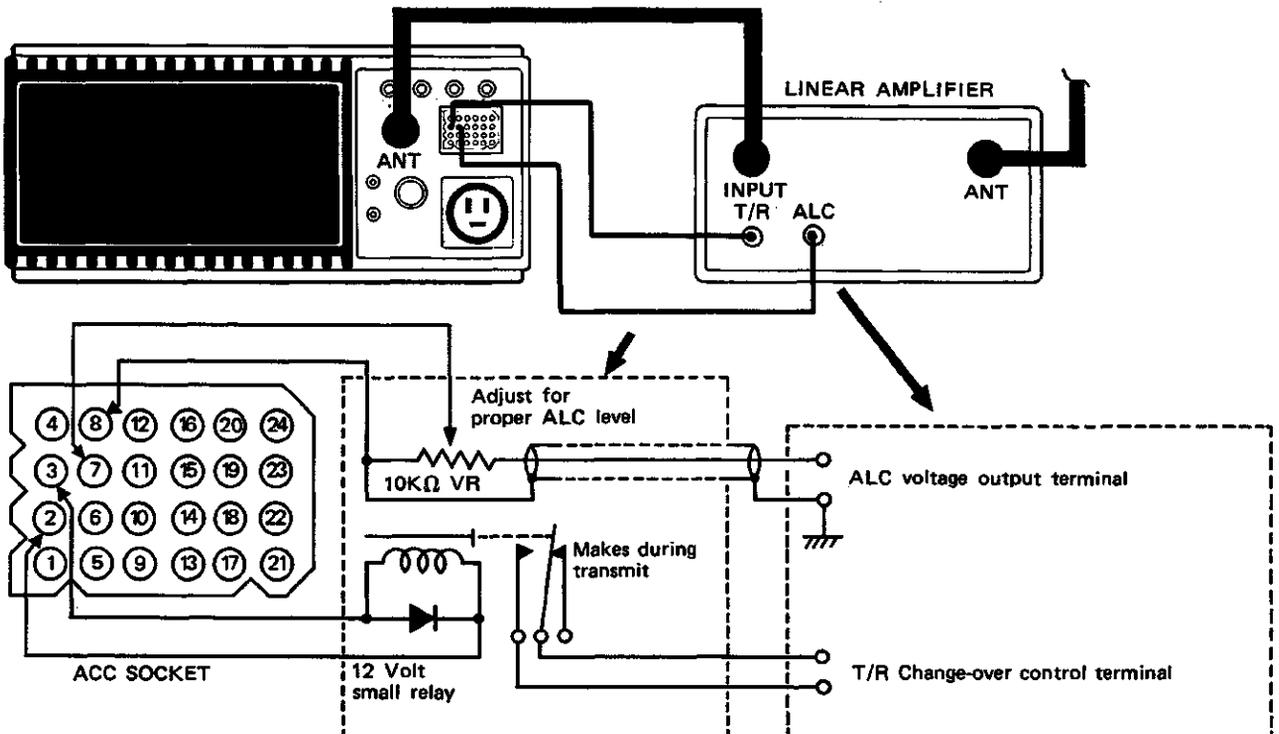
CONNECTION OF SEPARATE ANTENNAS



5-5-7 CONNECTION OF A LINEAR AMPLIFIER

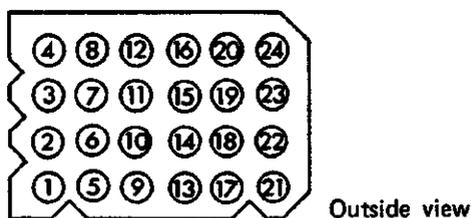
Connect the output connector (ANT) to the input of the linear amplifier. If the linear does not have an RF sense line, it will be necessary to use the circuit shown below to key the Amplifier. If ALC is available from the amplifier, it should be of negative polarity and about 5 volts peak for complete shut down. The ALC voltage should be connected to the ALC connector of the IC-701 and the ALC connector of the amplifier.

CONNECTION OF A LINEAR AMPLIFIER



5 - 5 - 8 CONNECTION OF ACCESSORY SOCKET

Various adapters can be used through terminals in this SOCKET for frequency control input, modulation output, receiver output, T/R change-over control and so on. The table below shows terminal connections of this connector. Care should be taken not to apply voltages other than -0.5 V to +5V to terminals between No. 14 and No. 24 as they are connected with the C-MOS IC.



ACC SOCKET CONNECTIONS

PIN No.	FUNCTION
1.	NC (no connection)
2.	13.8 Volts DC in conjunction with the power switch operation.
3.	Connected to Push-to-talk, T/R change-over switch. When grounded, the set operates in the transmit mode.
4.	Output from the receiver detector stage. Fixed output regardless of AF output or AF gain.

PIN No.	FUNCTION
5.	Output from Transmitter MIC amplifier stage.
6.	8 Volts DC available when transmitting. (relay can not be directly actuated.)
7.	Input for external ALC voltage.
8.	Ground
9.	Input for RTTY keying (MARK: HIGH level, SPACE: LOW level).
10.	8 Volts DC available when the 28MHz band is selected.
11.	Input for TRANSVERTER control. When 8 Volts DC is applied, set can operate with a transverter.
12.	Output reference voltage for band switching.
13.	INPUT/OUTPUT for external band switching.
14.	CKE Output HIGH level at band edge.
15.	LOCK Input to lock dial externally.
16.	UDC Input to control Up/Down counter externally.
17.	SCAN Input to scan frequency.
18.	CL Input to clear frequency.
19.	FCL Input to clear counter in specified digit and input for MSB data.
20.	K0 Input for frequency control data.
21.	K1 Input for frequency control data.
22.	K2 Input for frequency control data.
23.	K4 Input for frequency control data.
24.	K8 Input for frequency control data.

SECTION 6 THEORY

OUT LINE

The IC-701 employs a digital phase locked loop (PLL) circuit as the local oscillator for both transmit and receive. The output of the PLL circuit is approximately 9MHz above the receive frequency, thereby spurious is kept to a minimum.

The frequency is determined as follows. A pulse generated by the optical chopper circuit, located at the tuning knob, is digitalized by the up/down counter in the LSI (ICOM's custom developed Large Scale Integrated circuit) and used to control a programmable divider, also located inside the LSI chip. The programmable divider controls the PLL circuit which determines the output frequency of the VCO (Voltage Controlled Oscillator).

In the receiver section, signals from the antenna are mixed with the local oscillator output from the PLL circuit. After passing through the pass band tuning circuit, the signals are amplified then detected to become audio signals, amplified again and sent to the speaker.

The transmitter uses a carrier of 9.013MHz for USB and 9.010MHz for LSB. The carrier and the voice signal are sent to a balanced modulator where the DSB suppressed

carrier signal is generated. The unwanted sideband is removed by a crystal filter, and an SSB signal of 9.0115MHz is obtained. This SSB signal is mixed with the local oscillator output from the PLL circuit, which is the same as that of the receiver section, and then amplified, filtered, and sent to the antenna.

6 - 1 RECEIVER

6 - 1 - 1 RF AMPLIFIER AND MIXER CIRCUIT

A signal from the antenna is passed through the ALC and filter units and applied to the antenna switching circuit located on unit "B". During reception, Q17, the antenna switching transistor, is ON, thus forward biasing D25 and D26. The incoming signal passes through these diodes, through the external receive antenna jumper, and is fed to the attenuator circuit. If the attenuator switch on the front panel is in the ON position, the signal passes through the attenuator and is attenuated 10dB. The signal is then fed to the RF unit through J3. If the attenuator switch is in the OFF position, the signal is fed directly to J3. D22 through D33 are switching diodes for the input and output signals through their respective RF circuits depending on the band selected. Q15 through Q20 are dual gate MOS FET amplifiers. The incoming signal is fed to the first

gate, and AGC voltage is fed to the second gate to increase the dynamic range of the amplifier. The signal is then passed through a Schottky Diode double balanced mixer composed of D11 through D14 where it is mixed with the local oscillator signal from the VCO. From the mixer, the resultant 9.0115MHz signal passes through a monolithic filter, having a bandwidth of 10KHz, and to the first gate of amplifier, Q3. AGC voltage from pin 6 of J7 is fed to the second gate of Q3 as it is to the gates of Q15 through Q20.

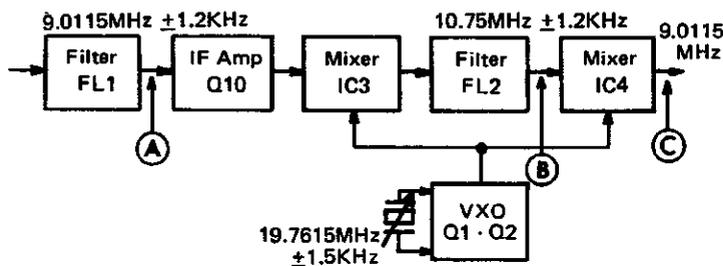
6-1-2 NOISE BLANKER CIRCUIT

The noise blanker circuit is composed of Q8 through Q12. Noise amplified by Q8-Q10 is detected by Q20 and Q21 and applied to the base of Q12. Since the collector of Q12 is connected to the bias circuit of D1 and D2, when Q12 conducts because of detected noise, it shuts off D1 and D2, cutting off the noise. Therefore only clean signals are fed through J1 to the IF circuit.

6-1-3 PASS BAND TUNING CIRCUIT

A VXO circuit composed of Q1, Q2, X1 and D4 in the "A" unit. The bias voltage of D4 is changed by turning the Pass Band Tune control, located on the front panel, which varies the VXO frequency (19.7615MHz) by ± 1.5 KHz. Signals from J1 in the RF unit are fed through J1 in the IF circuit to L5, then through FL1, a crystal filter. FL1's center frequency is 9.0115MHz with a band width of ± 1.2 KHz. From the filter, the signal is fed to Q10, where it is amplified and fed to the balanced mixer, IC3, where the 9.0115MHz signal is mixed with the 19.7615MHz VXO signal. From the balanced mixer, the resultant 10.750MHz signal is fed through FL2, a crystal filter having a center frequency of 10.750MHz and a band width of ± 1.2 KHz. This 10.750MHz signal is then again mixed with the VXO frequency, to result in the original 9.0115MHz signal to be fed to the IF amplifier circuit.

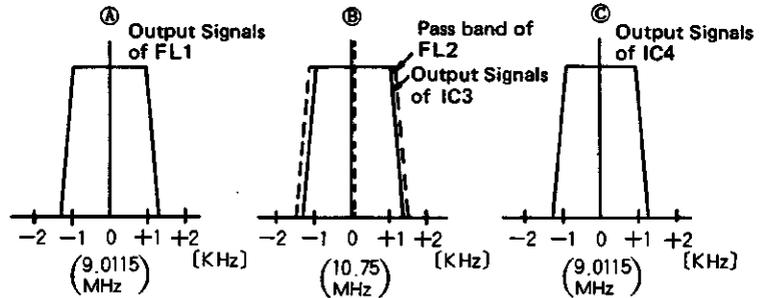
The diagrams will explain the operation of the Pass Band Tuning in more detail.



In the diagrams, "A" represents the 10.750MHz signals that are the result of mixing of the 9.0115MHz input signals with the 19.7615MHz VXO signal in IC3. These 10.750 MHz signals are represented by the solid line. The dotted line represents the pass band of the second crystal filter, FL2. With the Pass Band Tune control in the center position, signals between 9.0127MHz and 9.0103MHz

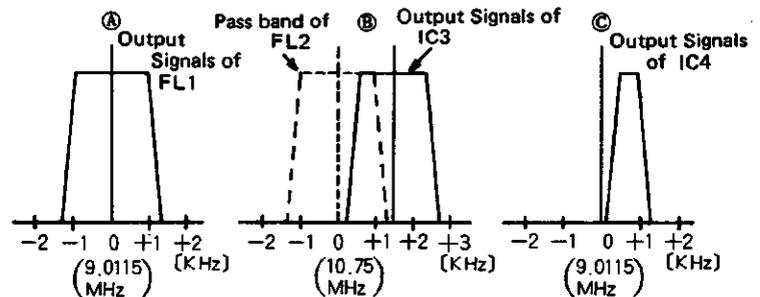
(9.0115 ± 1.2 KHz), when mixed with the VXO center frequency (19.7615MHz) result in signals of 10.7512MHz to 10.7488MHz ($10.750\text{MHz} \pm 1.2$ KHz), all signals will pass through the filter.

A : P.B.TUNE centered



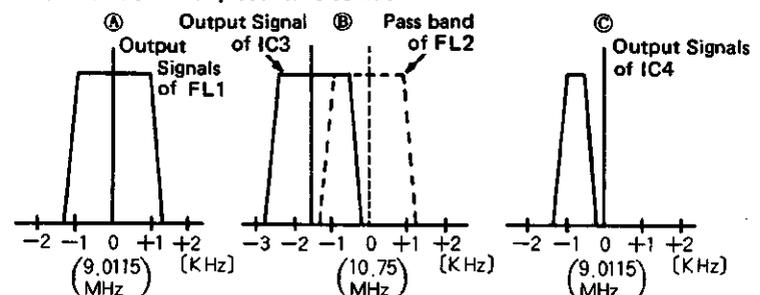
Turning the Pass Band Tune control fully clockwise, the VXO frequency will become 19.7630MHz ($19.7615 + 1.5$ KHz). When mixed with the incoming 9.0115MHz signals, the resultant frequencies become 10.7503MHz to 10.7527MHz. Since the upper frequency pass band limit of the filter is 10.7512MHz, not all the signals will be passed by the filter. Only the signals between 10.7503MHz and 10.7512MHz, the upper limit, will be passed. As you can see, the pass band has been narrowed to 900Hz to one side of the center frequency of the filter. Therefore, all signals outside the 10.7503MHz to 10.7512MHz range will not pass through the filter.

B : P.B.TUNE fully clockwise



By turning the pass band tuning control in the fully counter-clockwise position, the VXO frequency becomes 19.760MHz. The 10.750MHz signals from the mixer will be from 10.7497MHz to 10.7473MHz. Again with the 10.7512MHz to 10.7448MHz pass band width of the filter, not all signals will be passed. Only those between 10.7497 MHz to 10.7488MHz will be passed. Again, the pass band has been narrowed to 900Hz and offset to the opposite side of the center frequency of the crystal filter.

C : P.B.TUNE fully counter-clockwise



This action, and what this circuit really can do for you is better explained with an example.

If you are receiving a signal, whose frequency at the input of the mixer is 9.0115MHz and another signal whose frequency is 9.0125MHz is present, after being mixed with the 19.7615MHz VXO frequency (center frequency) the resultant frequencies are 10.750MHz, the signal you are listening to, and 10.7490MHz, the other signal. Both are within the pass band of the filter, and therefore, both will be heard. However, by turning the pass band tune control until the VXO frequency reaches 19.7605MHz, the 9.0125 MHz interfering signal when mixed with the VXO frequency will have a resultant frequency of 10.7480MHz, which is out of the pass band of the filter. The signal you are listening to, however, when mixed with the VXO frequency becomes 10.749MHz. This is within the pass band of the filter, and when mixed again with the VXO frequency in the next stage, becomes it's original 9.0115MHz signal and is sent to the IF amplifier. You have pushed the interfering signal out of the pass band of the filter, and therefore eliminated it and the interference it caused.

In the CW mode (CW or CW-N position of the MODE switch), a fixed bias voltage is applied to D4 of the VXO circuit, so the VXO frequency shifts to 19.7634MHz (+1.9KHz). The bandwidth then becomes $\pm 250\text{Hz}$ for clear CW reception.

The signal which has passed through the pass band tuning circuits are amplified by Q11 and Q12, dual gate MOS FET's. AGC voltage is applied to the second gate of both. The amplified IF signal is then sent to the detector circuit.

6-1-4 DETECTOR CIRCUIT

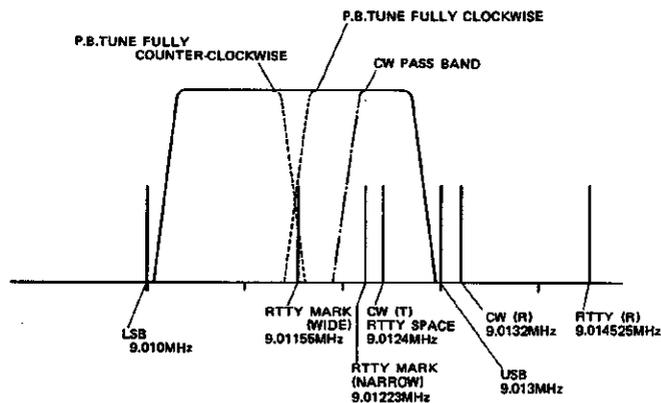
The detector circuit is composed of IC2, a differential amplifier that consists of a pair of similar input circuits and a constant-current source. The IF signals are fed to the both input circuits of IC2 180 degrees out of phase, and the BFO signal is fed to the constant-current source. The detected audio signal is sent to the CW AF filter circuit and the AF amplifier circuits through the MODE switch.

6-1-5 BFO CIRCUIT

The BFO circuit is composed of Q18 to Q20 and crystals X1 and X2, all located on the "B" unit. Q19 is the crystal oscillator, O18 is the buffer, and Q20 is the FSK (Frequency Shift Keying) keyer for RTTY.

The crystals are switched by D12 and D13, and the oscillation frequencies are shifted by D14 to D18, according to the mode used. Crystal X1 (9.013MHz) is used for USB transmitting and receiving, CW receiving, and RTTY receiving. Crystal X2 (9.010MHz) is used for LSB transmitting and receiving, CW transmitting, and RTTY transmitting. In the RTTY transmitting mode, the bias voltage to D19 (a varicap) is changed according to the MARK and SPACE RTTY signals, by Q20. S3 (RTTY shift switch) changes the shift width for Narrow (170Hz) shift, and Wide (850Hz) shift.

BFO FREQUENCY CHART



	MODE	FREQUENCY
RECEIVE	LSB	9.010 MHz
	USB	9.013 MHz
	CW	9.0132 MHz
	RTTY	9.014525MHz
TRANSMIT	LSB	9.010 MHz
	USB	9.013 MHz
	CW	9.0124 MHz
	RTTY	9.0124 MHz (SPACE)
		9.01223 MHz (170MARK)
		9.01155 MHz (850MARK)

6-1-6 CW AF FILTER CIRCUIT

After detection, the audio signal is applied to IC6. Next, the signal is fed to a CW AF filter composed of high and low pass filters. The filters have a pass band of about 800Hz $\pm 100\text{Hz}$. If the mode switch on the front panel is turned to CW-N, the signal passes through this bend pass filter and goes to the audio amplifying section in the next stage. During other modes, the detected output is directly applied to the audio amplifier circuit.

6-1-7 AUDIO AMPLIFIER CIRCUIT

The audio amplifier circuit is composed of Q44 to Q46. The amplified signal passes through an AF gain control on the front panel, and then goes into Q27, an audio low pass filter, where signal components above 3KHz are removed. After that, the signal is amplified in the audio power amplifier circuit, IC1, which then drives the speaker.

6-1-8 AGC CIRCUIT

The AGC circuit is composed of Q28 to Q36 on the "A" unit. After detection, for CW-N after passing through the CW AF filter, the AF signal is amplified by Q28. It then undergoes peak hold detection by detector Q32. The detected DC voltage charges C101 and C102. The DC voltage is then amplified further by Q33 and Q34. Thus, the AGC voltage for RF and IF amplifier circuits is

obtained from the collector of Q34.

When the AGC switch is in the SLOW position, the output signal of Q28 is fed to another peak hold detector, Q29. The detected DC voltage charges C99 and switches Q30 ON and Q31 OFF. When the input signal is removed, the charged voltage of C99 is discharged through R159 and Q31 is switched ON, and the charged voltage in C101 and C102 is discharged through R164 and Q31. Thus the AGC voltage is held during the time constant of C99 and R159.

When the AGC switch is in the FAST position, -9 Volts is applied to the base of Q30, turning it OFF, and turning Q31 ON. Therefore the AGC voltage does not hold, and the AGC release time is determined by C101 and R192. The AGC attack time varies with the mode of operation, so when the mode is changed, so are the determining fixed resistors. C101 and C102 are the capacitors that are in the circuit for all modes, and determine the time constant along with the following resistors: For USB and LSB, R166 and R192 (in parallel), both of which are located on the "A" unit. For CW and RTTY, R192 and R48, which are located on the mode switch (in series); As the time constant circuit may be charged during transmit, Q35 is designed to be ON during transmit time, through D33, and shunts C101 and C102. When the power is turned ON, a pulse through R173 and C103 turns ON Q35 temporarily to discharge C101 and C102 in order to allow the AGC voltage to recover to a non-signal bias state.

Q36 superimposes the RF Gain control voltage on the AGC line by way of the emitter follower, forming a threshold-type RF gain control. As the RF AGC circuit operates as the ALC circuit during transmit, Q36 is biased through D31 to remove interference troubles.

6 - 2 TRANSMITTER

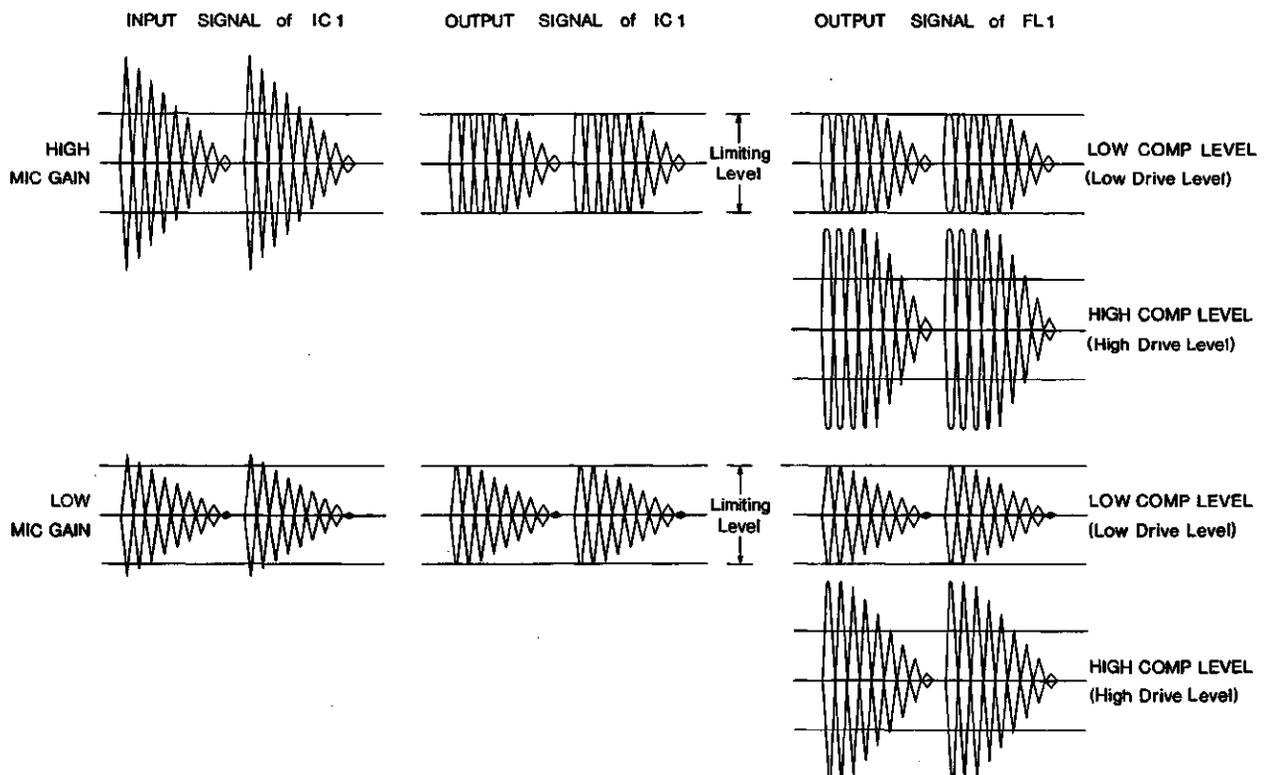
6 - 2 - 1 MICROPHONE AMPLIFIER CIRCUIT

Voice signals from the microphone are passed through the MIC Gain control on the front panel and applied to Q17, the ALC attenuator, located on "A" unit. The ALC voltage is applied to the gate of Q17, which changes the source-to-drain impedance. Accordingly, the ALC controls the voice signal passing through Q17. The signal is then fed to amplifiers Q15 and Q16 and then sent to the balanced modulator.

6 - 2 - 2 BALANCED MODULATOR CIRCUIT

IC5 is a balanced modulator where the BFO signal, fed through Pin 10, and the voice signal, fed through Pin 4, result in a double side band signal output from Pins 6 and 12. When the mode switch is in the CW, CW-N or RTTY mode, power for the MIC AMP circuit is turned OFF and therefore, the voice signal is not applied to IC5. Instead, a voltage from D23 is applied to Pin 4, IC5 loses balance, and the BFO signal becomes the output. This signal is used for the carrier in CW and RTTY operation. During RTTY operation, the frequency is shifted according to the keying in the BFO. During CW operation, a key is connected to the base of Q14. The key turns Q14 ON and OFF at the keying rate, therefore, turning OFF D22 and the carrier, ON and OFF at the keying rate.

When the speech processor circuit is not used, the output signal of IC5 (DSB signal) is fed to the crystal filter FL1 through D21, D22 and D34. FL1 cuts the unwanted side band of the DSB signal and the output is a Single Side Band signal. (CW and RTTY signals pass through FL1 as unchanged wave forms). From the filter, the signal is sent to the RF unit through P1.



When the speech processor circuit is used, the output signal of IC5 is fed to the speech processor circuit through D22 and D13.

6-1-3 SPEECH PROCESSOR CIRCUIT

The speech processor circuit uses the pass band tuning circuit. While transmitting with the processor ON, a DSB signal output in the balanced modulator is passed through D22 and D13, and applied to IC3, where it is mixed with the VXO output frequency of 19.7615MHz. It passes through the crystal filter FL2 and becomes an SSB signal of 10.750MHz. This signal is mixed with the VXO output at IC4 again to become an SSB signal of 9.0115MHz, amplified by Q11, and passed to IC1. Q11 as well as Q12 and D19 are in the compressor level detection circuit, and the output for the compressor level indicator is obtained from the source of Q11. IC1 is the limiting circuit, and it clips signals above a fixed level, as shown in the figure on page 23.

Q4 is the drive level control circuit. By turning the COMP (compressor) control, the bias voltage is changed and gain is controlled. Some splatter is included in the signal which has passed through the limiter circuit, but this is completely removed by the crystal filter FL1, and the final signal is a clean and powerful SSB signal.

In the CW and RTTY modes, the speech processor circuit works the same as in the SSB mode, and the output power is adjustable from 0 to about 100 Watts by turning the COMP control knob.

6-2-4 TRANSMIT MIXER CIRCUIT

The output from P1 of the "A" unit is passed to J1 on the RF unit, and after passing through the T/R switching diode, D4, it is applied to Q1, the attenuator of the drive level control for each band. Each band has its own drive level control in the RF amplifier circuit. In order to compensate for this, the voltage fixed at R92 to R97 is applied to the gate of Q1, and proper drive for each band is obtained.

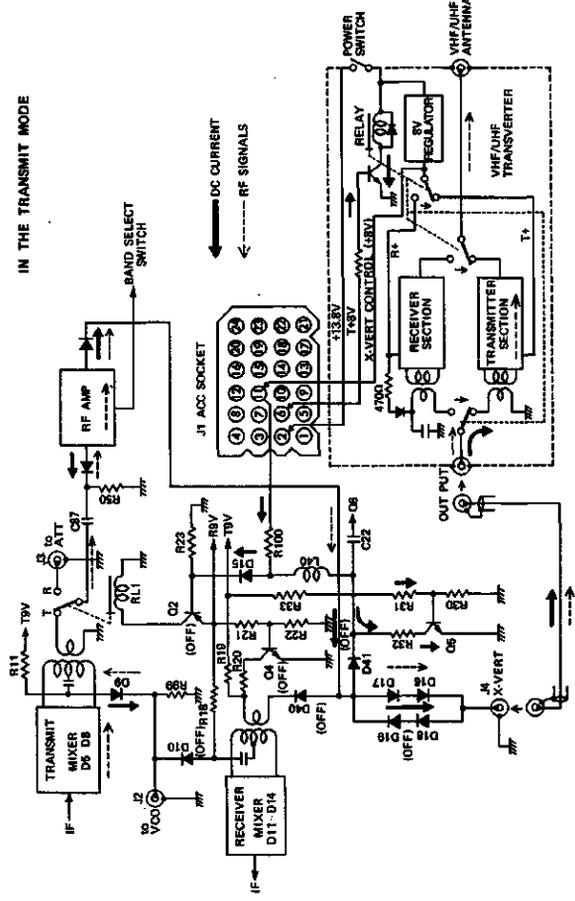
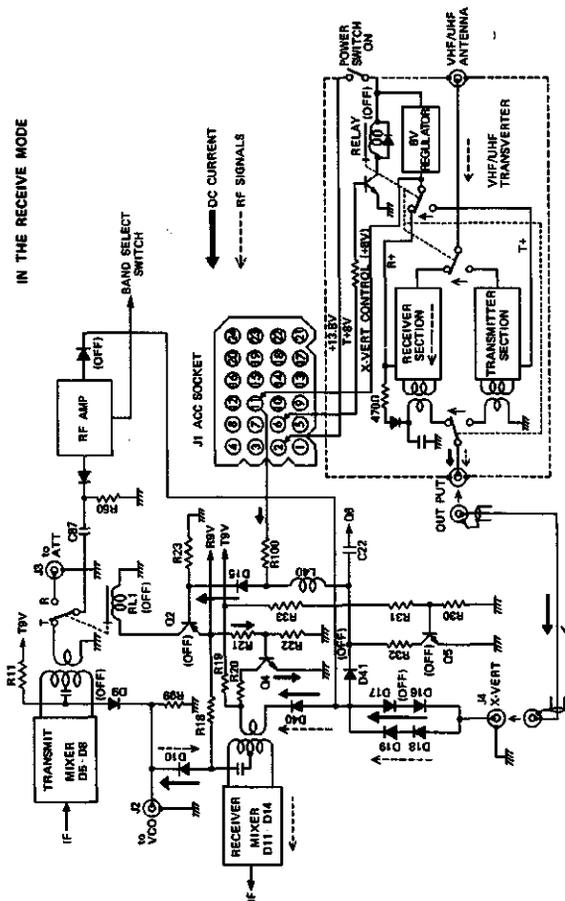
The output is mixed with the VCO output frequency in the Schottky diode double balanced mixer circuit composed of D5 to D8, and the output becomes the final transmitting frequency. The mixed output is fed to RF amplifier through the T/R relay RL1.

6-2-5 RF AMPLIFIER CIRCUIT

The mixed output is applied to the RF amplifier circuit (the same one as is used in the receiver section) composed of Q15 to Q20 which are set for each band. L20 to L25 and L26 to L31 are multiple coupled resonant coils which are staggered to obtain a flat wide band pass. The output is fed through T/R switch diode D41, and is amplified by Q9, the predrive circuit in the next stage, to a level of about 50 milliwatts. This is then fed to J6.

6-2-6 TRANSVERTER OUTPUT CIRCUIT

When a transverter is used for VHF or UHF operation, connect the input/output cable from the transverter to J4, X-verter Jack on the rear panel. Also apply +8 Volts DC to pin 11 of J1 (ACC socket) on the rear panel.



In the receiving mode using the transverter, +9 Volts DC is applied through a 470 ohm resistor to J4, superimposed on the incoming signal from the transverter. Then D18, D19, and D40 are turned ON, and D28 to D33 of the RF amplifier circuit for each band are turned OFF by +9 Volts DC from J4, and D41 is turned OFF by +8 Volts DC from pin 11 of J1. Thus the incoming signal from the transverter is fed to the mixer circuit composed of D11 to D14, through J4, D18, D19, and D40.

In the transmitting mode, J4 is grounded for DC, then D16, D17 and one of D28 to D33 are turned ON and D40 is turned ON and D41 and Q2 are turned OFF by +8 Volts from Pin 11 of J1. Thus T/R relay RL1 turns to the transmit side so the mixed SSB signal from the transmit mixer circuit is fed to the RF amplifier circuit, and the amplified signal is sent to J4 through D16 and D17.

The output level is about 200mV across a 50 ohm load.

6-2-7 EXCITER CIRCUIT

The output from the RF unit J6 is applied to Q1 on the "B" unit and amplified class A to about 500mV. L3 and R6 are negative feedback circuits for obtaining the wide band characteristics from 1.8 to 30MHz. This signal is amplified class B push pull to about 7 Watts by Q2 and Q3. The junction voltage of D2 and D3 is amplified by Q9 and applied to the bases of Q2 and Q3, from the emitter of Q9, for the bias voltage.

D2 and D3 are connected to the cases of Q2 and Q3 respectively, and compensate the bias voltage by the temperature of Q2 and Q3. L8, L9, R10, and R11 are negative feedback components for obtaining a wide frequency range.

6-2-8 FINAL POWER AMPLIFIER CIRCUIT

The 7 Watt signal from the exciter stage is amplified class B push-pull in the final amplifier stage transistors, Q1 and Q2 of the PA unit to a level of about 100 Watts. Bias voltage is treated by D1, D2 and Q3, the same as D2, D3 and Q9 of the "B" unit, and fed to the bases of Q1 and Q2. L4, R7, to R10, C3, C4 and C8 are negative feedback circuits.

6-2-9 ANTENNA SWITCHING CIRCUIT

The Antenna switching circuit is composed of D24 to D26, Q4 and Q5 located on the "B" unit. Q4 is ON when the voltage for transmit (+9V) is applied to its base through R14 during transmit. Accordingly, Q5 is turned ON, which turns on D24 which passes the RF. Q17 is OFF since its emitter voltage is 0, and D25 and D26 are cut OFF. The signal from the PA unit is conducted to the filter section through D24. In this case, the high frequency voltage is rectified by D25 and D26 and this voltage, about 80V, is fed to the cathode of D25. This reverse biases D25 and

D26 and turns them OFF, so output signals do not feed into the receive section.

In the receive mode, the base of Q4 goes to 0V, so Q4 and Q5 are turned OFF, and -9V is applied to D24 through R19 and R108, thus D24 is turned OFF completely. Also +9V is applied to D25 and D26, through Q17 and R70, so D25 and D26 are turned ON, thus incoming signals are fed to the receiving section.

6-2-10 LOW PASS FILTER CIRCUIT

The PA output, which has passed through the antenna switching circuit, is fed through the low pass filter of the filter unit, which serves to attenuate harmonics by about 40dB in order to get a clean transmit output. The low pass filter is set for each band, and is switched by the rotary relay switch. The relay switch is driven by Q1 to Q8. The voltage set in the band switching circuit on the front panel is applied to one of Q1's bases through R8, while the voltage set according to the contact of the rotary relay is applied to the other base through R12. If there is a potential difference between both bases of Q1, Q2 and Q3 amplify the DC current, and Q4 applies the bias voltage to Q5's base. As a result, the multivibrator composed of Q5 and Q6 operate and the rotary relay is driven by Q7 and Q8. The relay continues to operate until the voltage of both bases of Q1 are the same. At that time, Q1's collectors are equal, Q2, Q3, and Q4 are all OFF. Therefore Q7 and Q8 are OFF and the relay stops.

6-2-11 ALC CIRCUIT

L3 of the ALC circuit is an SWR pick up coil, and D3 and D4 are the detection diodes. The voltage of a traveling wave is detected by D3 as a negative DC voltage, and a reflected wave is detected as a negative DC voltage by D4. They pass through the SWR switch (forward-reverse switching) and operate the meter on the front panel. D1 is the ALC detection diode.

The bias voltage determined by R24 in the "B" unit is applied to the cathode of D1 through R2 in the ALC circuit. If RF voltage applied to the D1 anode is above this bias voltage, negative ALC voltage is detected by D1 and fed to the base of Q10 in the "B" unit, through L2 and R3.

The ALC output voltage is DC amplified by Q10, Q11, Q12, and Q13 in the "B" unit and then applied to the gate of Q17 in the "A" unit and controls the driving power to the PA stage. If the antenna connected to the transmitter is inferior in SWR, the detected voltage of the reflected wave is applied to the ALC output circuit Q10 in the "B" unit through D2 and R1. Therefore, the ALC output negative voltage becomes extremely high and so the driving power to the final PA stage becomes low and protects the PA transistors.

6-3 PLL (Phase Locked Loop) UNIT

6-3-1 LOCAL OSCILLATOR CIRCUIT

Q1-Q6 are local oscillator circuits for the conversion type PLL. Each circuit, Q1-Q6 is functional for each band respectively; for example, Q1 is for the 1.8MHz band and Q6 is for the 28MHz band. Each oscillator circuit is basically the same, being a doubler with 3rd overtone crystals.

BAND	XTAL FREQUENCY	OUTPUT FREQUENCY
1.8MHz	27.125MHz	54.25MHz
3.5MHz	28.125MHz	56.25MHz
7 MHz	30.125MHz	60.25MHz
14 MHz	33.625MHz	67.25MHz
21 MHz	37.125MHz	74.25MHz
28 MHz	40.625MHz	81.25MHz

For example, X1 of the 1.8MHz oscillator is a 3rd overtone crystal (27.125MHz) and has L2 in series, which is for the X1 oscillating frequency adjustment. The output tuning circuit connected to the Q1 collector is tuned for 54.25 MHz, which is double the frequency of the fundamental crystal oscillator frequency. The operation of Q2-Q6 is exactly the same as that of Q1, except the output frequencies are different.

6-3-2 VXO (Variable Frequency Crystal Oscillator) CIRCUIT

This circuit is for the oscillation, by 100Hz steps, of the lowest two digits (0.0 - 9.9KHz) of the VCO output frequency of the PLL circuit. The VXO circuit consists of

Q7, Q8, D9, X7, L18, etc. Q8 oscillates on the 15MHz range using X7, D9, D10, and L18, and the oscillating frequency is altered in 100Hz steps by the voltage supplied from the D/A (Digital to Analog) converter.

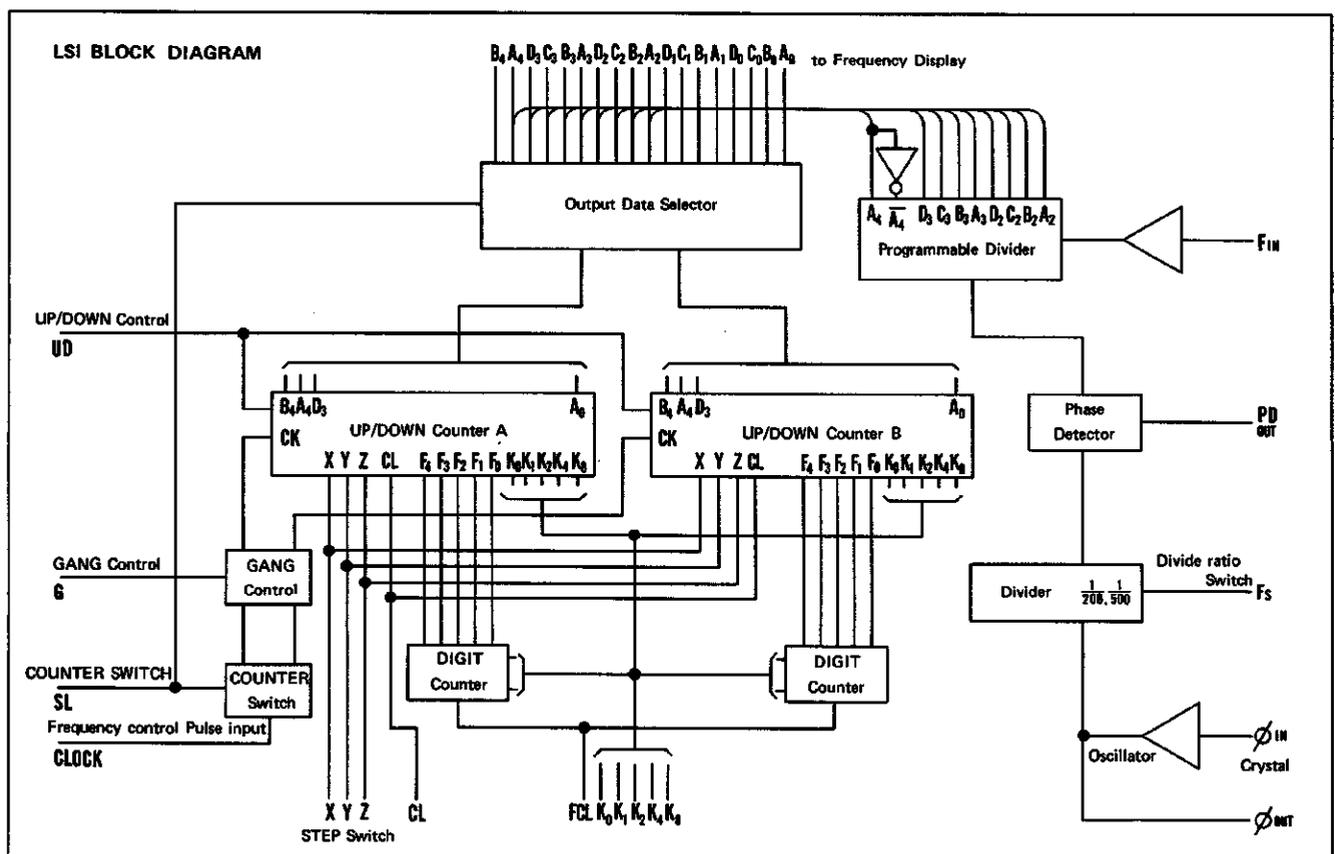
Q7 triples the output signal from Q8 and supplies about 45MHz signal to the mixer circuit of IC7.

6-3-3 MIXER AND AMPLIFIER CIRCUIT

The local oscillator output and VCO output are mixed by IC6 and the mixed signal is taken from Pins 3 and 13 of IC6. The output signal is filtered by L13 and L14, tuned for 42-44Hz, to supply the signal to IC7. The output signal of the VXO circuit is also supplied to IC7 to be mixed with the signal from IC6. The difference of these two signals (1MHz - 3MHz) is filtered by a low-pass filter and is amplified by IC8 up to the level of more than 1.6Vp-p.

6-3-4 PLL LSI CIRCUIT

The output from IC8 is the input to the LSI, Pin 2. The LSI contains PLL and control circuits. To drive the reference frequency of the PLL, a 5MHz crystal oscillator is connected to Pins 3 and 4. This frequency is divided into 1/500 by the fixed frequency divider in the LSI and becomes the accurate frequency of 10KHz. The input from Pin 2 is divided by the Programmable divider in the LSI from 1/100-1/299 which is decided by the controller, fed into the phase frequency detector with the reference frequency of 10KHz. The output is taken from Pin 40 in the form of a pulse from the phase detector according to these two input's difference of phase. The controller circuit consists of two sets of a 4-1/2 digit BCD up-down



counter for the A and B VFOs, the digit selector to control the up-down counters by a parallel data, A/B counter selector, etc.

The CLOCK (CK) and UP/DOWN (UD) signals, which is the output from the tuning knob/chopper circuit, are fed to Pins 12 and 19 respectively. The clock signal is in proportion to the rotation of the tuning knob, and the up/down signal is dependant on the direction of rotation of the tuning knob. When the UD signal is HIGH, the up/down counter increases the frequency in proportion to the number of CK pulses, and when the UD is LOW, the frequency is decreased in the same way. The signal which is determined by the "A" or "B" position of the VFO switch on the front panel, becomes the SL signal. This is fed to Pin 14 and selects either "A" or "B" set of up/down counters.

If you push the Tuning Speed button on the front panel, the signal is locked in the driver unit and fed to Pin 17 as the Y signal. In this case the Kilohertz and Hundred Hertz digits are cleared to 0, the CK signal is fed to the 10KHz digit counter, which results in high speed tuning.

The FCL K0-K8 pins of IC1 are connected to the ACC socket on the rear panel through the driver unit. When a high level signal is supplied to FCL Pin 11 and K0-K2 at the same time, the 10MHz digit figure can be reset, according to the input data at K0-K2. Then when the high level signal is supplied from K0-K8, any figure can be set according to the data from the higher digit to the lower.

For use with an EXT controller, resetting of the frequency can be accomplished in the same manner through the ACC socket.

When the Band Select switch is at the 7-28MHz position, no signal to these pins is required; however, at the 1.8MHz or 3.5MHz position, since a figure of 8 or 5 is necessary on the 100KHz digit for the bottom of each band, when switching bands or turning the power ON, a signal is supplied to these terminals. Therefore, when the Band Select switch on the front panel is at the 1.8MHz or 3.5MHz position, the high level signal is supplied to K8 to set 8 on the 100KHz digit or K4 and K1 to set 5 on the 100KHz digit respectively at the bottom of each band after the FCL input.

By the input of CK and UD signals, or by resetting the up/down counter, the output signals appear at A0-B4 and are supplied to the driver unit. The output signals of A2-A4 are supplied to the PLL programmable divider for determination of the divider ratio in accordance with the up/down.

6-3-4 LOOP FILTER CIRCUIT

The phase-detected signal from Pin 40 of IC1 is fed to the loop filter circuit of IC3. This circuit changes the output of the phase detector, which is a pulse, into a DC voltage, and also decides the response time of the whole loop. The

output is used as the voltage to control the VCO frequency.

6-3-5 VCO CIRCUIT

Q11-Q16 of the VCO unit are FET modified Clapp oscillators, used separately for each band. Q11 is for the 1.8MHz band and Q16 is for the 28MHz band. To obtain the desired frequency, the D/A converter tracks the variation of the voltage supplied to the tuning diodes D11, D14, D17, D20, D23 or D26. Also, the output signal from the loop filter is supplied to the tuning diodes D12, D15, D18, D21, D24, or D27 to lock the VCO frequency.

The oscillation frequency of the VCO is equal to the displayed frequency plus about 9MHz. This output is supplied from P1 to the mixer circuit of the RF unit after the buffer amplifier Q1 and Q2. Also, the VCO output is supplied to the PLL mixer through the buffer amplifier of Q4. Q5-Q10 and D4-D9 are to switch the DC power source to Q11-Q16 according to the operating band.

6-3-6 D/A (DIGITAL TO ANALOG) CONVERTER CIRCUIT

The PLL unit has two D/A converters. R89-R94 are connected to A2-B4 of IC1 and act as a D/A converter for the higher digits. The output signal is amplified by half of IC2 for VCO tracking.

The other D/A converter is for the VXO circuit. The signals from A0-D1 of IC1 are buffered by IC4 and IC5 and supplied to R101-R109 for D/A conversion. The converted signal is fed to the other half of IC2 for off-set voltage setting and amplification to control the frequency of the VXO circuit. This D/A converted voltage changes in steps to give 100Hz step variation to the VXO frequency. Q14 and Q15 are employed to maintain linearity of the D/A converter output for changes in mode. R68 and R70 adjust the correction factor.

6-3-7 POWER CIRCUIT

The power circuit of the PLL consists of Q12, Q13, and IC9. Q13 is turned ON by C144 and applies voltage to IC9. IC9 is a voltage regulator whose output is 8.2 Volts. A part of this output is divided at R55 and R56 and is added to the base of Q12. The emitter of Q12 is connected to the collector of Q13 through a Zener diode (6.1 Volt) and therefore the collector voltage is held at 10.6 Volts. As you can see, the output of the power circuit of the PLL has outputs of 8.2 and 10.6 Volts regulated DC.

Q10 and Q11 are negative voltage regulators. One of Q11's bases is grounded and the other is connected to R53 and R54. R53 is connected to +8.2 Volts, and R54 to the collector of Q10. -9 Volts is added to the emitter of Q10 and its base is connected to the collector of Q11. Q11 works as a differential amplifier, so if there is a difference between the divided voltage at R53 and R54 and the potential of the other base, which is ground, the difference is amplified until the voltage of Q10's collector is the same value as the positive output of IC9. Therefore, the output of -8.2 Volts is achieved.

6-3-8 RIT ON/OFF CONTROL CIRCUIT

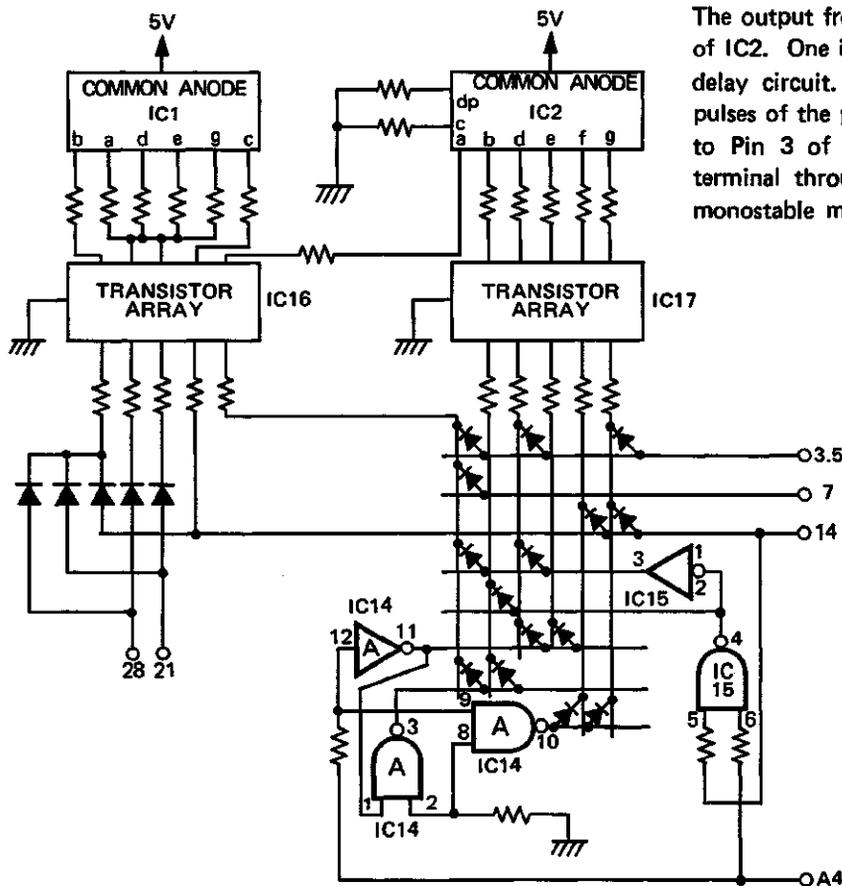
Q9 and Q17 make up the RIT ON/OFF control circuit. The voltage of the RIT IC terminal of J6 is fed to the emitter of Q17 through R51. The terminal is 0 Volts when the RIT is ON and 5 Volts when it is OFF. Q17 prevents the reverse flow from the base of Q9. The collector of Q9 is fed voltage from the RIT control and shunt to ground when the RIT is OFF.

6-4 DRIVER UNIT

6-4-1 FREQUENCY DISPLAY

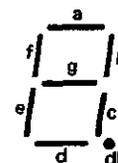
IC1-IC6 are the 7 segment LED indicators for the frequency

10MHz · 1MHz DIGITS DISPLAY SCHEMATIC DIAGRAM



IC14 is switched ON in 28MHz band.

BAND MHz	BAND DATA						10MHz DIGIT							1MHz DIGIT						
	3.5	7	14	21	28	A4	a	b	c	d	e	g	a	b	d	e	f	g		
1.8	L	L	L	L	L	L	-	-	-	-	-	-	ON	-	-	-	-	-		
3.5	H	L	L	L	L	L	-	-	-	-	-	-	ON	ON	ON	-	-	ON		
7	L	H	L	L	L	L	-	-	-	-	-	-	ON	ON	-	-	-	-		
14	L	L	H	L	L	L	-	ON	ON	-	-	-	ON	-	-	ON	ON	ON		
15	L	L	H	L	L	H	-	ON	ON	-	-	-	ON	-	ON	-	ON	ON		
21	L	L	L	H	L	L	ON	ON	-	ON	ON	ON	-	ON	-	-	-	-		
28	L	L	L	L	H	L	ON	ON	-	ON	ON	ON	ON	ON	ON	ON	ON	ON		
29	L	L	L	L	H	H	ON	ON	-	ON	ON	ON	ON	ON	-	-	ON	ON		



The segments "dp" and "c" of 1MHz digit are turned on in any band.

display. IC3-IC6 are driven by IC22-IC25. The input for the drivers is taken from the BCD output of IC1, A0-D3 in the PLL.

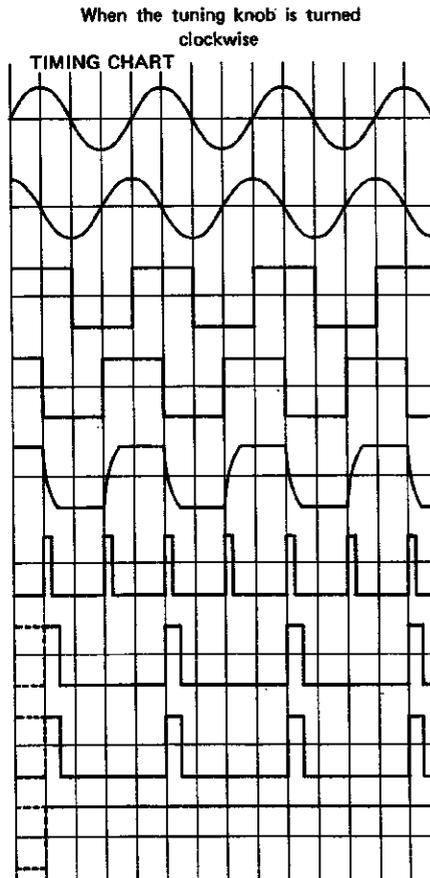
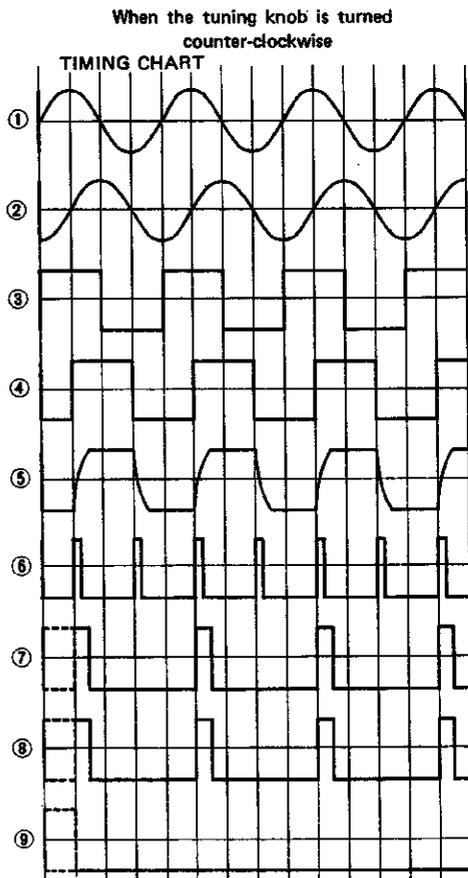
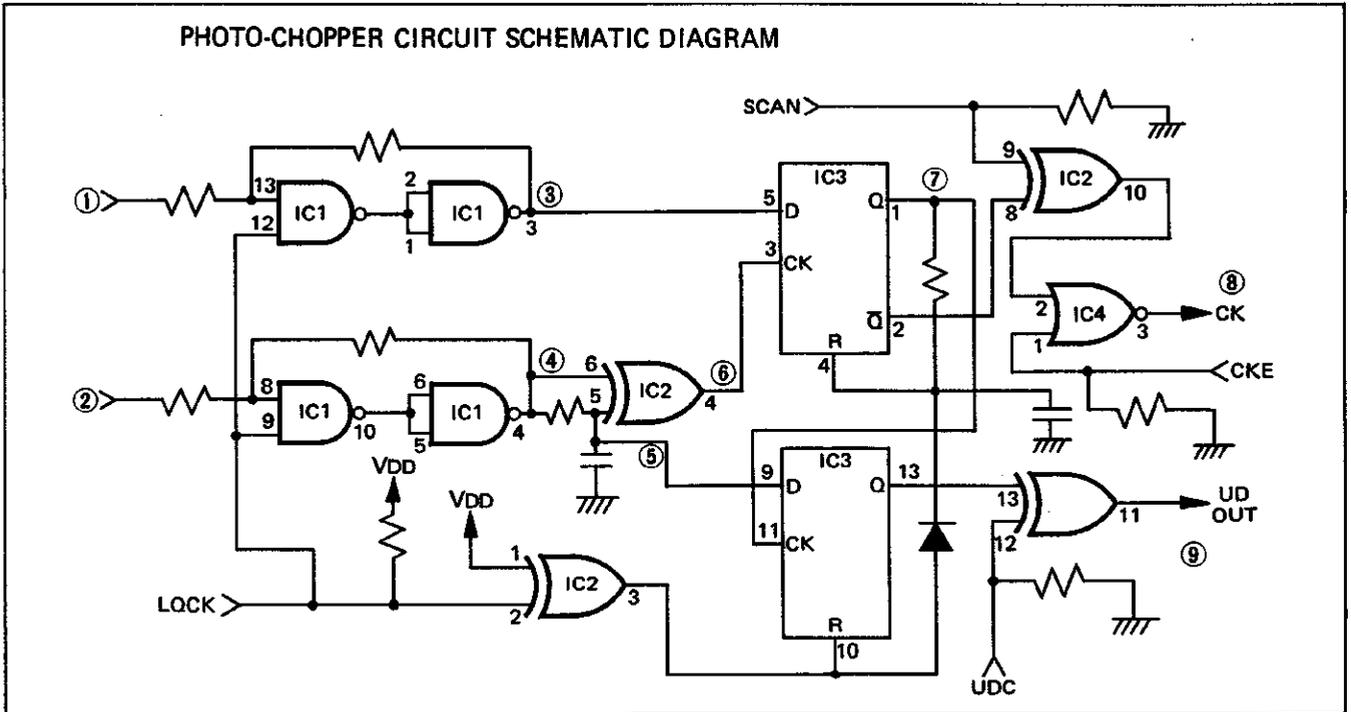
IC1 and IC2 are driven by IC16 and IC17. The input of each band is decoded by IC14 and IC15 and D21-D35 to indicate the MHz and 10MHz digits.

6-4-2 PHOTO-CHOPPER CIRCUIT

D38 and D39 are infrared LED diodes and Q1 and Q2 are the photo transistors. D38 sends a continuous light source to Q1 and D39 sends it to Q2. Between the diodes and the transistors is a windowed chopper disk which is connected directly to the shaft of the tuning knob. The diodes and photo transistors are set up so that the phase difference of 90 degrees results when the disc revolves. The signal from Q1 and Q2 is fed to IC1 on the driver unit. Pin 3 of IC1 is connected to the data input pin (Pin 5) of flip-flop IC3. The output from Pin 4 is divided, and fed to Pins 5 and 6 of IC2. One is fed directly and the other is fed through a delay circuit. IC2 is an exclusive OR gate IC, and both pulses of the positive edge and the negative edge are added to Pin 3 of IC3 and Q output is added to IC3's reset terminal through the delay circuit so IC3 operates as a monostable multivibrator. Q is fed to Pin 8 of IC2, the

scan signal from the ACC socket (if used) is fed to Pin 9 and the output signal is fed to Pin 4 of IC4. This is controlled by the band edge detection signal, fed to Pin 1, and is sent to the LSI in the PLL unit as the Clock (CK) signal. Pin 1 of IC3 is connected to Pin 11 and the delay circuits of R13 and C3 are connected with the data input Pin 9. IC3 is a D type flip-flop, so it latches the signal of the delay circuit.

IC3 gives a High (H) level signal when the tuning knob is turned clockwise, and a Low (L) level signal when the knob is turned counter-clockwise. This (H) or (L) signal goes to Pin 13 of IC2 along with the signal from the ACC socket (if used), which is fed to Pin 12. The output of IC2 is added to the LSI as the UP/DOWN (UD) signal.

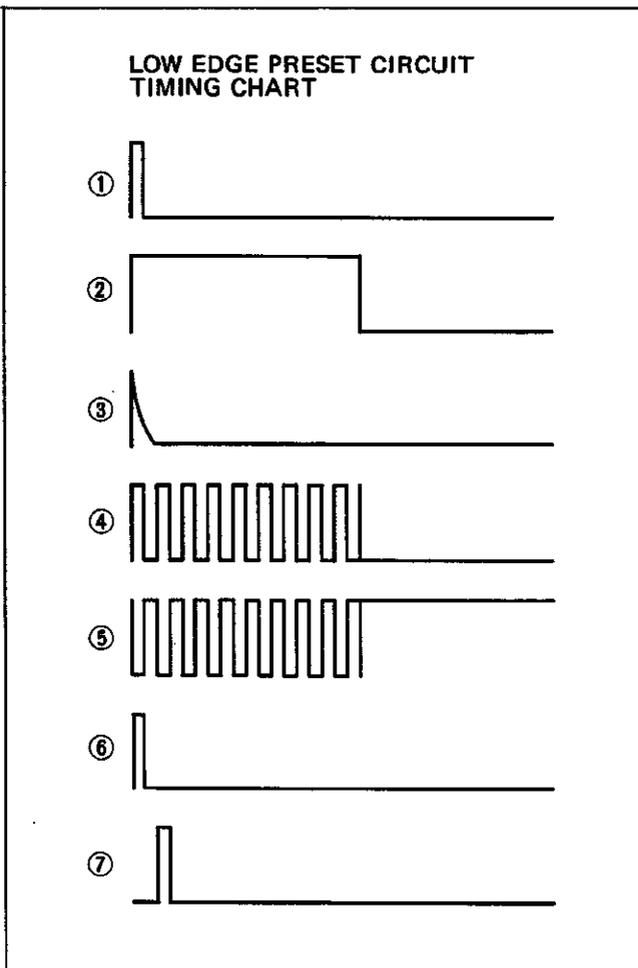
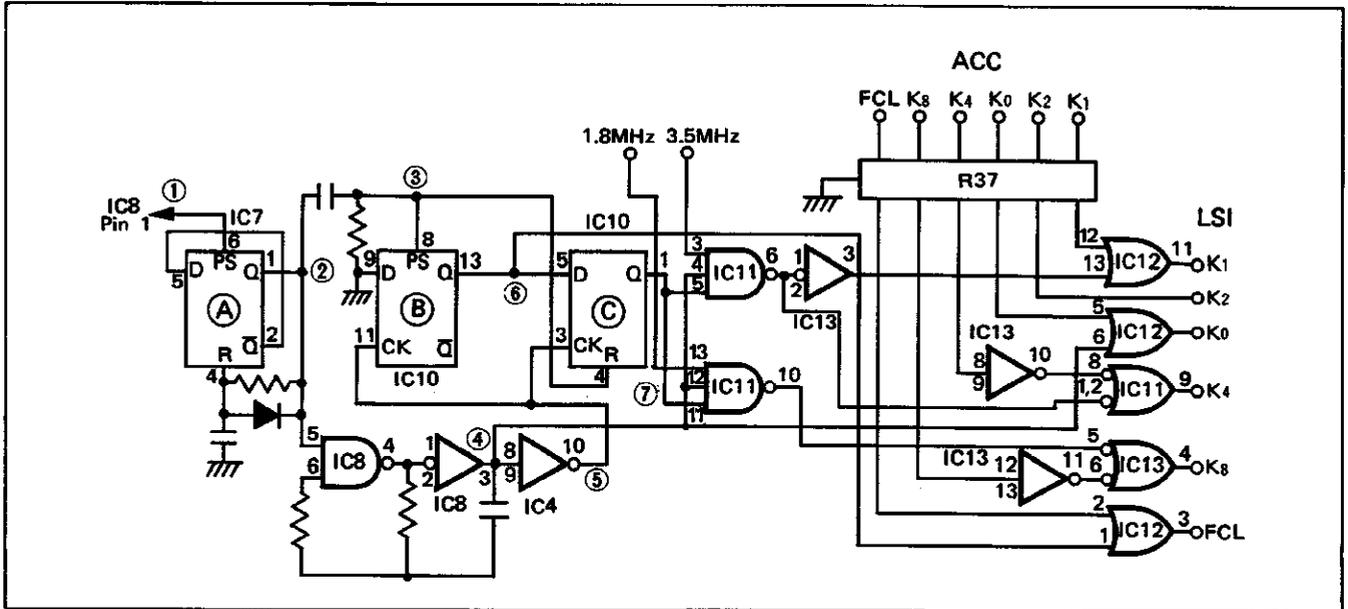


goes LOW and Pin 1 goes HIGH. When the band is 1.8MHz, Pin 13 of IC11 is HIGH and Pin 11 is connected with Pin 1 of IC10, so it is HIGH.

As Pin 12 is connected to the front of the inverter of the oscillator circuits, the three input gates of IC11 are all HIGH in the next half cycle, and the output from Pin 10

goes LOW. IC13's output HIGH level pulse is fed to IC1 in the PLL as K8, therefore the 100KHz digit is set at 8.

When the band is 3.5MHz, it is input to K1 and K4 through the same process and the 100KHz digit is set to 5. As the 100KHz digit begins with 0 in upper bands from 7MHz, this operation is not necessary. When the band is 7MHz or



above, the band data of 1.8MHz and 3.5MHz are LOW, so pin 6 and Pin 10 of IC11 are always HIGH and this circuit does not work.

After the third pulse from IC8, it is used together with the low edge of the band and fed to IC1 in the PLL as the K0 input.

6 - 4 - 5 MATRIX CIRCUIT

The output A0-A4 of the LSI and the signal of each band make up the matrix circuit and this decides the edges of each band.

The band edges are as follows:

BAND	LOW EDGE	HIGH EDGE	FAST TUNING
1.8MHz	1.800.0	1.999.9	1.990.0
3.5MHz	3.500.0	3.999.9	3.990.0
7 MHz	7.000.0	7.499.9	7.490.0
14 MHz	14.000.0	15.199.9	15.190.0
21 MHz	21.000.0	21.499.9	21.490.0
28 MHz	28.000.0*	29.999.9*	29.990.0*

* In the 28MHz band the tuning does not stop when the band edge is reached. It automatically reverts to the other band edge and continues tuning either up or down.

IC21, IC20, and IC18 of the driver unit detect the 0 in the 10KHz, 1KHz, and 100Hz digits. Since they are AND gate circuits, they detect when all three figures are 0, and are used to control the low edge. IC19 detects 0 of the 100KHz level, but it is necessary to distinguish the 1.8MHz

and 3.5MHz bands from other bands since their 100KHz low edge digits are not 0.

D42-D45 detect the lower two figures (9.9) and D40 and D41 detect (9) of the 10KHz level. It is necessary to distinguish these from others since the lower two figures are reset to 0 when the dial tuning speed button is pushed. Q6 is the gate for this distinction. The MHz level and the 100 KHz level of the band edge are determined by the detection of the (0) and (9), the figures which are chosen according to the band of the matrix unit and the matrix which composes the necessary data out of A0-B4 of the output from IC1 in the PLL section "line". Moreover, three kinds of AND gates with the UD signal are taken out and if they agree with each other, CKE output is achieved. If this controls IC4 Pin 4 and becomes the band edge, the CK signal is stopped.

6 - 4 - 6 RIT CONTROL CIRCUIT

Once the RIT switch is pushed down, Pin 12 and Pin 13 of IC27 become HIGH. Any chatter from the switch is absorbed by the delay circuit of R66 and C20. Also the two gates of IC27 are used as inverters and the square pulse is achieved by the Schmitt trigger circuit. The output from IC27 Pin 10 goes to IC28 Pin 3 of the CK terminal. This IC works as a T type flip-flop so the level is reversed every time the RIT switch is pushed. The signal at Pin 1 of IC28 (O output) is fed to Q15 through R62 and turns ON this transistor. Q15's collector is connected with D4, an LED, so when Q15 conducts, the LED is lit and indicates that the RIT circuit is ON. The LED also indicates the center of the RIT control.

Also Pin 1 of the IC28 is fed to Pin 6 of IC27 as well and it functions with the send signal HIGH level during receiving, of Pin 5. The output of Pin 4 switches Q9, in the PLL circuit, and controls the RIT circuit.

6 - 4 - 7 DIAL LOCK CIRCUIT

If the dial lock button is pushed, Pin 1 and Pin 2 of IC27 go LOW. Any switch chatter is absorbed by R6 and C19. The output of IC27 inverter is fed to Pin 11 (CK terminal) of IC28. IC28 also works as a T type flip-flop and output levels are repeated every time the input is applied. The Q output from Pin 13 turns Q4 ON, through R69, and shows that the dial lock button has been pushed, by lighting the dial lock indicator LED, D2. Also Pin 9 and Pin 12 of IC1, and Pin 2 of IC2 are connected to Q4's collector and they go to LOW level, so the clock pulses from Q1 and Q2, the photo transistors, are not fed to the LSI circuit, and the frequency is locked.

6 - 4 - 8 TUNING SPEED CIRCUIT

If the tuning speed button is pushed, the input goes to Pin 3 of IC29 through C16. R59 and C17 absorb any chattering. IC29 is also a T type flip-flop. The Q output of Pin 1 is combined with Q3. This makes the dial fast LED indicator, D1, light. Also, the signal from IC29 Pin 2 (\bar{Q}) is sent to the LSI circuit and the Y terminal (Pin 17) of IC1 in the PLL unit goes HIGH, and so fast tuning is

achieved.

6 - 5 POWER UNIT

6 - 5 - 1 13.8 VOLT DC REGULATED POWER SUPPLY CIRCUIT

Voltage supplied to Pin 1 (ground) and Pin 2 (positive) of the power source connector J6 on the rear panel is fed to the reverse polarity protection circuit composed of diodes D9, D10, and Fuse F1. This voltage serves as a direct power source for the final amplifier stage of the IC-701. Power for all other circuits is fed through the power switch located on the front panel. Voltage passing through the power switch is fed to the PLL unit and distributed to the 13.8 Volt regulated power supply circuit. This supply is composed of transistors Q13 through Q15 in the PA unit and will reduce a voltage applied to the power source connector to 13.8 Volts when the voltage exceeds that. The output voltage of this circuit is fed to the "B" unit, AF power amplifier circuit, dimmer circuit, and at the same time, serves as an input power source for each 9 Volt regulated power circuit.

6 - 5 - 2 9 VOLT REGULATED POWER SUPPLY CIRCUIT

Regulated -9 Volts is produced using the DC-DC converter, IC2, in the "B" unit. The regulated power supply circuit for +9 Volts is composed of Q40, Q42, Q43, and D36 in the "A" unit. D36 detects the difference between output voltages and Q42 and Q43 amplify it. Then Q41 controls the voltage. The regulated 9 Volt supply for receive is taken from the emitter follower circuit of Q37 to the base of which the 9 Volt standard voltage of D36 is fed. In the transmit mode, the base of Q37 is connected to ground by the PTT switch or Q9 through D26 and the output voltage ceases. The 9 Volt supply for transmission is taken from the emitter of Q39, and its base is supplied with the 9 Volt standard voltage through D30. In the receive mode, the base of Q39 is connected to ground through Q38 being ON and the output voltage ceases.

6 - 6 OTHER CIRCUITRY

6 - 6 - 1 THERMO-SENSOR CIRCUIT

This circuit controls forced air cooling for the final stage of the transmitter when the temperature of the final transistors become excessively high. The temperature of the transistors is detected by the POSISTOR R13 to vary base bias of Q4. Q4 and Q5 compose a Schmitt trigger circuit whose output is amplified by Q6 and Q7 to turn ON and OFF the DC fan motor. Another Schmitt trigger circuit composed of O8 and O9 is provided to turn ON Q10 if the final stage transistor temperature rises dangerously high. The voltage built up by D6 and R38 through Q10 conducted varies the bias for the constant voltage circuit in the motor to run it at high speed. With this high speed, the fan gives more air flow through the heatsink which keeps the temperature of the final stage transistors low

AC POWER SUPPLY

enough to prevent damage. To indicate the dangerously high temperature, all the luminous indicators on the front panel begin blinking. This is done by the autodimmer circuit composed of Q14, and Q15. The output of the free running multivibrator, Q11 and Q12, and driven by the voltage through Q10 is applied to the circuit.

6-6-2 AUTODIMMER CIRCUIT

R20 in the driver unit is a cadmium sulfide (CdS) element to detect brightness of the surrounding light. Its output is applied to the Schmitt trigger circuit composed of Q14 and Q15 through the dimmer change over switch in the "B" unit. The output of this circuit is connected to Q16 which conducts when the surroundings are bright and a voltage of about 12.5 Volts is obtained from the emitter of Q16. This voltage turns ON lamps through D9. The output from the emitter of Q16 through R66 is fed to the Zener diode (D37) in the "A" unit to obtain 5 Volts which is driven by Q1 to light each LED. When the surroundings are dark, Q16 turns OFF and the voltage for lamps on the front panel is fed from the 13.8 Volt line through R20 (47 ohms) which drops the supply voltage for lamps to about 8.5 Volts. With Q16 in the "B" unit is OFF, a voltage of about 7 Volts is obtained from the Zener diode D8 and fed to R66 through D28. This voltage is reduced by half with R66 and R247 in the "A" unit and driven by Q1 to light each LED with about 2 Volts.

6-6-3 VOX AND CW BREAK-IN CIRCUIT

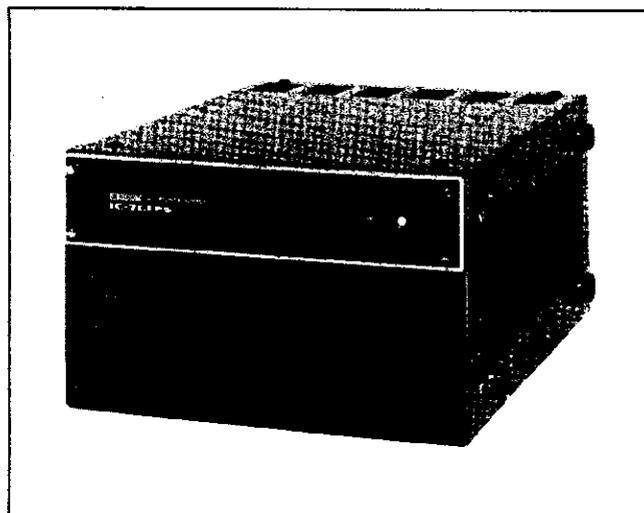
The voice signals from the microphone are amplified by Q18 through Q20 and detected by Q21 to drive Q22. Q22 switches the power supply line ON and OFF corresponding to voice signals. C78, R120 (VOX DELAY control) and R121 compose a circuit to determine the time constant to the VOX circuit. The output of this circuit is amplified by Q7 and Q8 to drive Q9 to switch the transmit circuit ON and OFF. The anti-VOX circuit prevents erroneous operation of the VOX circuit caused by sound from the speaker. Signals from the speaker are amplified by Q24, Q25, and Q26 through R131 (ANTI VOX GAIN Control). These are then fed to Q23 to raise emitter voltage of Q21 which reduces the sensitivity of the detector circuit.

The CW break-in circuit is composed of Q5 and Q6 amplifying the signals from the key at Q5 and feeding it to Q6 for switching the power supply voltage ON and OFF. This ON-OFF voltage is fed to the circuit composed of C14, R32 (CW DELAY Control) and R33 having the specified time constant to control the transmit circuit by way of Q7, Q8, and Q9.

6-6-4 CW MONITORING CIRCUIT

Q7 in the "B" unit is a CR phase shifting oscillator whose frequency is set to about 900Hz. Q8 is a switching circuit to make and break a ground path for C45 so as to vary the gain of Q7 for switching the oscillation ON and OFF. The oscillating output passing through R32 (CW MONI volume control) is fed to the 8th Pin of the AF amplifier circuit IC1 to drive the speaker.

This set is designed to operate on 13.8 Volts DC. However, a special AC power supply, the IC-701PS is available for AC operation. The IC-701PS provides 13.8 Volts at 20 Amps. When connected to the IC-701 transceiver, the power supply ON-OFF operation is remotely controlled by the power ON-OFF switch on the IC-701.



■ SPECIFICATIONS

- Number of Semi-conductors: Silicon diode 3
LED 1
- Input Voltage: 117/230V AC (50/60Hz)
(suitable voltage):
- Allowable voltage fluctuation: $\pm 10\%$ of input voltage (suitable line voltage)
- Input Capacity: 360VA (at 20A load)
- Output Voltage: DC13.8V (at 20A load)
- Max. Load Current: 20A (10 minutes ON/10 minutes OFF 50% duty cycle)
- Polarization: Negative ground
- Internal Speaker: 125 x 77 mm
2 watt nominal input
4 watt maximum
- Dimensions: 110(H) x 180(W) x 260(D) mm
- Weight: Approximately 8 kg
- Accessories included: AC power cord 1
Spare fuse 2
Speaker cord 1

SECTION 7 MAINTENANCE AND ADJUSTMENT

This set has been completely adjusted and checked at the factory so no adjustment should be necessary. However, during prolonged use, the preadjusted oscillating frequency, etc. may need slight adjustment. If this becomes necessary, the following procedures should be followed. As the adjustments other than those specified herein require high skill and expensive equipment, for best results the set should be sent back to the dealer or ICOM distributor for adjustment.

7-1 MEASURING INSTRUMENTS REQUIRED FOR ADJUSTMENT

- (1) FREQUENCY COUNTER
 FREQUENCY RANGE 1 - 90MHz
 ACCURACY BETTER THAN ± 1 ppm
 SENSITIVITY 100mV or BETTER
- (2) SIGNAL GENERATOR
 FREQUENCY RANGE 1.8MHz - 30MHz
 OUTPUT VOLTAGE -20 - 90dB (0dB = 1 μ V)
- (3) MULTIMETER
 20K Ω /Volt or better
- (4) AC MILLIVOLTMETER
 MEASURING RANGE 100-200mV
- (5) RF VOLTMETER
 FREQUENCY RANGE 1.8-30MHz
 MEASURING RANGE 0.1-1V
- (6) RF WATTMETER
 MEASURING RANGE 20-100 Watts
 FREQUENCY RANGE 1.8-30MHz
 IMPEDANCE 50 OHMS
 SWR LESS THAN 1.1
- (7) AF OSCILLATOR
 OUTPUT FREQUENCY 200-3000Hz
 OUTPUT VOLTAGE 0-100mV
 DISTORTION LESS THAN 0.1%
- (8) OSCILLOSCOPE SYNCROSCOPE
- (9) DIGITAL VOLTMETER
 MEASURING RANGE 0.001-10V 3-1/2 DIGITS

7-2 RECEIVER ADJUSTMENT

7-2-1 BFO FREQUENCY ADJUSTMENT

The BFO oscillator frequencies differ in accordance with

each mode and for transmitting and receiving operation. Adjust corresponding trimmer capacitors and trimmer resistors as shown in the table below. Connect the input terminals of the frequency counter across the lead wire of the BFO output D10, or D11, in the B unit, and ground. Connect a digital voltmeter to the cathode of D36 of the "A" unit and check voltage for 8V DC ± 100 mV. Adjust R202 as necessary for 8V. This is a critical adjustment.

- (a) Place the set in receive. Make frequency adjustments as follows.

RECEIVE OPERATION

MODE	ADJUSTING POINT	BFO OUTPUT FREQUENCY
USB	C66 (B unit)	9.01300 MHz (note 2)
LSB	C75 (B unit)	9.01000 MHz (note 2)
CW (CW-N)	C69 (B unit)	9.01320 MHz
RTTY	C72 (B unit)	9.014525MHz

- (b) Turn ON the speech processor (COMP) but keep the knob fully counterclockwise to suppress RF power from being transmitted. Then place the set in the transmit mode.

MODE	ADJUSTING POINT	BFO OUTPUT FREQUENCY
CW (CW-N)	C78 (B unit)	9.01240MHz
RTTY	R87 (B unit) (S3 in B unit to N)	9.01223MHz (note 1)
	R85 (B unit) (S3 in B unit to W)	9.01155MHz (note 1)

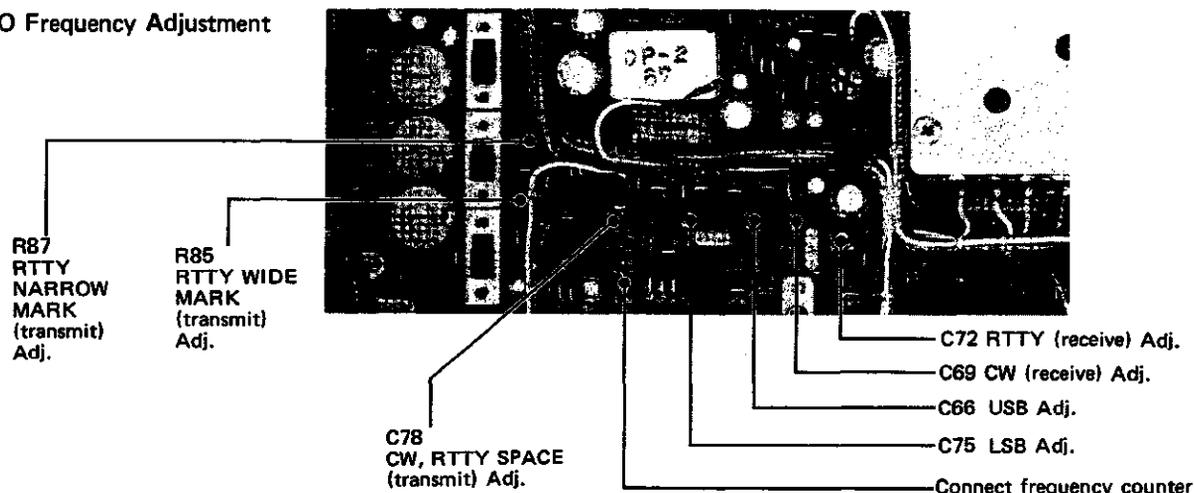
NOTE 1:

S3 in the "B" unit is a change over switch (RTTY Switch) for narrow modulation (170Hz shift width) and wide modulation (850Hz shift width) RTTY transmission, which is located under the access cover.

NOTE 2:

Fine adjustment for the SSB carrier point should be performed in accordance with adjustment "4", CARRIER POINT ADJUSTMENT, in the transmitter section adjustment procedures.

BFO Frequency Adjustmt



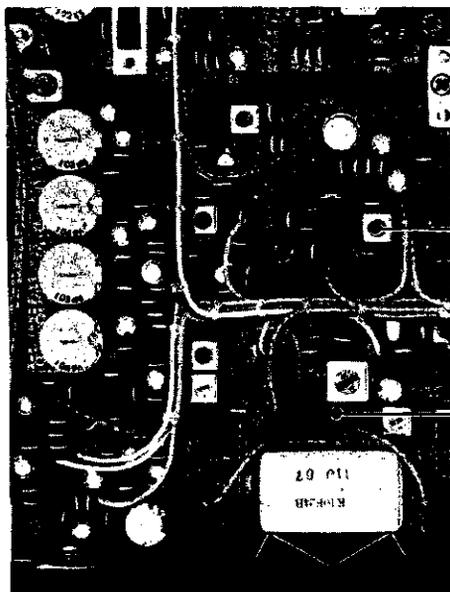
7-2-2 VXO FREQUENCY ADJUSTMENT

Connect the input terminals of the frequency counter across CP1 and ground of the "A" unit and perform the following adjustments.

- (a) Place the set in receive in the CW (P.B. TUNE control centered) and adjust coil L1 for a frequency of 19.76330MHz.
- (b) Then set the mode to USB or LSB and adjust R28

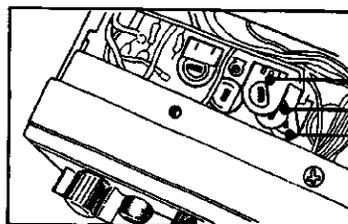
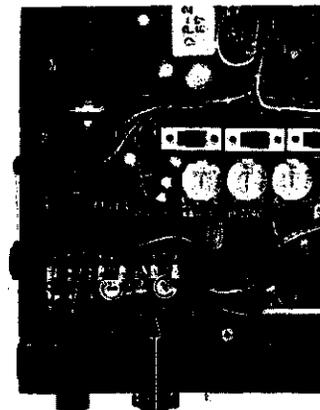
on the front inside of the set to obtain 19.75970MHz with the P.B. TUNE knob turned fully counter-clockwise. Set P.B. TUNE knob to the center position and adjust R27 for 19.7615MHz. R28 and R27 adjustments interact so repeat several times until no further change is noted. Set P.B. TUNE knob fully clockwise. If previous adjustments are correct, frequency should read 19.7631MHz.

VXO Frequency Adjustment



L1
Adjust core for
19.7633MHz
in CW mode

CP1
Connect frequency
counter

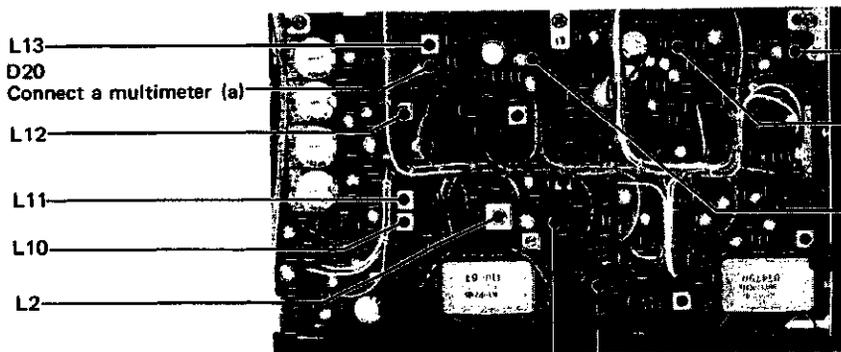


R40
R28
R27 Adjust for
19.7597MHz
described in (b)

7-2-3 AGC SENSITIVITY AND S METER ADJUSTMENT

With the receive frequency set to 14MHz, mode to USB AGC fast, RF GAIN fully clockwise, connect the signal generator to the antenna connector. **MAKE SURE THAT THE SET IS IN THE RECEIVE MODE!! IF NOT, THE GENERATOR WILL BE DAMAGED!!**

- (a) Adjust the signal generator output to minimum and connect the multimeter set to the 10 Volt DC range across the plus lead of D20 in the "A" unit and ground. Make sure no signal is applied to the set. Adjust R168 in the "A" unit to get 4.0 Volts on the multimeter.



L13
D20
Connect a multimeter (a)
L12
L11
L10
L2

R168
Adjust to get 4.0V on the
multimeter (a)
R153
Adjust to get 200mV on the
AC millivoltmeter (c)
R44
Connect AC millivoltmeter (c)

(A unit)

R16
Adjust to make the S meter read S=9+60dB

R22 Adjust to make the S meter read S=1.

Sensitivity and S meter adjustment

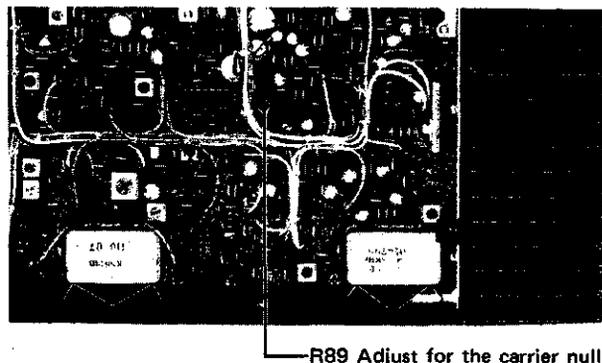
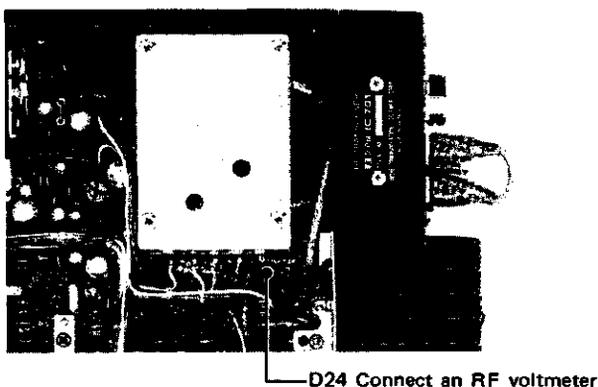
- (b) Adjust the signal generator frequency to the receive frequency (so that a beat note is heard) and set the signal generator output to make the S meter read S=5. Adjust coils L2, L10, L11, L12, and L13 in the "A" unit to get the maximum deflection on the S meter. If operating properly, a 4dB signal at the antenna terminal should read S1. If the reading is slightly off, minor adjustment of L1 and L5 on the RF unit may correct the discrepancy.
- (c) Increase the signal generator output to +90dB (+84dB loaded) and adjust R153 in the "A" unit to obtain 200 milli Volts on the AC millivoltmeter connected across the lead of R44 and ground.
- (d) S meter adjustment begins with setting the signal generator output to 4dB and adjusting R22 to make the S meter read S1. Then increase the signal generator output to 84dB and adjust R16 to make the S meter read S9+60dB. If this adjustment is required, readjustment of the speech processor is necessary.
NOTE:
DO NOT ADJUST L5, L6 or L9 as ripple will result.

7-3 TRANSMITTER ADJUSTMENT

First, connect an RF wattmeter (impedance of 50 ohms) to the antenna connector.

7-3-1 CARRIER SUPPRESSION ADJUSTMENT

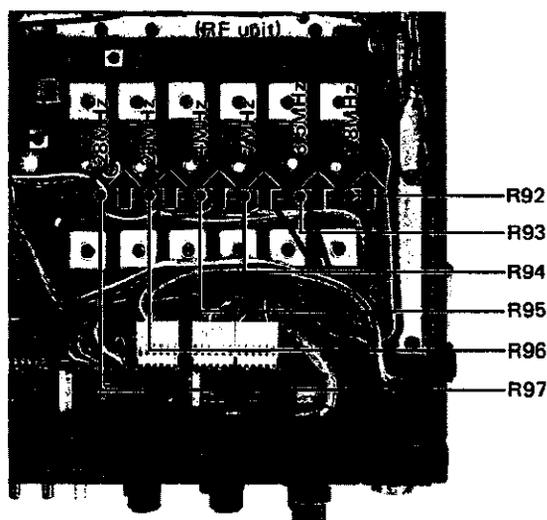
Place the set in the transmit mode in USB or LSB, 14MHz band, the compressor OFF and the MIC GAIN fully counter-clockwise. Connect an RF voltmeter (set to a high voltage scale) across the lead of D24 in the "B" unit and ground. With an appropriate meter range to have suitable deflection for reading, adjust R89 in the "A" unit for minimum.



7-3-2 GAIN ADJUST (FOR EACH BAND)

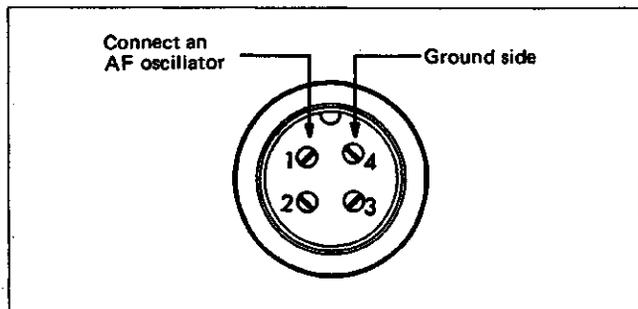
Each amplifier stage, mixer stage, etc., has variations in gain at different frequencies. This requires individual adjustment of the ALC and Compressor levels at every band selection. This set provides a gain adjustment circuit for each band and by presetting these circuits, easy operation can be expected. The following is the adjustment procedure.

- (a) Place the set in transmit, RTTY, 1.8MHz and adjust the compressor knob to obtain 20 Watts indication on the RF wattmeter. Turn R92 in the RF unit for maximum RF power and adjust the compressor knob again for 20 Watts indication on the RF wattmeter. (DON'T READJUST THE COMPRESSOR KNOB HEREAFTER).
- (b) Set to receive, change bands to 3.5MHz and set again to Transmit. Adjust R93 in the RF unit for maximum RF output. Turn R93 in the direction as indicated by the arrow to set the RF power to 20 Watts.
- (c) Similarly, adjust R94, R95, R96, and R97 for maximum RF output in the 7, 14, 21, and 28MHz respectively and turn each resistor in the direction as indicated by the arrow to set the RF output to 20 Watts.

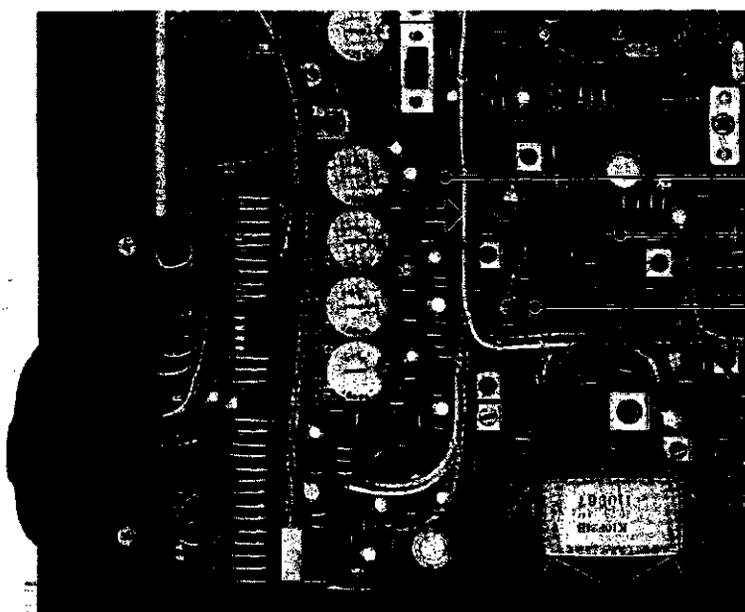


7-3-3 COMPRESSOR CIRCUIT ADJUSTMENT

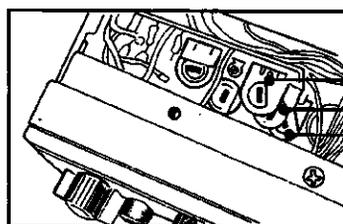
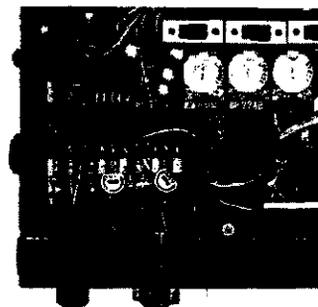
Place the set in the transmit mode in USB at 14MHz and turn the compressor and mic gain knobs fully clockwise. Then as shown in figure, connect the AF oscillator across Pin 1, the input, and pin 4, ground, of the mic connector and apply an input of 100mV, at 1000Hz. Turn R68 fully in the direction as indicated by the arrow. (clockwise)



Compressor circuit adjustment



- R68
Described in (b)
- R25
Described in (a)
- R213
Described in (c)



- R40 Described in (d)
- R28
- R27

- (a) Place the meter switch to ALC and adjust R25 so that the meter indicates 60% on the Po scale.
- (b) Adjust R68 in the "A" unit so that the meter indicates 50% on the Po scale.
- (c) Place the meter switch to COMP and adjust R213 in the "A" unit so that the meter indicates its full scale.
- (d) Then turn the mic gain knob on the front panel fully counter-clockwise and adjust R40 so that the meter just begins to swing and set it in that position.
- (e) Performing the adjustment shown in (d) will affect the adjustments done in (c). Repeat adjustments in (c) and (d) until no difference is present.

7-3-4 CARRIER POINT ADJUSTMENT

Place the set in transmit in USB at 14MHz and turn the compressor OFF with the mic gain knob turned fully clockwise. Connect the AF oscillator across Pin 1 (input) and Pin 4 (ground), of the mic connector.

- (a) Set the frequency of the AF oscillator to 270Hz and the output at a level that gives approximately 20 Watts output on the watt meter.
- (b) Then vary the AF oscillator frequency to 2700Hz and adjust C66 in the "B" unit to obtain 20 Watts on the meter.
- (c) Performing the adjustments shown in (b) will affect the adjustment done in (a). Repeat adjustments in (a) and (b) until no difference is present.

- (d) Place the MODE switch to LSB and adjust C75 the same as the adjustments done in (a), (b) and (c) using C75.

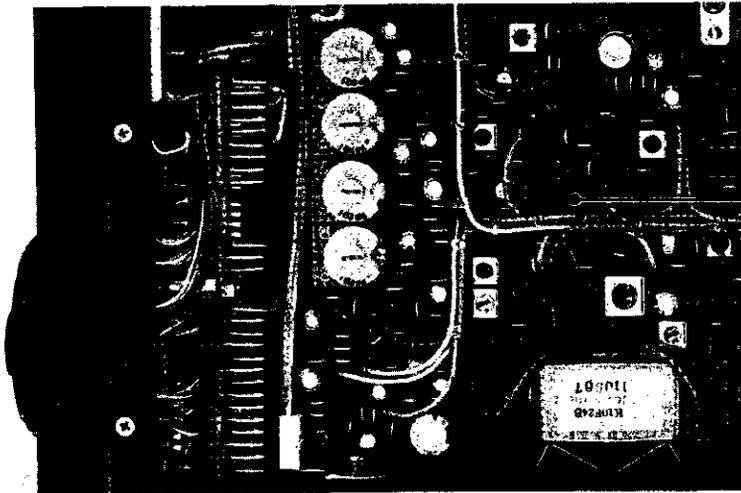
7-3-5 SPEECH PROCESSOR ADJUSTMENT

When the compressor is turned ON, the SSB signal passes through the crystal filters FL1 and FL2 in the "A" unit. Since the rated center frequencies of these filters differ from the actual center frequencies by a certain amount, deterioration of audio quality results due to narrowing of the SSB band width. To eliminate this problem, the VXO frequency is shifted to cover the drift mentioned above in the following adjustments. Place the set in the transmit mode in USB at 14MHz with the mic gain and compressor knobs turned fully clockwise and connect the AF oscillator output across of the mic connector as before with the frequency set at 2600Hz.

- (a) With the MODE switch in USB, adjust the AF oscillator output level to obtain 20 Watts output on the Watt meter.
- (b) Then change the mode to LSB and adjust R2 in the

"A" unit to obtain the same 20 Watts.

- (c) Repeat the adjustments in (a) and (b) until the RF output for USB and LSB are the same. (20 Watts).



Compressor pass band adjustment

R2 Described in (b)

(A unit)

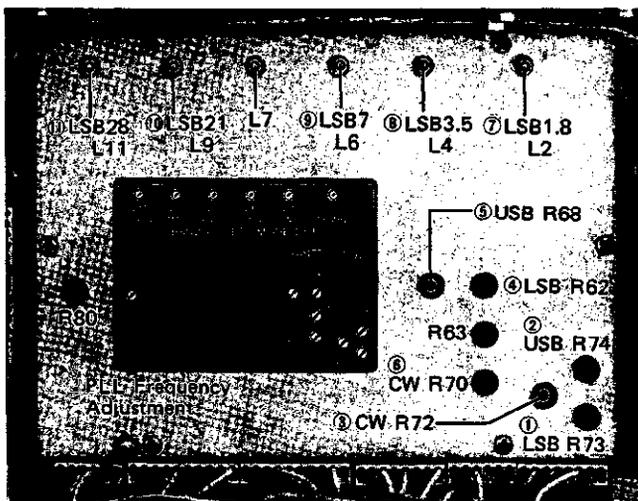
7-4 PLL UNIT ADJUSTMENT

7-4-1 FREQUENCY ADJUSTMENT

This adjustment can be performed without removing the PLL unit seal. Remove the plug from J2 in the RF unit and connect the counter to coax 2 – output of VCO low

pass filter. Select the 1.8MHz band and open the top cover to place the FREQ SET control to the 12 o'clock position. Align the frequencies referring to the counter display as shown in the following table.

MODE	FREQUENCY DISPLAY	ADJUSTING POINT	DESIRED FREQUENCY
① LSB	14.000.0	R73	23,010.0KHz
② USB	14.000.0	R74	23,013.0
③ CW	14.000.0	R72	23,012.4
④ LSB	14.009.9	R62	23,019.9
⑤ USB	14.009.9	R68	23,022.9
⑥ CW	14.009.9	R70	23,022.3
⑦ LSB	1.800.0	L2	10,810.0
⑧ LSB	3.500.0	L4	12,510.0
⑨ LSB	7.000.0	L6	16,010.0
⑩ LSB	21.000.0	L9	30,010.0
⑪ LSB	28.000.0	L11	37,010.0



7-4-2 VCO COIL ADJUSTMENT

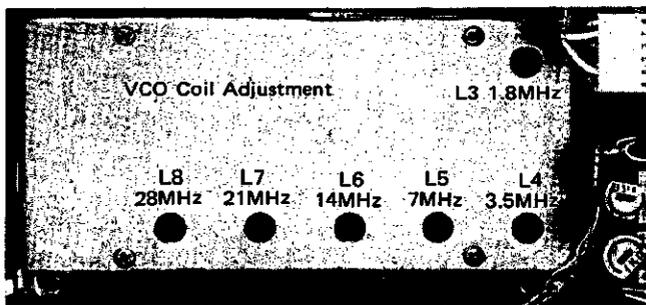
Measure the voltage at Pin 5 (PD) of J1 in the VCO unit with the multi-meter or oscilloscope taking care to avoid accidental shorts or grounds. As the measured voltage will vary to both polarities, a calibrated, syncroscope should be used for ease of adjustments. The adjustments should be performed as follows.

The ideal voltage range as we tune from one end of a band to the other is -3V to 0V. The -3 Volt level is maximum and must not be exceeded. Initially check for the maximum positive voltage and set it to zero with the adjustment shown for the band. Then tune through the band for the low voltage. If it is more positive than -3V, no further adjustment is necessary. If the low point is more

negative than -3 Volts, readjust the control for -3 Volts maximum negative. You will then find that the hi point will swing positive. Up to +4 Volts is acceptable.

Tune over all bands with the tuning knob placed in the dial fast mode and make sure the voltage at J1 PIN 5 falls

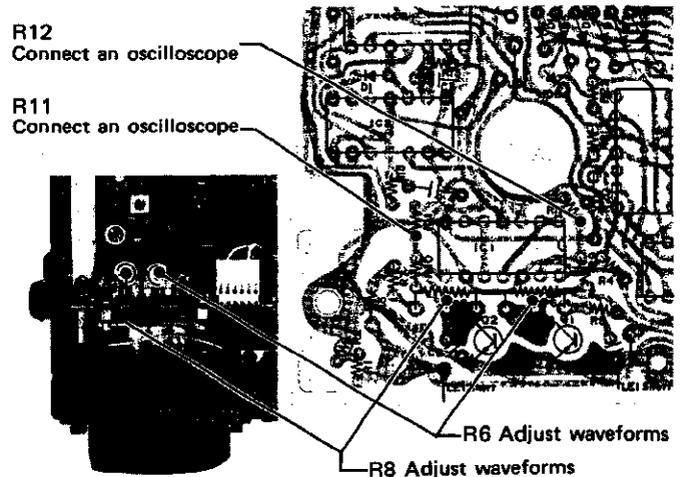
BAND	VOLTAGE AT J1 PIN 5	ADJUSTING POINT
1.8MHz	-1 ~ -2V	L3
3.5MHz	"	L4
7 MHz	0 ~ -1V	L5
14 MHz	+1 ~ 0V	L6
21 MHz	0 ~ -1V	L7
28 MHz	+1 ~ +2V	L8



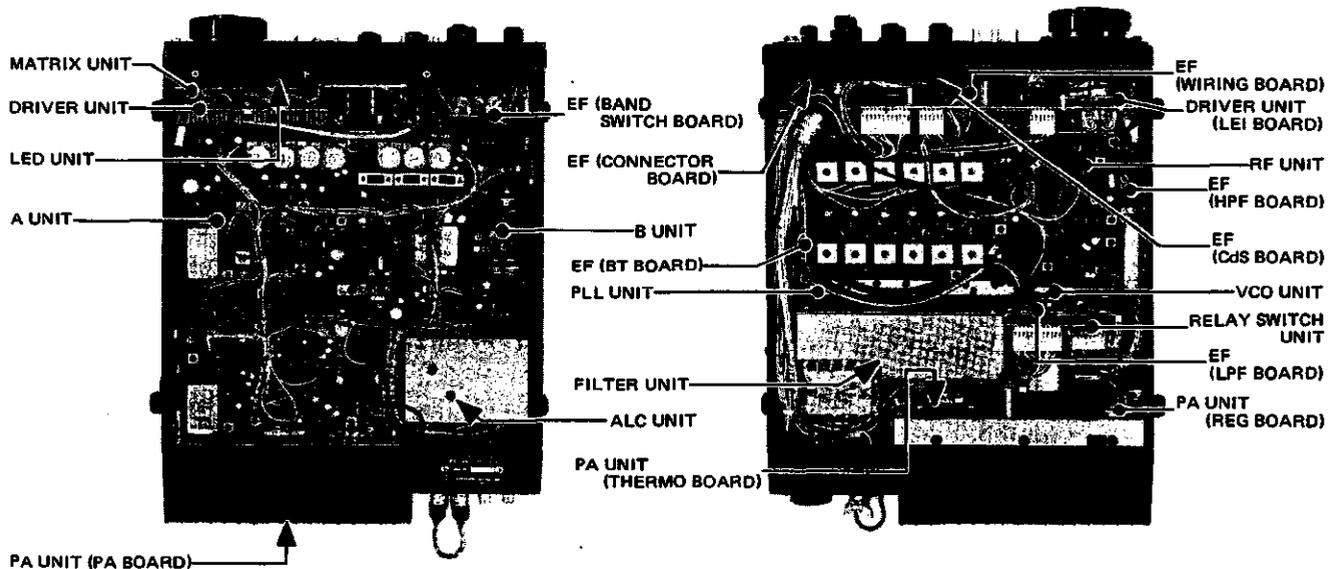
within -3 to +4 Volts. If not, repeat the adjustments.

7-4-3 DISPLAY UNIT ADJUSTMENTS

Set the oscilloscope to the DC 1 to 2V/cm and connect it to R11 and release the brake of the tuning knob. Then turn fast the tuning knob and adjust R8 so that vertically symmetrical waveforms are obtained. Similarly, connect the oscilloscope to R12 and adjust R6. After adjustment, lock the tuning knob again.



UNIT LAYOUT



SECTION 8 TROUBLE SHOOTING

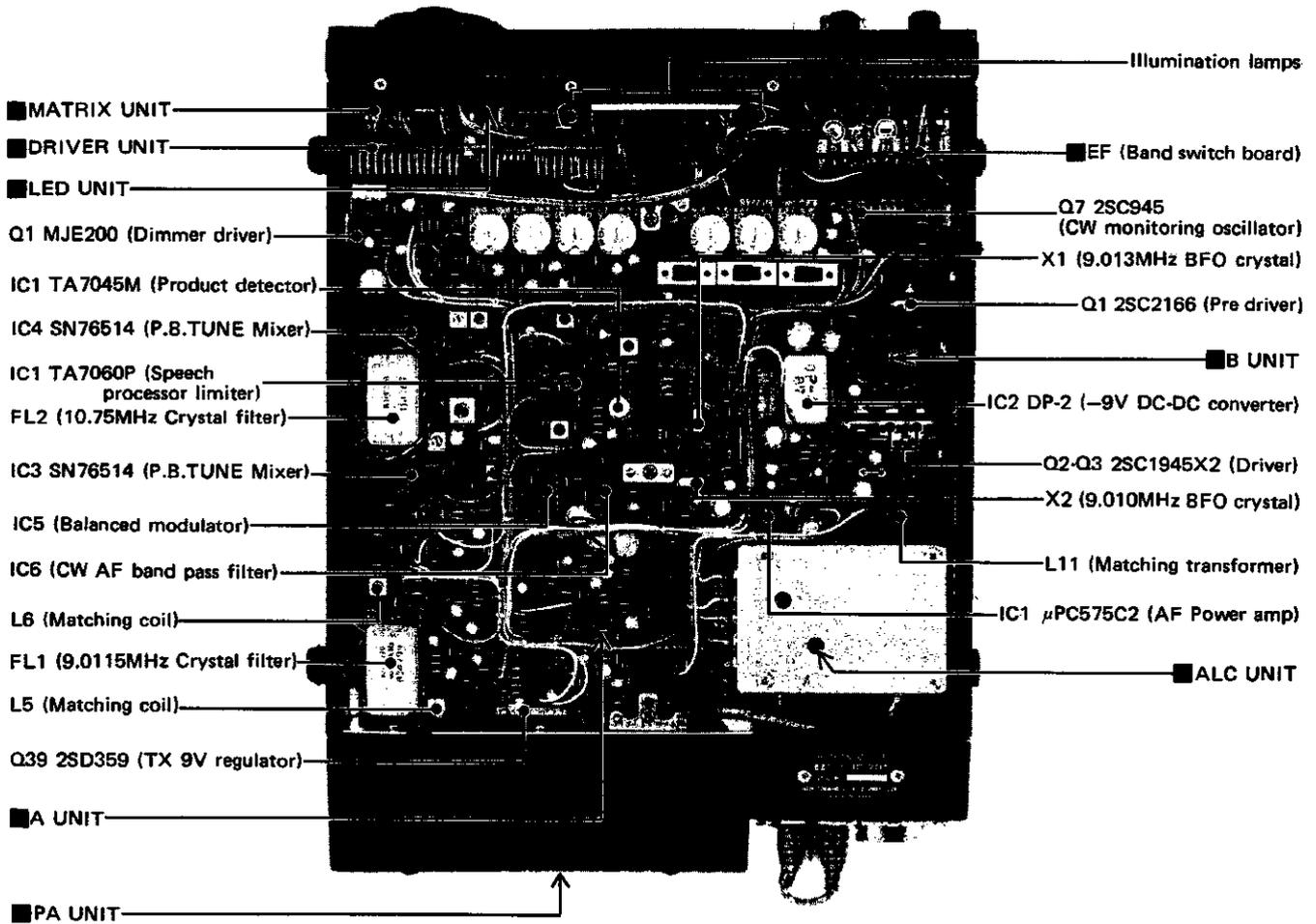
Your IC-701 has been tested very carefully at the factory before shipping. The chart below has been designed to help you correct any problems which are not equipment malfunctions. If you are not able to locate the problem and/or solve it through use of this chart, please contact your dealer or ICOM distributor for assistance.

Problem	Possible Cause	Solution
1. Power does not come on when the switch is pushed.	Power cable is improperly connected. Power cord is connected with the polarity reversed. Blown fuse. IC-701PS MAIN switch is in the OFF position.	Carefully reconnect power cable. Disconnect the power cord, replace the blown fuse, then reconnect the power cord observing proper polarity. Check for the cause, then replace the fuse with a spare one. If using an IC-701PS, also check it's fuse. Place MAIN switch in ON position.
2. The frequency display on the LED's keeps changing bands and a clicking noise is heard.	The Band Select switch is in the EXT. position.	Set the Band Select switch to the desired band.
3. No sound comes from the speaker.	AF GAIN Control knob is completely counterclockwise. The unit is in the transmit mode, either by the T/R switch or the PTT switch on the microphone. The external speaker cable is connected to the unit but not to the external speaker jack on the power supply. The internal speaker cable is not connected. A headphone is connected to the PHONES jack.	Turn the knob clockwise to a suitable level. Put the unit in the receive mode. Connect the cable to the speaker jack on the IC-701PS. Connect the speaker connection. Unplug the headphone.
4. Sensitivity is low and only strong signals are audible.	RF GAIN Control knob is turned fully Counterclockwise. The antenna feed line is cut or shorted.	Turn the RF Gain control knob fully clockwise. Check the feed line and correct any improper condition.
5. During receive, the "S" meter is higher than the "0" position even when there is no signal.	RF GAIN Control knob is turned fully counter clockwise.	Turn the knob fully clockwise.
6. Only incomprehensible voices are heard during SSB reception.	Receiving the wrong sideband.	Set the MODE switch to the sideband other than the one you are using, either USB or LSB.
7. The tone of the received signal is very high or low pitched.	Maladjustment of the P.B. Tune control.	Set the P.B. Tune control knob to the center (12 o'clock) position.
8. No output power or low output power.	The COMP Control knob is set too low when the Speech Processor is ON. MIC GAIN setting is too low. When SSB is desired, but the MODE switch is in the CW or CW-N position. PTT switch is not functioning due to improper connection of the mic connector. The antenna feed line is cut or shorted.	Turn the Speech Processor OFF, or turn the COMP Control knob clockwise to obtain the proper output power. Turn the MIC GAIN Control clockwise to the center (12 o'clock) position. Set the MODE switch to the SSB (USB or LSB) position. Using a screw driver, slightly spread the prongs of the mic connector for better contact. Check the antenna feed line and correct any improper connection.
9. The Po meter does not function when signals are transmitted.	The SWR switch is set at the SWR position. SWR SET control is not set at the proper level.	Open the Access Cover and move the SWR switch to the SWR SET position. Open the Access Cover and turn the SWR SET control clockwise while watching the Po meter, until the needle moves properly.
10. The receive mode functions properly and your signals are transmitted, but you are unable to make contact with another station.	The VFO Select is in the RA-TB or TA-RB position. The receive frequency is offset from the transmit frequency by the RIT control.	Set the VFO Select switch to the A or B position. Turn OFF the RIT, or set the RIT control knob to the center (12 o'clockw) position.

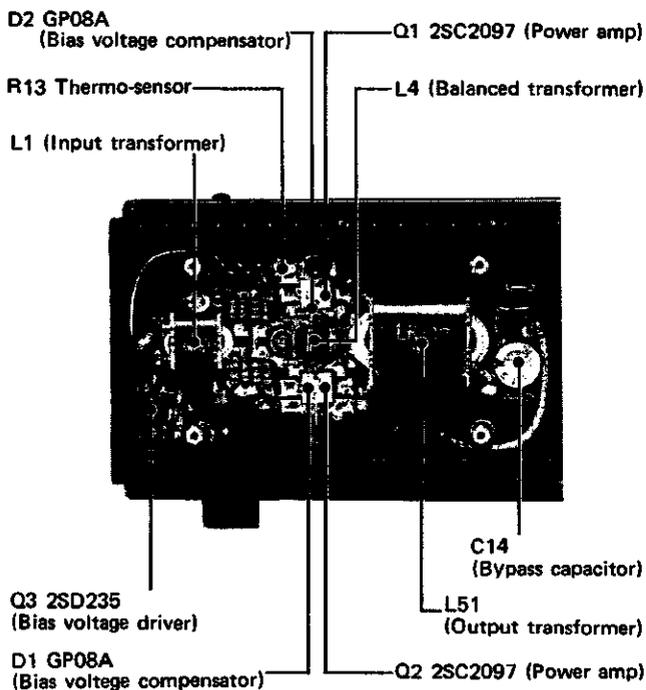
Problem	Possible Cause	Solution
11. The left side of the unit case becomes a little hot.	As the left side of the case is used as a heat sink for the transmitter driver transistor, even during normal operation, it may get as hot as 35°C (63°F) above room temperature.	Make sure that the area around the unit has as much ventilation as possible.
12. The heat sink on the rear panel becomes hot.	The heat sink on the rear panel is for the final transistors and may become as hot as 40°C (72°F) above room temperature.	Make sure the area around the unit has as much ventilation as possible.
13. Hot air suddenly or occasionally eliminates from the heat sink on the rear panel.	The cooling fan has been automatically triggered ON due to the temperature of the final transistors.	This is a normal condition and you may continue operation. However, if the fan starts running at high speed and the frequency display LED's and meter lamps start blinking, discontinue operation immediately, allow the heat sink to cool and check to see that you have the best ventilation possible, or correct a high SWR condition, etc.
14. Frequency display LED's and the meter lamps start blinking, and the fan starts running at high speed.	Insufficient ventilation of the heat sink on the rear panel. Long operation on RTTY or CW. Excessive antenna SWR.	Make sure the heat sink on the rear panel has as much ventilation as possible, after stopping transmit operation and allowing the heat sink to cool. Reduce the RF output power by turning the COMP control knob counterclockwise, or set the unit in the receive mode as quickly as possible. Check the antenna feed line and correct any improper condition of mismatch.
15. The RIT LED goes out even though you did not depress the RIT ON/OFF switch to turn it OFF.	The RIT is automatically turned OFF when the Tuning Knob is rotated. The Power switch has been turned OFF,	You may change the unit so that the automatic RIT OFF function does not operate when you rotate the Tuning knob. (See page 13.) Depress the RIT ON/OFF switch again to turn ON the RIT.
16. The frequency does not change by rotating the Tuning knob.	DIAL LOCK is engaged.	Disengage the DIAL LOCK by pressing the DIAL LOCK button.
17. The scale on the Tuning knob is not in calibration with the displayed frequency.	The DIAL FAST button was depressed when the Tuning knob was set with one of the small increments at the center point, the Tuning knob was rotated when the DIAL LOCK was engaged, or the dial was rotated beyond the band edge.	Set the Tuning dial so that one of the large increments is at the center point, and push the DIAL FAST button TWICE, and the scale will be calibrated.
18. The last two digits of the memorized or displayed frequency has changed to 0.0.	The DIAL FAST button was pressed, or the VFO Select switch was turned while in the DIAL FAST mode.	Release the DIAL FAST function by pressing the DIAL FAST button again, and set the unit to the desired frequency with the Tuning knob.
19. An abnormal figure is displayed on an LED, or an LED fails to illuminate.	The Power switch was turned OFF and ON too quickly.	Turn the Power switch OFF, wait a few seconds, and then turn it ON again.
20. An unusual amount of distortion and/or noise in the transmit signal.	The MIC GAIN is too high when the RF Speech Processor is ON.	While watching the meter in the COMP mode, turn the MIC GAIN control counterclockwise so that the meter indicates approximately 0 dB at the voice peak.
21. Unable to key when using an electronic keyer.	The Switching circuit of the electronic keyer does not close completely. Polarity of the connection wires from the keyer is reversed. The MODE switch is not in the CW or CW-N position.	Replace the keyer with one having a voltage of less than 0.4V between the two keying lines at the key down position, or modify your keyer so that the voltage is less than 0.4 Volts. Connect the keyer wires observing proper polarity. Place the MODE switch in the CW or CW-N position.

SECTION 9 INSIDE VIEW

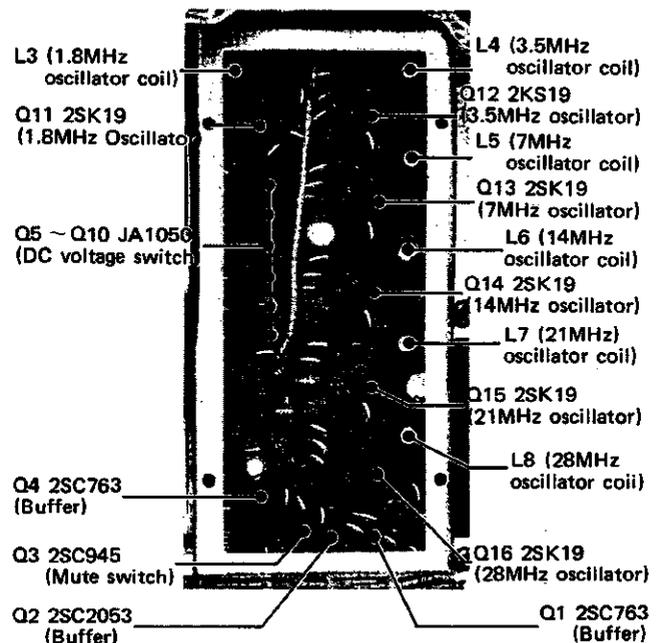
9-1 TOP VIEW



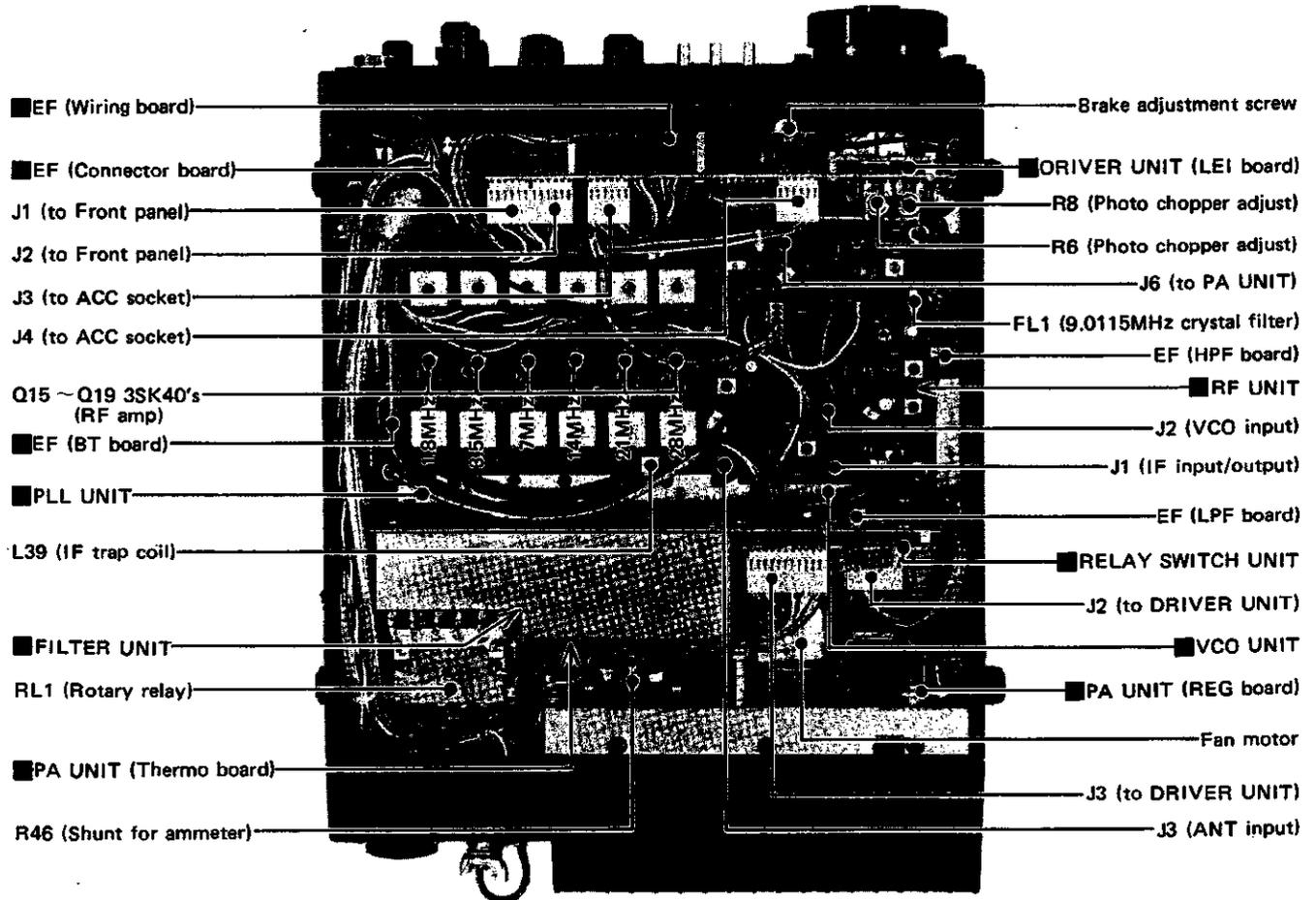
9-2 PA UNIT



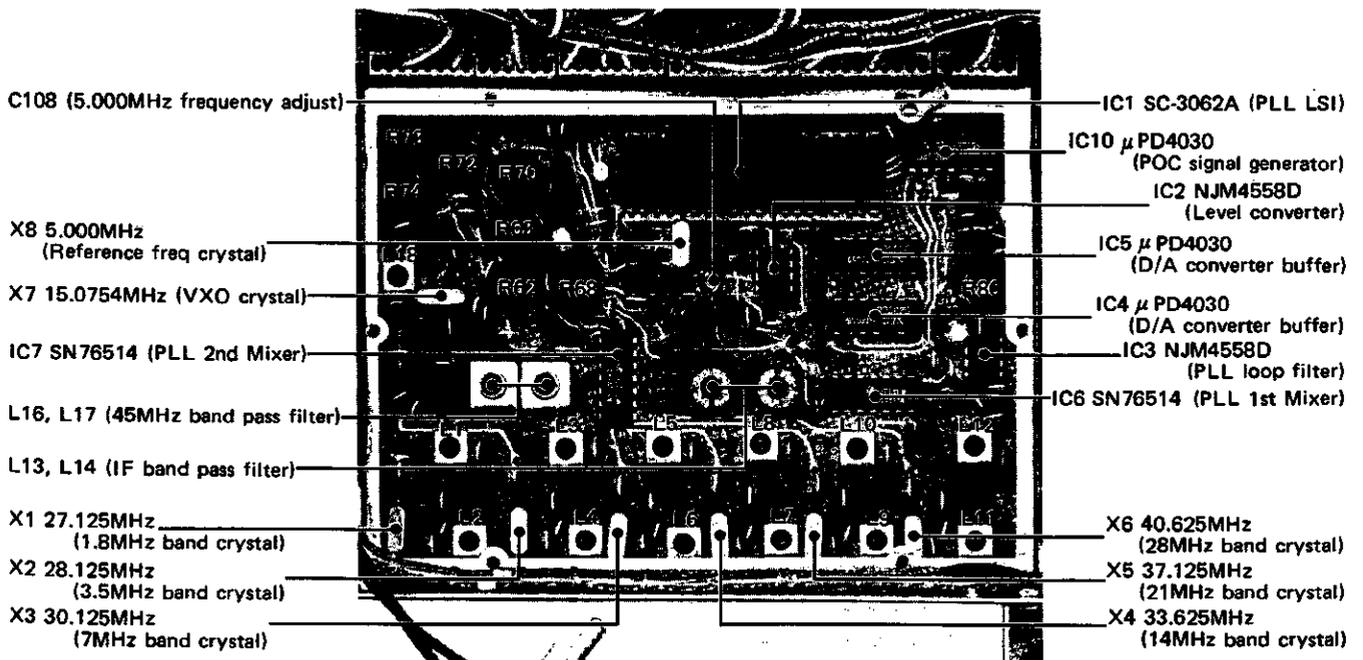
9-3 VCO UNIT



9-4 BOTTOM VIEW



9-5 PLL UNIT



SECTION 10 VOLTAGE CHARTS

10-1-1 "A" UNIT TRANSISTORS

TR No.	TRANSMIT				RECEIVE				REMARKS
	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	
Q 1	2.2		6.9	1.5	2.2		6.9	1.6	USB COMP ON
Q 2	1.7		6.6	1.2	1.7		6.6	1.2	"
Q 3					0.6		1.1	GND	NO SIGNAL INPUT
Q 4	-3.3		6.7	GND					COMP ON
Q 5	0		5.7	GND	0.7		0.2	GND	VOX ON CW KEY ON
Q 6	5.7		6.2	5.0	0		6.4	0	"
Q 7					0.5		2.9	1.3	NO SIG INPUT VOX ON
Q 8					1.9		1.4	1.3	"
Q 9					0.3		8.2	GND	"
Q10					0		5.9	0	
Q11	0	4.4	7.1	0.6	0	3.7	7.2	0.6	COMP ON
Q12	0	4.4	6.9	0.2	0	3.7	7.0	0.2	"
Q13	0.6		2.3	GND					
Q14	0		2.3	GND					
Q15	6.0		4.8	6.7	6.0		4.8	6.7	
Q16	5.6		6.0	5.5	2.6		6.0	5.5	
Q17	3.8		3.4	3.4	3.8		3.4	3.4	
Q18	2.6		3.3	1.4	2.6		3.3	1.4	
Q19	5.4		6.0	5.0	5.4		6.0	5.0	
Q20	6.0		4.7	6.6	6.0		4.7	6.6	
Q21	0		6.6	0	0		6.6	0	NO SIGNAL INPUT
Q22	0		0	6.6	0		0	6.6	
Q23	6.6		0	6.6	6.6		0	6.6	
Q24	0		6.6	0	0		6.6	0	
Q25	0		6.6	0	0		6.6	0	
Q26	1.5		4.7	1.0	1.5		4.7	1.0	
Q27	4.5		7.6	3.9	4.5		7.6	3.9	
Q28	-4.8		0.5	-5.5	-4.8		0.5	-5.5	
Q29	-8.5		8.0	-9.0	-8.5		8.0	-9.0	
Q30	-9.0		-3.0	-9.0	-9.0		-3.0	-9.0	
Q31	-8.5		-9.0	-9.0	-8.5		-9.0	-9.0	
Q32	-8.6		8.0	-9.0	-8.6		8.0	-9.0	
Q33	-9.0		8.0	-7.5	-9.0		8.0	-7.5	
Q34	-7.9		4.4	-8.5	-7.9		3.9	-8.5	
Q35	-8.5		-9.0	-9.0	-8.5		-9.0	-9.0	
Q36	5.5		-9.0	4.4	3.4		-9.0	3.9	
Q37					8.8		12.9	8.2	
Q38	0.7		8.8	0.3	1.7		1.0	0.8	
Q39	8.7		12.8	8.0					
Q40	12.8		8.0	13.4	12.8		8.0	13.4	
Q41	0		5.0	0	0		5.0	0	
Q42	0.5		12.5	GND	0.6		0.5	GND	
Q43	0.6		0.5	GND	0.5		12.8	GND	
Q44					3.3		4.4	2.8	
Q45					4.8		2.6	4.8	
Q46					2.6		4.8	1.9	

10-1-2 "A" UNIT IC IN TRANSMIT MODE

IC No.	PIN No.														REMARKS
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	
IC1	1.5	1.5	E	6.3	6.3										COMP ON
IC2															
IC3	0	7.0	6.3	3.5	2.3	GND	0	0	2.3	3.5	3.5	3.5	6.3	0	COMP ON
IC4	0	7.0	6.3	3.5	2.3	GND	0	0	2.3	3.5	3.5	3.5	6.3	0	COMP ON
IC5	1.8	1.1	1.1	1.8	1.2	6.9	0	3.4	0	3.4	3.4	6.9	0	GND	
IC6	0	0	0	-9	0	0	0	8.0							

10-1-3 "A" UNIT IC IN RECEIVE MODE

IC No.	PIN No.														REMARKS
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	
IC1															
IC2	3.7	1.5	GND	1.0	3.7	5.8	4.3	4.6							
IC3	0	7.1	6.4	3.5	2.3	GND	0	0	2.3	3.5	3.5	3.5	6.4	0	
IC4	0	7.1	6.4	3.5	2.3	GND	0	0	2.3	3.5	3.5	3.5	6.4	0	
IC5															
IC6	0	0	0	-9	0	0	0	8.0							

10-2-1 "B" UNIT TRANSISTORS

TR No.	TRANSMIT				RECEIVE				REMARKS
	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	
Q 1	1.2		13	0.5					
Q 2	0.6		16.7	GND					
Q 3	0.6		16.7	GND					
Q 4	0.7		0	GND					
Q 5	12.7		13.9	13.5	17.0		-8.0	17.0	
Q 6	0.7		0	GND					28MHz
Q 7	1.6		3.3	0.5	1.6		3.3	0.5	CW MONI
Q 8	0		0.4	GND	0		0.4	GND	
Q 9	1.3		13.8	0.7					
Q10	0		-9.0	0					
Q11	0		-8.5	0.5					
Q12	-8.5		1.5	-9.0					
Q13	1.5		-9.0	2.0					
Q14					1.0		13.7	3.4	DIM ON
Q15					4.0		5.2	3.4	"
Q16					5.2		14.0	5.7	"
Q17	0.4		0	0.5	6.3		5.8	7.1	SSB NO SIG INPUT
Q18	2.2		7.0	1.7	2.2		7.0	1.7	
Q19	2.3		7.0	1.7	2.2		7.0	1.8	
Q20	0.6		4.0	GND					RTTY N
	0.6		7.6	GND					RTTY W

10-2-2 "B" UNIT IC IN TRANSMIT MODE

IC No.	PIN No.														REMARKS
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	
IC 1	1.9	14	14	0	0.5	14.1	0	4.2							

10-2-3 "B" UNIT IC IN RECEIVE MODE

IC No.	PIN No.														REMARKS
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	
IC 1	1.9	14	13.4	7.9	6.7	14.1	0	0.9							

10-3 RF UNIT

TR No.	TRANSMIT				RECEIVE				REMARKS
	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	
Q 1	3.9		6.8	3.9					
Q 2	0.5		0	0.5	7.3		7.9	8.1	
Q 3					2.9	0	6.8	0.5	
Q 4	0.2		8.0	GND					
Q 5	0.7		0	GND					
Q 6	1.1		6.4	0.2					
Q 7					0		8.0	0.4	
Q 8					0		4.1	0.4	NB ON
Q 9 (1)					2.5		5.7	1.9	"
Q 9 (2)					2.5		6.4	1.9	"
Q10 (1)					2.5		6.4	1.9	"
Q10 (2)					2.5		7.2	1.9	"
Q11					0		2.5	GND	"
Q12					0		2.7	GND	"
Q13									DELETED
Q14									DELETED
Q15	0	2.8	7.8	1.1	0	2.8	7.9	1.1	
Q16	0	2.8	7.8	1.1	0	2.8	7.9	1.1	
Q17	0	2.8	7.8	1.1	0	2.8	7.9	1.1	
Q18	0	2.8	7.8	0.7	0	2.8	7.9	0.7	
Q19	0	2.8	7.8	1.0	0	2.8	7.9	1.0	
Q20	0	2.8	7.8	0.8	0	2.8	7.9	0.8	

10 - 4 RELAY SWITCH UNIT

TR No.	TRANSMIT				RECEIVE				REMARKS
	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	
Q1 (1)					*		6.2	3.6	7MHz
Q1 (2)					*		6.4	3.6	"
Q2					6.2		8.0	6.4	
Q3					6.4		8.0	6.2	
Q4					▲		8.0	8.1	
Q5					▲		7.4	GND	
Q6					0.7		0	GND	
Q7					0.1		16.4	0.1	
Q8					0.1		16.2	GND	

* High impedance: multimeter does not indicated actual voltage.

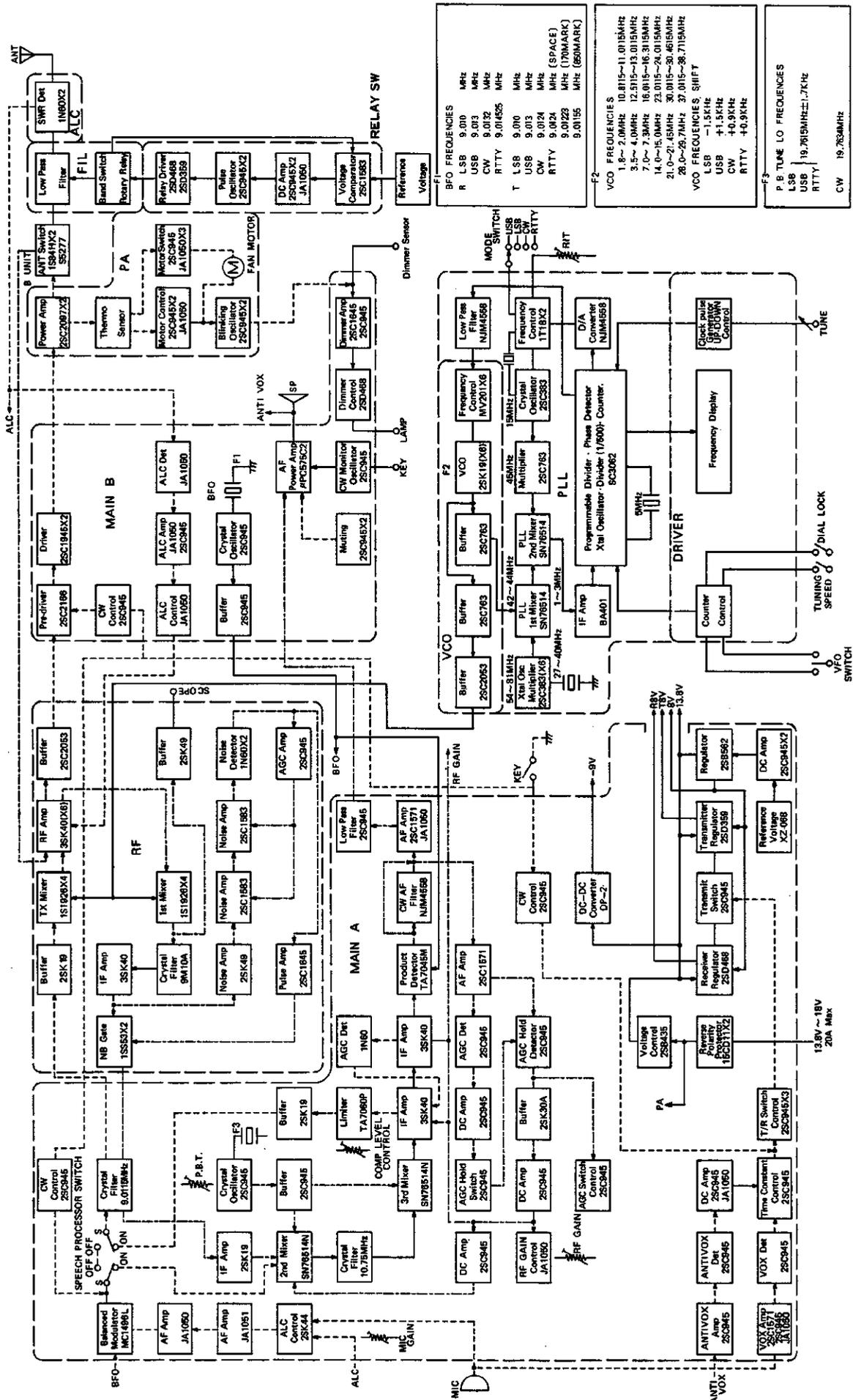
▲ Pulse activated

10 - 5 PA UNIT

TR No.	TRANSMIT				RECEIVE				REMARKS
	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	BASE OR GATE 1	GATE 2	COLLECTOR OR DRAIN	EMITTER OR SOURCE	
Q 1	0.7		16.0	GND					PA BOARD
Q 2	0.7		16.0	GND					"
Q 3	1.3		16.0	0.7					"
Q 4	7.5		8.0	7.9					THERMO BOARD FAN OFF
Q 5	8.0		8.0	8.0					"
Q 6	7.9		13.9	GND					"
Q 7	13.9		0	13.9					"
Q 8	7.5		8.0	8.0					"
Q 9	8.0		13.9	8.0					"
Q10	13.9		0	13.9					"
Q11	0		0	GND					"
Q12	0		0	GND					"
Q13	0.7		0.8	GND					REG BOARD
Q14	0.6		12.2	GND					"
Q15	15.8		14.2	16.6					"

SECTION I | BLOCK DIAGRAM

IC-701 BLOCK DIAGRAM





ICOM INCORPORATED
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